



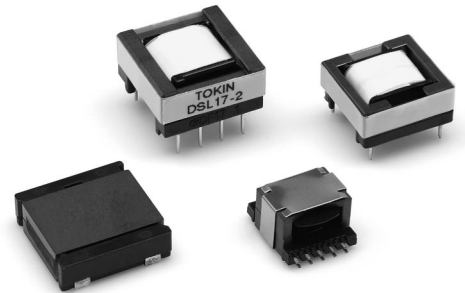
Devices thru Material Innovation

NEC/TOKIN

Vol.

02

# Transformers



TRANSFORMERS

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# INTRODUCTION

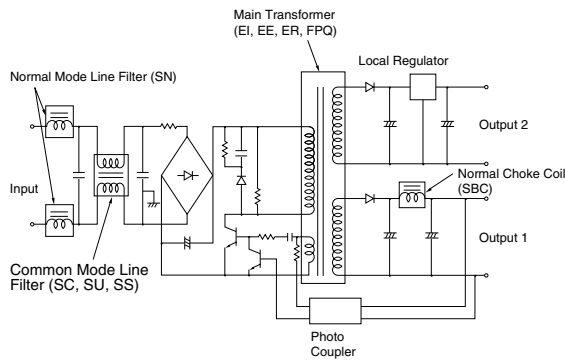
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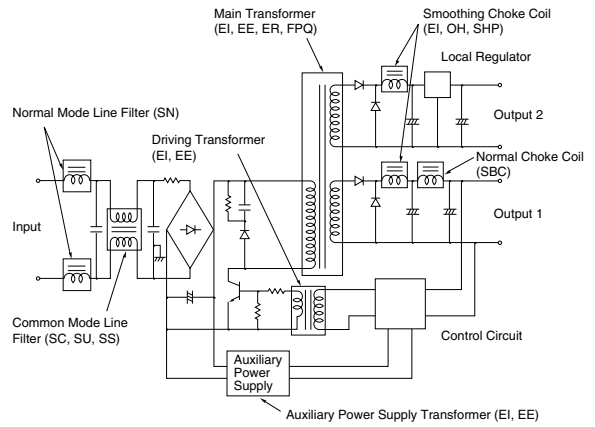
As semiconductor devices become ever more advanced, the demand for thinner, more compact devices with higher efficiency and functionality are increasing. Owing to technologies such as high-density mounting for the switching power supplies, requirements for transformers and chokes are becoming increasingly rigorous. Under the motto "Reliability based on high quality material," NEC TOKIN uses selected excellent materials to provide diverse transformers and choke coils that can meet the requirements for a wide range of applications, such as small toroidal structure types with little heat effect and minimal emission noise to peripheral parts.

# Example of Use

**Forward Method**



**Flyback Method**



# Materials for Design

## Design Method

At present, there are two major types of circuits for the typical switching power supply units: the forward converter method and the flyback method. (See Figure 1.)

The following section introduces the design method of the high frequency transformers for each of the two types mentioned above.

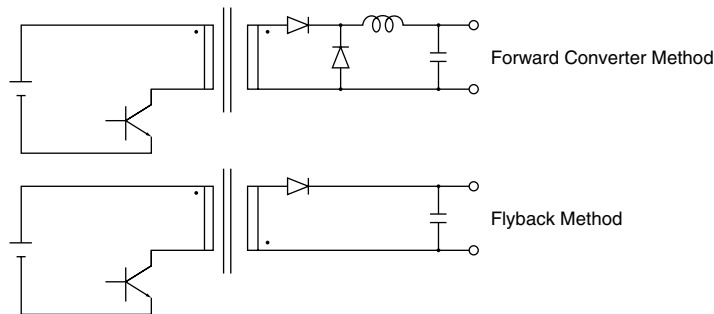


Figure 1 Circuit Method of Switching Power Supply

(1)Forward Converter Method  
1)Transformers

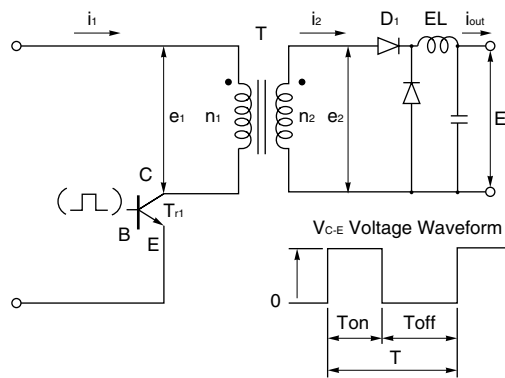


Figure 2 Circuit Diagram of Forward Converter

Figure 2 is the basic circuit diagram of the forward converter method.

When the bias (pulse) of forward direction is applied to the base of the switching transistor ( $Tr_1$ ),  $Tr_1$  is ON,  $e_1$  (V) is impressed to the primary windings  $n_1$  of the transformer "T", and at the same time the voltage as found by the following formula is generated to the secondary windings  $n_2$ :

$$e_2 = \frac{n_2}{n_1} \times e_1(V) \dots\dots\dots \textcircled{1}$$

- $e_1$ : Input voltage of transformer
- $e_2$ : Output voltage of transformer
- $n_1$ : Number of primary windings
- $n_2$ : Number of secondary windings

Therefore, the output voltage is found by the ratio of  $n_1$  and  $n_2$ . Also, from the law of equal ampere turns, the following formula is obtained:

$$i_2 = \frac{n_1}{n_2} \times i_1 \text{ (A)} \dots\dots\dots \textcircled{2}$$

The primary winding is determined by the following formula:

$$n_1 = \frac{e_1 \times T_{on}}{\Delta B \times A_e} \times 10^4 \dots\dots\dots \textcircled{3}$$

- $T_{on}$ : Transistor ( $T_{r1}$ ) "on" time (sec)
- $\Delta B$ : Usage magnetic flux density (T)
- $A_e$ : Core effective cross-section ( $\text{cm}^2$ )

At this point it is important to be aware of the value of  $\Delta B$ . It must be set especially carefully, because when the magnetic flux is saturated, the inductance drops abruptly, in addition to the core loss and temperature rise. The magnetic flux density available for use is within the range from effective saturation magnetic flux density ( $B_{rms}$ ) to the effective saturation residual coercive force ( $B_{rms}$ ). However, it is necessary to set this value  $\Delta B$  as shown below, considering the inevitable factors such as calorific value of the core:

1. Set the upper limit of allowable temperature of core heat.
2. The energy (Wattage) equivalent to the core loss at that time (set in the previous step 1).
3. The next time that same loss occurs and at what value (T).

