

# High Efficiency Synchronous Step-Up DC/DC Converter

## **FEATURES**

- High Efficiency 93% ( $V_{IN}$ =2.4V,  $V_{OUT}$ =3.3V,  $I_{OUT}$ =200mA).
- Output Current up to 500mA. (AIC1620 at V<sub>IN</sub>=2.4V and V<sub>OUT</sub>=3.3V)
- 20μA Quiescent Supply Current.
- Power-Saving Shutdown Mode (0.1μA typical).
- Internal Synchronous Rectifier ( no external diode )
- · On-Chip Low Battery Detector.
- . Low Battery Hysteresis.
- Space-Saving Package: MSOP-8

## APPLICATIONS

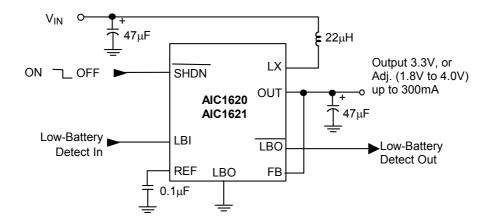
- Palmtop & Notebook Computers.
- PDAs
- · Wireless Phones
- · Pocket Organizers.
- · Digital Cameras.
- Hand-Held Devices with 1 to 2-Cell of NiMH/NiCd Batteries.

# DESCRIPTION

The AIC1620/AIC1621 are high efficiency step up DC-DC converter. The start-up voltage is as low as 0.8V and operate with an input voltage down to 0.7V. Consuming only  $20\mu A$  of quiescent current. These devices offer a built-in synchronous rectifier that reduces size and cost by eliminating the need for an external Schottky diode and improves overall efficiency by minimizing losses.

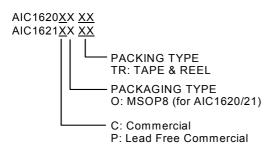
The switching frequency depends on the load and the input voltage can range up to 500KHz. The output voltage can be easily set by two external resistors from 1.8V to 4.0V, connecting FB to OUT to get 3.3V.The peak current of the internal switch is fixed at 0.8A (AIC1620), at 0.45A (AIC1621) for design flexibility.

# TYPICAL APPLICATION CIRCUIT





## ORDERING INFORMATION

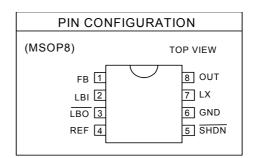


Example: AIC1620COTR

→ In MSOP8 Package & Taping & Reel Packing Type

AIC1620POTR

→ In MSOP8 Lead Free Package & Taping & Reel Packing Type



# ■ ABSOLUTE MAXIMUM RATINGS

Supply Voltage (OUT to GND)	8.0V
Switch Voltage (LX to GND)	V <sub>OUT</sub> + 0.3V
SHDN, LBO to GND	6.0V
LBI, REF, FB, to GND	V <sub>OUT</sub> +0.3V
Switch Current (LX)	-1.5A to +1.5A
Output Current (OUT)	-1.5A to +1.5A
Operating Temperature Range	-40°C ~ +85°C
Maximum Junction Temperature	125°C
Storage Temperature Range	-65°C ~150°C
Lead Temperature (Soldering 10 Sec.)	260°C

Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

# TEST CIRCUIT

Refer to Typical Application Circuit.



# **ELECTRICAL CHARACTERISTICS** (V<sub>IN</sub>=2.0V, V<sub>OUT</sub>=3.3V (FB=OUT), R<sub>L</sub>=∞, T<sub>A</sub>=25°C, unless otherwise specified.) (Note1)