According to the above stated procedure, the appropriate  $\Delta B$  can be set.

The characteristics of the NEC TOKIN BH2 compound are shown in Figures 3 and 4.

Then, determine the output voltage " $e_2$ " of the transformer from the desired output voltage " $E_o$ ".

$$e_2: \frac{E_o}{\text{duty}} + (e_d + e_l) \dots\dots\dots \textcircled{4}$$

- duty:  $T_{on} / T$   $T = T_{on} + T_{off}$
- $e_d$ : Output rectification diode loss voltage (V)
- $e_l$ : Line loss voltage (V)

The number of secondary windings is determined by formula  $\textcircled{1}$ .

$$n_2 = \frac{e_2 \cdot n_1}{e_1}$$

When the number of secondary windings is determined, again modify the value of " $n_1$ ", the number of primary windings:

$$n_1 = \frac{e_1 \cdot n_2}{e_2} \dots\dots\dots \textcircled{5}$$

At that time, it may be very convenient if the values  $\Delta B$ , core loss and temperature rise of core are found.

Then determine the current  $i_2$  applied to the secondary windings:

$$i_2 = \frac{i_{out}}{\text{duty}} \dots\dots\dots \textcircled{6}$$

The primary current is determined by formula w.

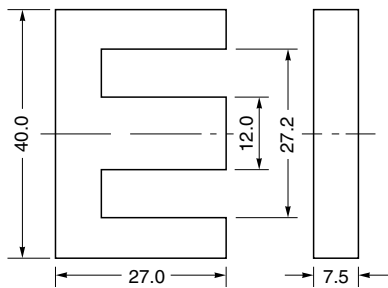
$$i_1 = \frac{n_2 \cdot i_2}{n_1}$$

The material of windings is determined by the following formula:

$$d\phi = \sqrt{\frac{4 \cdot i}{\delta \cdot \pi}} \dots\dots\dots \textcircled{7}$$

- dφ : Diameter of winding (mmφ)
- δ : Density of current (A/mm<sup>2</sup>)
- i : Average current (A)

When the current is large, the winding becomes very thick. Considering the bobbin structure and the efficiency of work, it is more convenient to use numerous windings with the diameter not exceeding 1.0. The higher the frequency used, the greater the loss by the skin effect of the windings. Therefore, applying many of the thinner windings instead of using the single thick winding is recommended.



Core Constant	Σ ℓ /A	cm <sup>-1</sup>	5.19
Effective Cross-section Area	Ae	cm <sup>2</sup>	1.48
Effective Magnetic Circuit Length	ℓ e	cm	7.68
Effective Volume	Ve	cm <sup>3</sup>	11.4
Cross-section Area of Middle Leg	Acp	cm <sup>2</sup>	1.36
Core Frame Area	Acw	cm <sup>2</sup>	1.63
Weight		g/set	61.0
AL		nH	4750

Figure 3 FEI40 (BH2) Core Constant

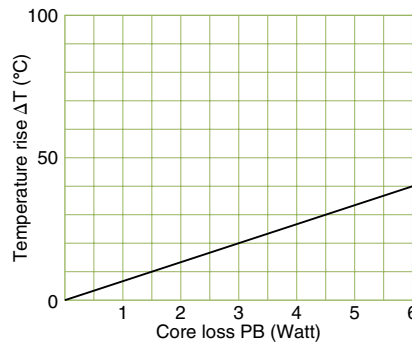


Figure 4 FEI40 core loss - temperature characteristics

2)Choke coils

Because the switching regulator's secondary side choking circuit is superimposed with direct current, a choke coil with good direct current superimposition characteristics must be selected to prevent saturation of the core. Therefore, for designing choke coils the data on the relationship between the gap and AL value is required.

For example, if the gap of the FEI40 (BH2) is 0.2 mm, then from Figure 5 showing the relationship of the FEI40's air gap AL, the AL value is 800nH. In this case, from the direct current superimposition characteristics it is understood that the range in

which the direct current magnetic field's magnetic permeability does not decrease is 40AT. For making a choke coil, the following formula must apply between the actual number of windings  $N_1$  and the maximum direct current superimposition current  $I_0$ :

$$NI > N_1 I_0 \dots\dots\dots \textcircled{8}$$

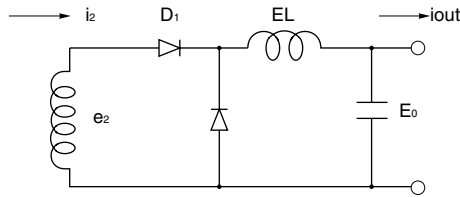
Thus, in this case the number of windings and current capacity are selected according to  $40AT > N_1 I_0$ .

On the other hand inductance is represented by the following formula:

$$L = AL \cdot N^2 \cdot 10^{-9} \text{ [H]} \dots\dots\dots \textcircled{9}$$

$$N = \sqrt{\frac{L \cdot 10^9}{AL}} \text{ (Turn)} \dots\dots\dots \textcircled{10}$$

When choosing the choke with the switching regulator, considering from the previous example :



$$eL = e2 - ed - Eo$$

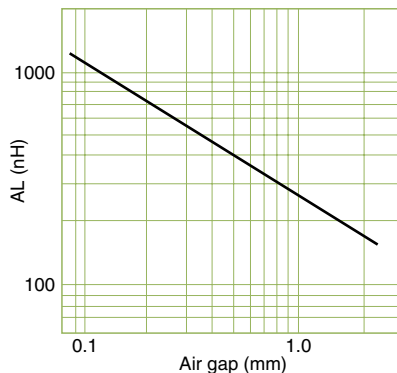
$$= L \frac{di}{dt}$$

$$\therefore L = \frac{eLdt}{di} = \frac{eL \cdot Ton}{iout}$$

Here, on account of the ripple and dummy load of  $iout$ , general guidelines for the inductance of the output choke are derived using the following formula:

$$L = \frac{5Eo \cdot Ton}{iout \text{ (max)}} \text{ [H]}$$

When a 0.1 mm gap is placed in the core's middle leg, the total gap is 0.1 mm, but when a 0.1 mm gap material is inserted overall, the total gap amount becomes 0.2 mm.



**Figure 5 Air gap-AL characteristics (FEI40, BH2)**



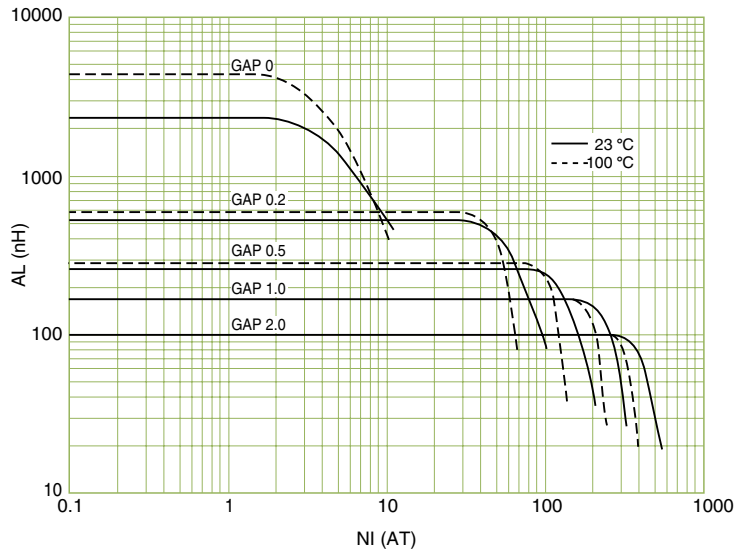


Figure 6 Direct current superimposition characteristics (FEI40, BH2)

(2) Flyback Method

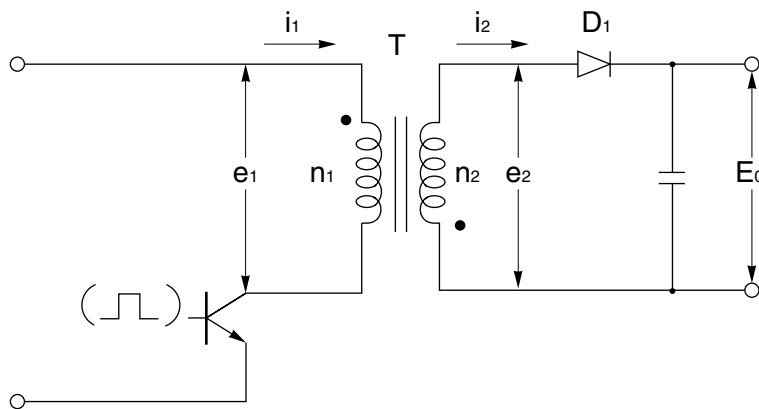


Figure 7 Circuit diagram of flyback method

The above Figure 7 is the circuit diagram of the flyback method. It is almost the same as the previous Figure 2 except for the polarity of the transformer. As the diode D<sub>1</sub> is located in the backward direction on the secondary output circuit, nothing is output when the switching transistor Tr<sub>1</sub> is turned ON. At that time, the following amount of energy is charged in the transformer T.

$$E = 1/2 Li^2 \dots \dots \dots \textcircled{11}$$

L: Primary inductance (H)

The above energy is emitted to the load R by way of the secondary diode D<sub>1</sub> when Tr<sub>1</sub> is turned OFF.

Determine the primary average current.

$$i_1 = \frac{E_o \cdot I_o}{e_1 \cdot \eta} \dots \dots \dots \textcircled{12}$$

- E<sub>o</sub> : Output voltage
- I<sub>o</sub> : Output current
- η : Energy conversion rate

Determine the primary peak current  $i_1 \text{ max.}$  by the following formula:

$$i_1 \text{ max.} = \frac{2 \cdot i_1 \cdot T}{T_{on}} \dots\dots\dots (13)$$

T: Driving cycle of switching (sec)

The inductance "Lp" necessary for the primary windings "n1" is determined by the following formula:

$$L_p \text{ min.} = \frac{e_1 \text{ min.} \cdot T_{on} \text{ max.}}{i_1 \text{ max.}} \text{ [H]} \dots\dots\dots (14)$$

The number of primary windings "n1" is determined by the following formula:

$$n_1 = \frac{L_p \text{ min.} \cdot i_1 \text{ max.}}{A_e \cdot \Delta B} \cdot 10^4 \dots\dots\dots (15)$$

In this case, a certain Gap is provided by the core. Therefore, its hysteresis curve becomes linear, unlike the nonlinear one for Gap = 0, and the value  $\Delta B$  can be kept somewhat larger than the case using the forward method.