PARAMETER	TEST CONDITIONS		MIN.	TYP.	MAX.	UNIT
Minimum Input Voltage				0.7		V
Operating Voltage			1.1		4.0	V
Start-Up Voltage	R <sub>L</sub> =3KΩ (Note2)			8.0	1.1	V
Start-Up Voltage Tempco				-2		mV/°C
Output Voltage Range			1.8		4.0	
Output Voltage	FB = V <sub>OUT</sub>		3.17	3.3	3.43	V
Steady State Output Current	FB=OUT	AIC1620	300	400		A
(Note 3)	$(V_{OUT} = 3.3V)$	AIC1621	150	220		- mA
Reference Voltage	I <sub>REF</sub> = 0	1.199	1.23	1.261	V	
Reference Voltage Tempco				0.024		mV/°C
Reference Load Regulation	I <sub>REF</sub> = 0 to		10	30	mV	
Reference Line Regulation	V <sub>OUT</sub> = 1.8V to 4V			5	10	mV/V
FB , LBI Input Threshold			1.199	1.23	1.261	V
Internal switch On-Resistance	I <sub>LX</sub> = 100m	ıA		0.3		Ω
LX Switch Current Limit	AIC1620 AIC1621		0.6	0.8	1.0	А
			0.3	0.45	0.6	
LX Leakage Current	V <sub>LX</sub> =0V, 4V; V <sub>OUT</sub> =4V			0.05	1	μА
Operating Current into OUT	V <sub>FB</sub> = 1.4V	' , V <sub>OUT</sub> = 3.3V		20	35	μА
(Note 4)						
Shutdown Current into OUT	SHDN = GND			0.1	1	μА
Efficiency	V <sub>OUT</sub> = 3.3	V ,I <sub>LOAD</sub> = 200mA		90		- %
	V <sub>OUT</sub> = 2V	<sub>UT</sub> = 2V ,I <sub>LOAD</sub> = 1mA 85			70	
LX Switch On-Time	V <sub>FB</sub> =1V , V <sub>OUT</sub> = 3.3V		2	4	7	μS
LX Switch Off-Time	V <sub>FB</sub> =1V , V <sub>OUT</sub> = 3.3V		0.6	0.9	1.3	μS
FB Input Current	V <sub>FB</sub> = 1.4V		0.03	50	nA	



# **ELECTRICAL CHARACTERISTICS** (Continued)

PARAMETER	TEST CONDITIONS	MIN.	TYP.	MAX.	UNIT	
LBI Input Current	V <sub>LBI</sub> = 1.4V		1	50	nA	
SHDN Input Current	V <sub>SHDN</sub> = 0 or V <sub>OUT</sub>		0.07	50	nA	
LBO Low Output Voltage	V <sub>LBI</sub> = 0, I <sub>SINK</sub> = 1mA		0.2	0.4	μΑ	
LBO Off Leakage Current	V <sub>LBO</sub> = 5.5V, V <sub>LBI</sub> = 5.5V		0.07	1		
LBI Hystereisis			50		mV	
SHDN Input Voltage	V <sub>IL</sub>			0.2Vout	V	
	V <sub>IH</sub>	0.8V <sub>OUT</sub>				

- **Note 1:** Specifications are production tested at  $T_A=25^{\circ}\mathbb{C}$ . Specifications over the -40 $^{\circ}\mathbb{C}$  to 85 $^{\circ}\mathbb{C}$  operating temperature range are assured by design, characterization and correlation with Statistical Quality Controls (SQC).
- **Note 2:** Start-up voltage operation is guaranteed without the addition of an external Schottky diode between the input and output.
- **Note 3:** Steady-state output current indicates that the device maintains output voltage regulation under load.
- **Note 4:** Device is bootstrapped (power to the IC comes from OUT). This correlates directly with the actual battery supply.

## TYPICAL PERFORMANCE CHARACTERISTICS

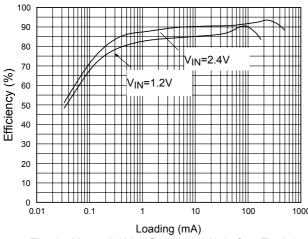


Fig. 1 V<sub>OUT</sub>=3.3V, AIC1620 (0.8A) (ref. to Fig.14)

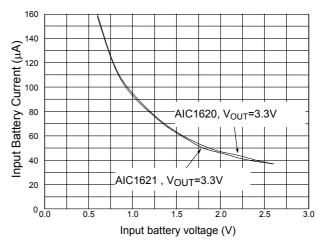


Fig. 2 No-Load Battery Current vs. Input Battery



# TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

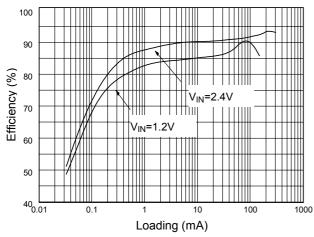


Fig. 3 V<sub>OUT</sub>=3.3V, AIC1621 (0.45A) (ref. to Fig.14)