The number of secondary windings "n2" is determined by the following formula:

$$n_2 = \frac{n_1 \cdot (E_o + e_d + e_l)}{e_1} \cdot \frac{T_{off}}{T_{on}} \dots\dots\dots (16)$$

$e_d$  : Secondary side diode loss voltage (V)

$e_l$  : Line loss voltage (V)

In the RCC method, the feedback winding is determined by the voltage "Ez" of the driving voltage, as shown in the following formula:

$$n_d = \frac{E_z + E_{BE}}{e_2 \cdot n_2} \dots\dots\dots (17)$$

The designing procedure has been completed by the above steps. However, a slight correction is actually required because of the differences from the various conditions initially set together with the linkage inductance of the transformer, floating capacity and transformer connecting conditions. In recent years, the driving frequency has risen.

$$N_p = \frac{e_1 T_{on}}{\Delta B \cdot A_e} \times 10^4 \dots\dots\dots (18)$$

The higher it becomes, the fewer windings are required to enable the transformer to be downsized, as shown in the above formula. However, loss could increase by the skin effect of the winding material, as it does for the high frequency. In addition, it might be difficult to cope with the safety standards of each country. Therefore, it is necessary to select the method that is the most appropriate for the various conditions required for each transformer when selecting the winding materials, winding order, winding method and insulation structure.

To select the core, attention must be paid to the following items.

1. The magnetic flux density is to be high.
2. The core loss is to be low.

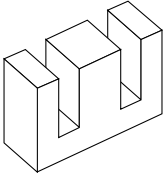
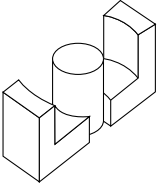
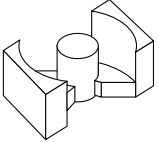
- 3. The magnetic permeability is not to decrease within the driving frequency range.
- 4 The Curie temperature is to be high.
- 5 The saturation magnetic flux density is not to be abruptly changed by the temperature.

Table 1 shows the material characteristics of TOKIN's ferrite cores.

**Table 1 Ferrite Core Material Characteristics**

Material Characteristics		Unit	BH1	BH2	5000B
Applied Frequency Range		MHz	<0.3	<0.3	
Initial Permeability	$\mu_i$		2300±20%	2300±20%	5000±20%
Effective Saturation Magnetic Flux Density (Approx.1200 A/m)	B <sub>ms</sub>	23°C	520	510	500
		100°C	410	400	360
Effective Retentivity	B <sub>rms</sub>	23°C	100	100	120
		100°C	55	55	90
Effective Coercivity	H <sub>cms</sub>	23°C	13	14.3	11.0
		100°C	5		6.5
Curie Temperature	T <sub>c</sub>	°C	220	220	180
Core Loss	100kHz 200mT	23°C	550	600	650
		60°C	350	450	500
		100°C	250	410	800
	500kHz 200mT	23°C			
		60°C			
		100°C			
1MHz 50mT	P <sub>cv</sub>	60°C			
Density	d	kg/m <sup>3</sup>	4.8 × 10 <sup>3</sup>	4.8 × 10 <sup>3</sup>	4.8 × 10 <sup>3</sup>

# Transformers/Choke Coils Series

Series	Shape of Core	Output Wattage			
		Forward Method		Flyback Method	
		50kHz(W)	100kHz(W)	50kHz(W)	100kHz(W)
<b>FEI•FEE</b> 	<b>FEI12.5</b>	3 to 8	4 to 10	2 to 5	3 to 6
	<b>FEI16</b>	10 to 15	13 to 19	3 to 8	4 to 10
	<b>FEE16</b>				
	<b>FEI19</b>	12 to 18	15 to 23	5 to 10	6 to 13
	<b>FEE19</b>				
	<b>FEI22</b>	15 to 20	19 to 26	8 to 15	10 to 19
	<b>FEE22</b>				
	<b>FEI22S</b>	15 to 20	19 to 26	8 to 15	10 to 19
	<b>FEI25</b>				
	<b>FEI28</b>	30 to 50	40 to 65	20 to 30	25 to 40
	<b>FEI30</b>				
	<b>FEE30</b>	50 to 70	65 to 90	30 to 40	40 to 50
	<b>FEI33</b>				
	<b>FEE33</b>	80 to 130	100 to 165	35 to 50	45 to 65
	<b>FEI35S</b>				
	<b>FEI40</b>	100 to 150	130 to 195	45 to 75	60 to 95
<b>FEE40</b>					
<b>FEER</b> 	<b>FEER25.5</b>	20 to 30	26 to 39	10 to 20	13 to 26
	<b>FEER28</b>	35 to 45	45 to 55	20 to 30	26 to 39
	<b>FEER28L</b>				
	<b>FEIR30</b>	30 to 50	40 to 65	25 to 35	33 to 45
	<b>FEER35</b>				
	<b>FEER35L</b>	100 to 150	130 to 195	50 to 65	65 to 80
	<b>FEER39L</b>				
	<b>FEER40</b>	140 to 220	180 to 285	75 to 95	100 to 120
	<b>FPQ</b> 	<b>FPQ2016-T-22</b>	20 to 30	26 to 39	10 to 20
<b>FPQ2020-T-22</b>		25 to 35	32 to 45	15 to 25	19 to 32
<b>FPQ2620-T-22</b>		45 to 60	60 to 75	25 to 35	32 to 45
<b>FPQ2625-T-22</b>		50 to 70	65 to 90	30 to 40	40 to 50
<b>FPQ3220-T-22</b>		50 to 70	65 to 90	30 to 40	40 to 50
<b>FPQ3230-T-22</b>		100 to 150	130 to 195	45 to 60	60 to 75
<b>FPQ3535-T-22</b>		130 to 180	170 to 230	70 to 80	90 to 100

Note: The output wattage is specified for conditions using TOKIN's BH2 material and the temperature rise of the transformer being  $\Delta T < 45^\circ\text{C}$  within range of the operating flux density.

## List of Core Set Shapes

### Ordering Code System

**EB 40 - P 12 12 - F**  
 ① ② ③ ④ ⑤ ⑥

- ① Series
  - EB .... Bobbins for FEI and FEE Cores
  - ERB .. Bobbins for FEER Cores
  - PQB .. Bobbins for FPQ Cores
- ② Size of Core
- ③ Type of Pin
- ④ Type of Placing 11: Horizontal Type, 12: Vertical Type
- ⑤ Number of Pins
- ⑥ Material F: Phenor Resin

### Description of Abbreviations

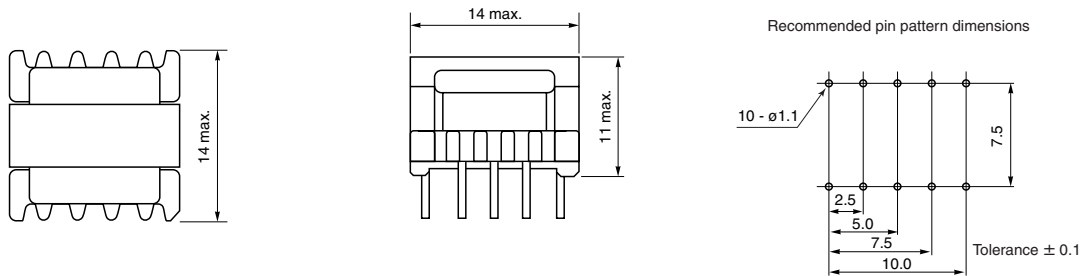
Ae : Section Area of Core (cm<sup>2</sup>)  
 W : Weight of Core (g/Set)

The dimension without the specification of tolerance indicates the typical value.

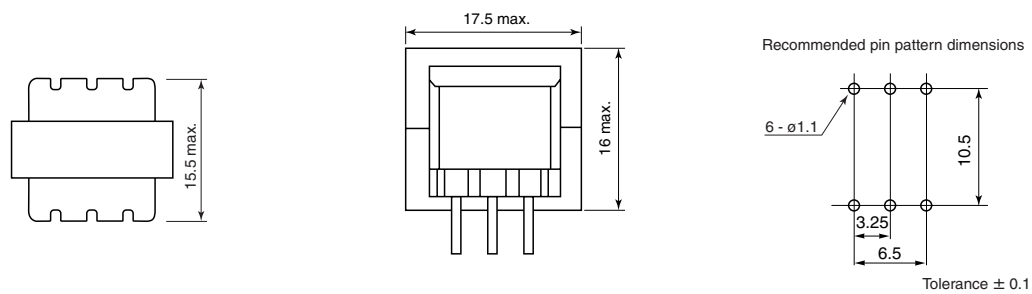
### Outline drawing

[mm] as the unit for dimensions not specified otherwise.  
 The top view is shown for the pin pattern dimensions not specified otherwise.

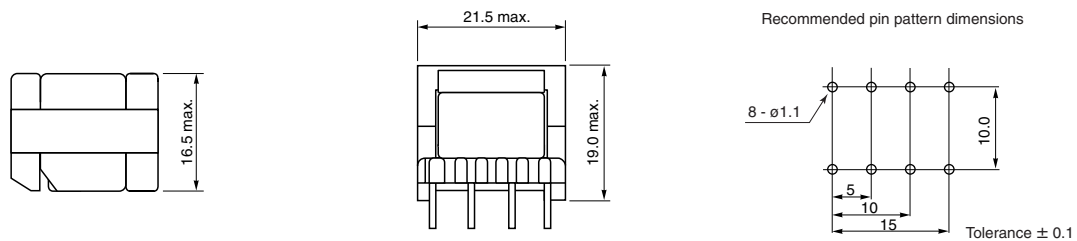
article name	Core Ae	W	Bobbin
FEI12.5	0.15	1.9	EB12.5-P1210-F



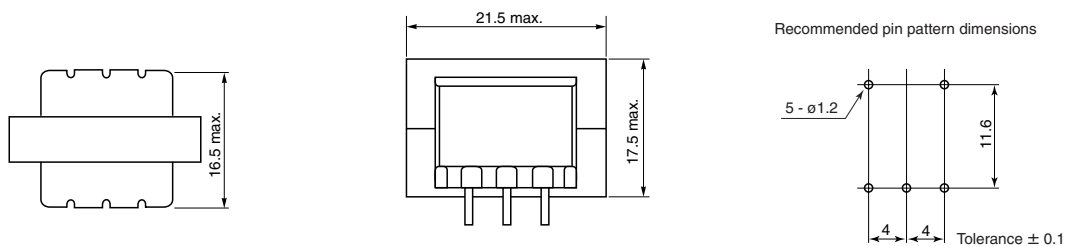
article name	Core Ae	W	Bobbin
FEI16	0.19	3.2	EB16-P1206-F
FEE16	0.19	3.3	



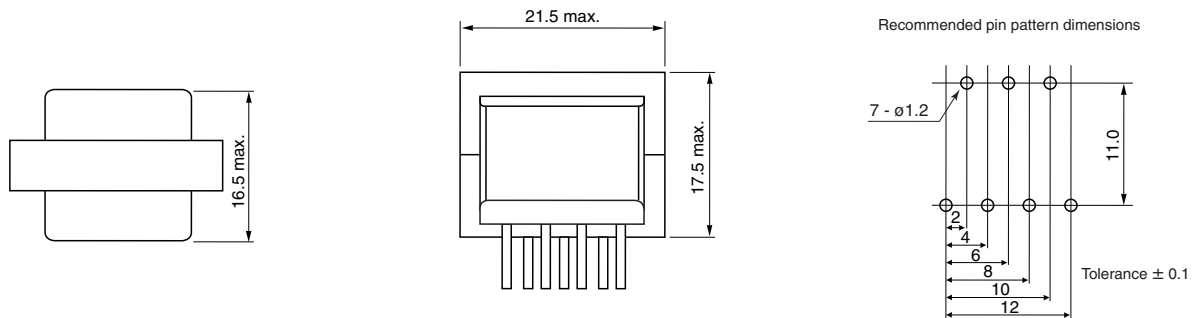
article name	Core Ae	W	Bobbin
FEI19	0.23	4.4	EB19-P1208-F
FEE19	0.23	4.8	