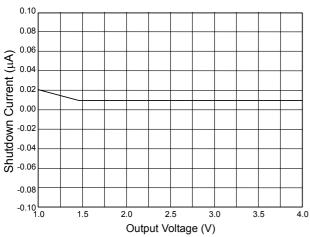
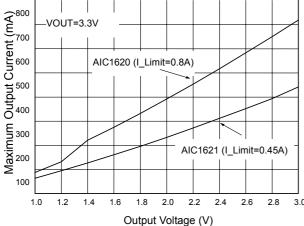


Fig. 5 Shutdown Current vs. Output Voltage



Output Voltage (V)
Fig.7 Maximum Output Current vs. Input Voltage

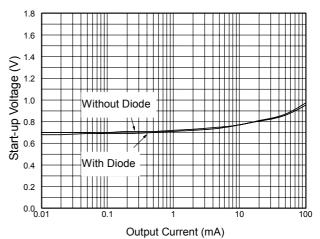


Fig. 4 Start-up Voltage vs. Output Current

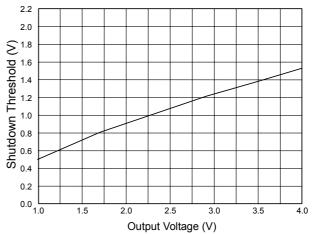


Fig. 6 Shutdown Threshold vs. Output Voltage

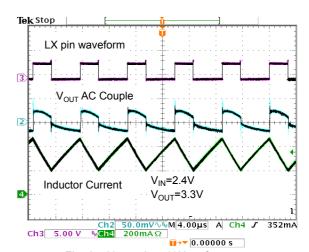
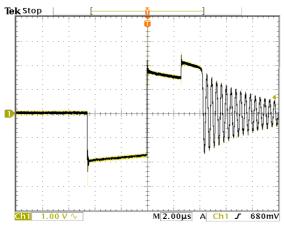


Fig. 8 Heavy Load Waveform



# TYPICAL PERFORMANCE CHARACTERISTICS (Continued)





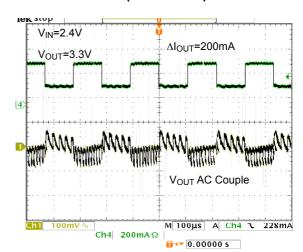
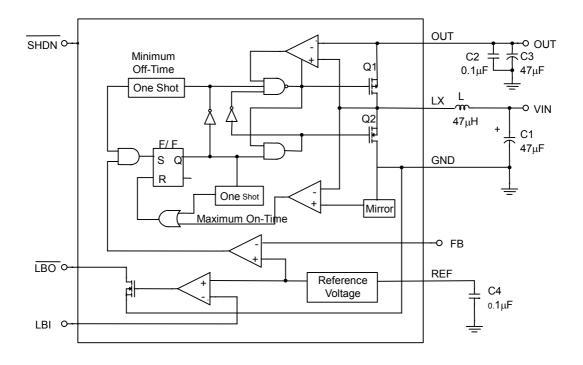


Fig. 10 Load Transient Response

# BLOCK DIAGRAM





## ■ PIN DESCRIPTIONS

#### AIC1620/ AIC1621

PIN 1: FB- Connect to OUT for +3.3V output.

Use a resistor network to set the output voltage from +1.8V to +4.0V.

PIN 2: LBI- Low-battery comparator input. In-

ternally set to trip at +1.23V.

PIN 3: LBO- Open-drain low battery comparator

output. Output is low when  $V_{LBI}$  is <1.23V.  $\overline{LBO}$  is high impedance

during shutdown.

PIN 4: REF- 1.23V reference voltage. Bypass with a  $0.1\mu F$  capacitor.

PIN 5: SHDN- Shutdown input. High=operating, low=shutdown.