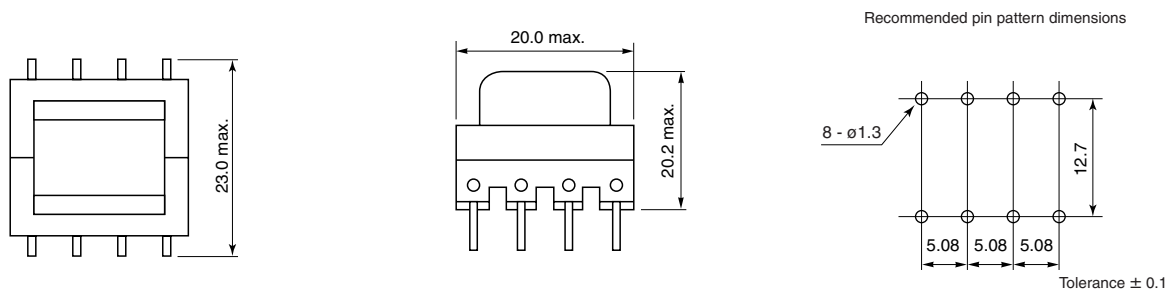
article name	Core Ae	W	Bobbin
FEI19	0.23	4.4	EB19-P1205-F
FEE19	0.23	4.8	



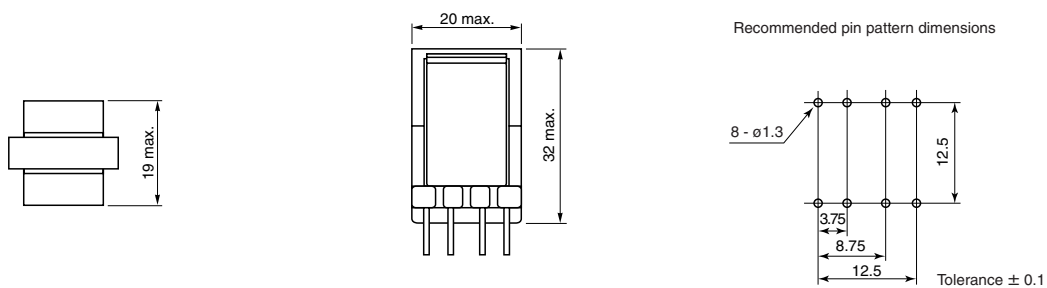
article name	Core Ae	W	Bobbin
FEI19	0.23	4.4	EB19-P1207-F
FEE19	0.23	4.8	



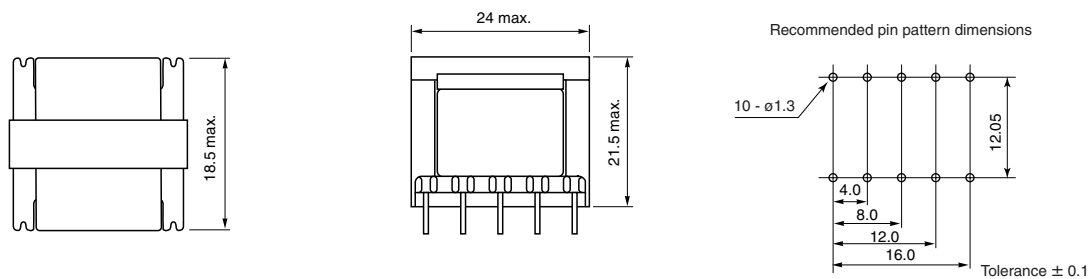
article name	Core Ae	W	Bobbin
FEI19	0.23	4.4	EB19-P1108-FA
FEE19	0.23	4.8	



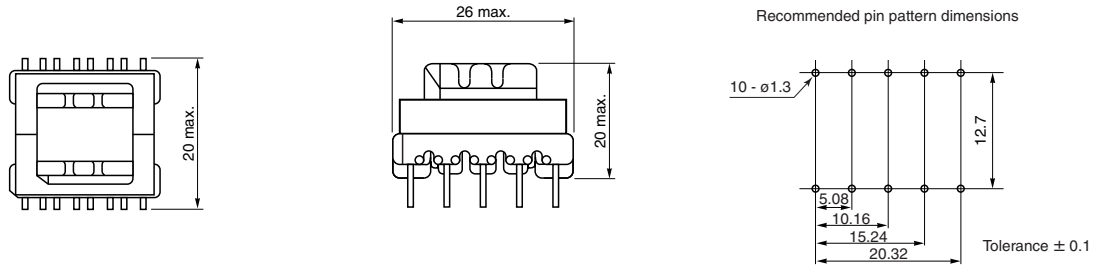
article name	Core Ae	W	Bobbin
FEI19W	0.224	7.0	EB19W-P1208-F



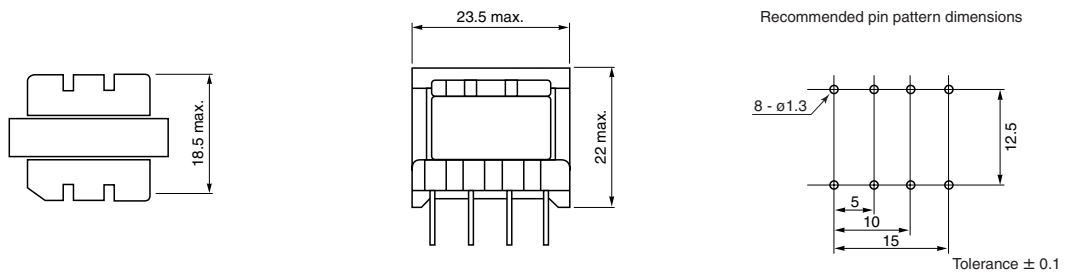
article name	Core Ae	W	Bobbin
FEI22	0.41	8.8	EB22-P1210-F
FEE22	0.42	8.8	