PIN 6: GND- Ground

PIN 7: LX- N-channel and P-channel power

MOSFET drain.

PIN 8: OUT- Power output. OUT provides boot-

strap power to the IC.

### APPLICATION INFORMATION

#### Overview

AIC1620/AIC1621 series are high efficiency, stepup DC-DC converters, designed to feature a built-in synchronous rectifier, which reduces size and cost by eliminating the need for an external Schottky diode. The start-up voltage of AIC1620/AIC1621 is as low as 0.8V and it operates with an input voltage down to 0.7V. Quiescent supply current is only  $20\mu A$ . The internal P-MOSFET on-resistance is typically  $0.3\Omega$  to improve overall efficiency by minimizing AC losses. The output voltage can be easily set by two external resistors from 1.8V to 4.0V, connecting FB to OUT to get 3.3V. The peak current of the internal switch is fixed at 0.8A (AIC1620) or 0.45A (AIC1621) for design flexibility. The current limit of AIC1620 and AIC1621 are 0.8A and 0.45A respectively. The lower current limit allows the use of a physically smaller inductor in space-sensitive applications.

#### **PFM Control Scheme**

The key feature of the AIC1620 series is a unique minimum-off-time, constant-on-time, current-limited, pulse-frequency-modulation (PFM) control scheme (see BLOCK DIAGRAM) with the ultra-low quiescent current. The peak current of the internal N-MOSFET power switch can be fixed at 0.8A (AIC1620) or 0.45A (AIC1621). The switch frequency depends on either loading condition or input voltage, and can range up to 500KHz. It is

voltage, and can range up to 500KHz. It is governed by a pair of one-shots that set a minimum off-time  $(1\mu S)$  and a maximum on-time  $(4\mu S)$ .

## **Synchronous Rectification**

Using the internal synchronous rectifier eliminates the need for an external Schottky diode. Therefore, the cost and board space is reduced. During the cycle of off-time, the P-MOSFET turns on and shunts the N-MOSFET. Due to the low turn-on resistance of MOSFET, the synchronous rectifier significantly improves efficiency without the addition of an external component. Thus, the conversion efficiency can be as high as 93%.

#### Reference Voltage

The reference voltage (REF) is nominally 1.23V for excellent T.C. performance. In addition, REF pin can source up to  $100\mu A$  to external circuit with good load regulation (<10mV). A bypass capacitor of  $0.1\mu F$  is required for proper operation and good performance



#### **Shutdown**

The whole circuit is shutdown when  $V_{\overline{SHDN}}$  is low. At shutdown mode, the current can flow from battery to output due to body diode of the P-MOSFET.  $V_{OUT}$  falls to approximately Vin-0.6V and LX remains high impedance. The capacitance and load at OUT determine the rate at which  $V_{OUT}$  decays. Shutdown can be pulled as high as 6V. Regardless of the voltage at OUT.

#### **Selecting the Output Voltage**

 $V_{\text{OUT}}$  can be simply set to 3.3V by connecting FB pin to OUT due to the use of internal resistor divider in the IC (Fig. 14). In order to adjust output voltage, a resistor divider is connected to  $V_{\text{OUT}}$ , FB, GND (Fig. 15). Vout can be calculated by the following equation:

R5=R6 [(V<sub>OUT</sub> / V<sub>REF</sub> )-1]

Where  $V_{REF}$  =1.23V and  $V_{OUT}$  may range from 1.8V to 4V. The recommended R6 is 240K $\Omega$ .

## **Low-Battery Detection**

AIC1620 series contains an on-chip comparator with 50mV internal hysteresis (REF, REF+50mV) for low battery detection. If the voltage at LBI falls below the internal reference voltage. LBO ( an open-drain output) sinks current to GND.