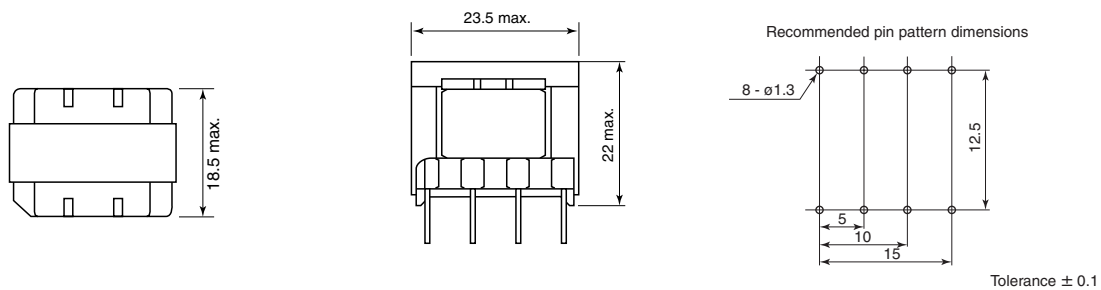
article name	Core Ae	W	Bobbin
FEI22	0.41	8.8	EB22-P1110-FA
FEE22	0.42	8.8	



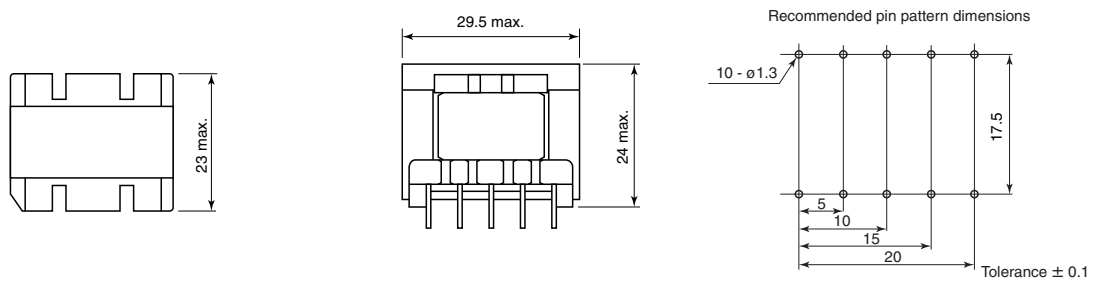
article name	Core Ae	W	Bobbin
FEI22S	0.36	7.7	EB22S-P1208-F



article name	Core Ae	W	Bobbin
FEI25	0.41	11.0	EB25-P1208-F

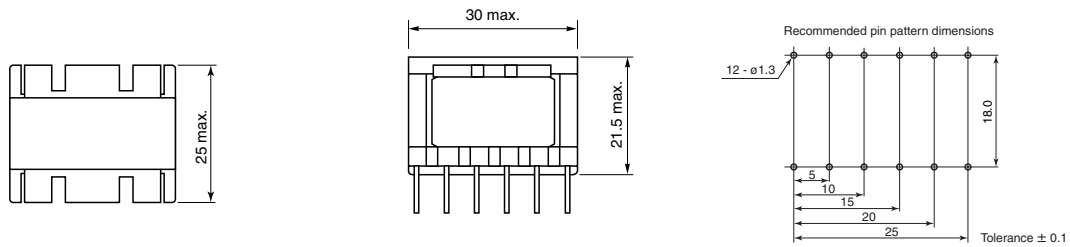


article name	Core Ae	W	Bobbin
FEI28	0.85	24.0	EB28-P1210-F
FEE28S	0.87	21.5	

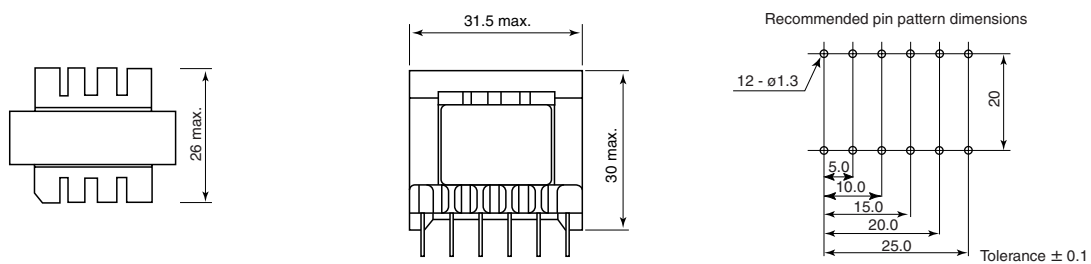




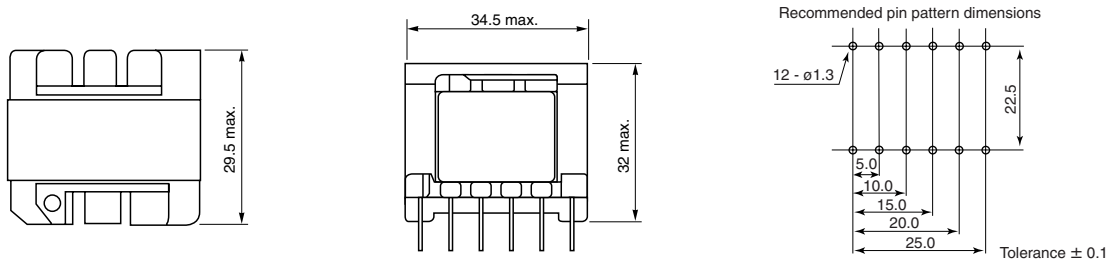
article name	Core Ae	W	Bobbin
FEI28	0.85	24.0	EB28-P1212-F
FEE28S	0.87	21.5	