#### Component Selection

#### 1. Inductor Selection

An inductor value of  $22\mu H$  performs well in most applications. The AIC1620 series also work with inductors in the  $10\mu H$  to  $47\mu H$  range. An inductor with higher peak inductor current tends a higher output voltage ripple (I<sub>PEAK</sub>×output filter capacitor ESR). The inductor's DC resistance significantly affects efficiency. We can calculate the maximum output current as follows:

$$I_{OUT(MAX)} = \frac{V_{IN}}{V_{OUT}} \Bigg[ I_{LIM} - t_{OFF} \Bigg( \frac{V_{OUT} - V_{IN}}{2 \times L} \Bigg) \Bigg] \eta$$

where  $I_{OUT(MAX)}$ =maximum output current in amps

 $V_{IN}$ =input voltage L=inductor value in  $\mu$ H  $\eta$  =efficiency (typically 0.9)  $t_{OFF}$ =LX switch' off-time in  $\mu$ S  $I_{LIM}$ =0.45A or 0.8A

#### 2. Capacitor Selection

The output voltage ripple relates with the peak inductor current and the output capacitor ESR. Besides output ripple voltage, the output ripple current also needs to be concerned. The smaller the capacitor ESR is, the higher the ripple current will be. A filter capacitor with low ESR is helpful to the efficiency and steady state output current of AlC1620 series. Therefore NIPPON tantalum capacitor MCM series with  $100\mu\text{F/6V}$  is recommended. A smaller capacitor (down to 47  $\mu$  F with higher ESR) is acceptable for light loads or in applications that can tolerate higher output ripple.

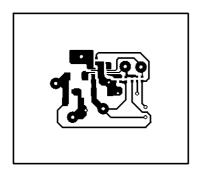
#### 3. PCB Layout and Grounding

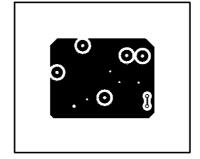
Since AIC1620's switching frequency can range up to 500kHz, it makes AIC1620 become very sensitive. So careful printed circuit layout is important for minimizing ground bounce and noise. IC's OUT pin should be as clear as possible. And the GND pin should be placed close to the ground plane. Keep the IC's GND pin and the ground leads of the input and output filter capacitors less than 0.2in (5mm) apart. In addition, keep all connection to the FB and LX pins as short as possible. In particular, when using external feedback resistors, locate them as close to the FB as possible. To maximize output power and efficiency and minimize output ripple voltage, use a ground plane and solder the IC's



GND directly to the ground plane. Following are

the recommended layout diagrams.





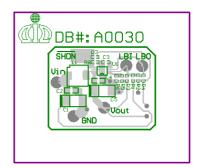
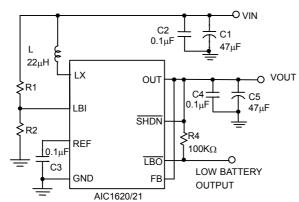


Fig.11 Top layer

Fig.12 Bottom layer

Fig.13 Placement

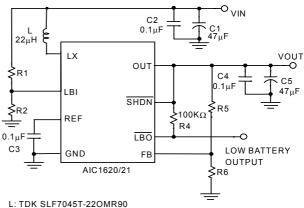
# APPLICATION EXAMPLES



L: TDK SLF7045T-22OMR90

C1, C5: NIPPON Tantalum Capacitor 6MCM476MB2TER

Fig.14  $V_{OUT} = 3.3V$  Application Circuit.



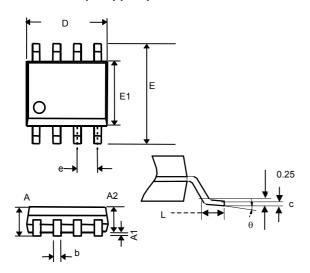
C1, C5: NIPPON Tantalum Capacitor 6MCM476MB2TER  $V_{\text{OUT}} = V_{\text{REF}}^* (1+\text{R5}/\text{R6})$ 

Fig.15 An Adjustable Output Application Circuit



# PHYSICAL DIMENSION (unit: mm)

# • 8 LEAD MSOP (CO)(PO)



MIN	MAX	
- 1.10		
0.05	0.15	
0.75	0.95	
0.25	0.40	
0.13	0.23	
2.90	3.10	
4.90 BSC		
2.90	3.10	
0.65 BSC		
0.40	0.70	
0°	0° 6°	
	- 0.05 0.75 0.25 0.13 2.90 4.90 2.90 0.65 0.40	

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