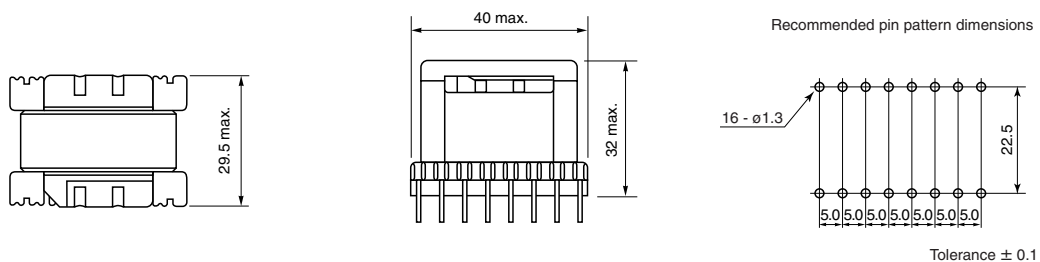
article name	Core Ae	W	Bobbin
FEI30	1.11	35.0	EB30-P1212-F
FEE30	1.11	33.0	



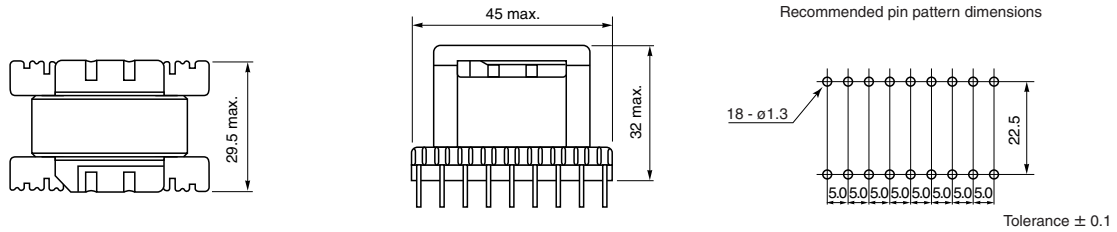
article name	Core Ae	W	Bobbin
FEI33	1.18	41.5	EB33-P1212-FS



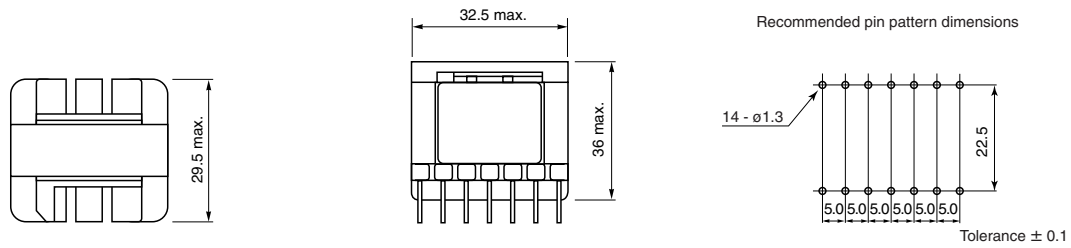
article name	Core Ae	W	Bobbin
FEI33	1.18	41.5	EB33-P1216-F



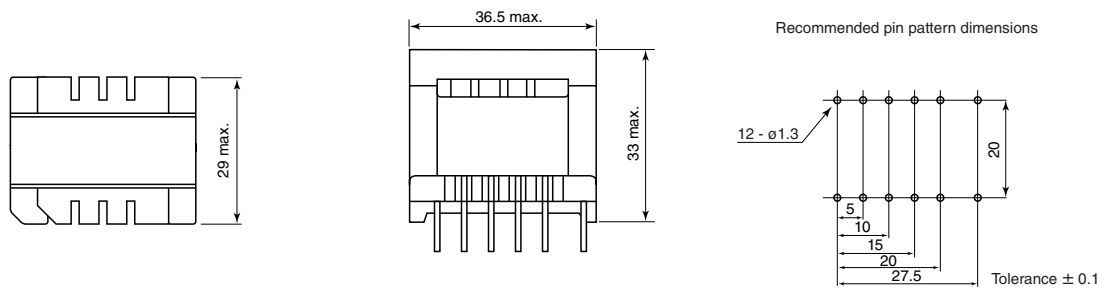
article name	Core Ae	W	Bobbin
FEI33	1.18	41.5	EB33-P1218-F



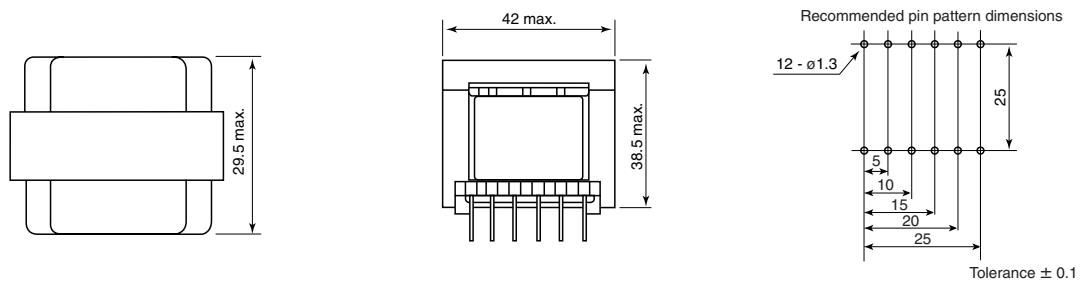
article name	Core Ae	W	Bobbin
FEI33	1.18	41.5	EB33-P1214-F1



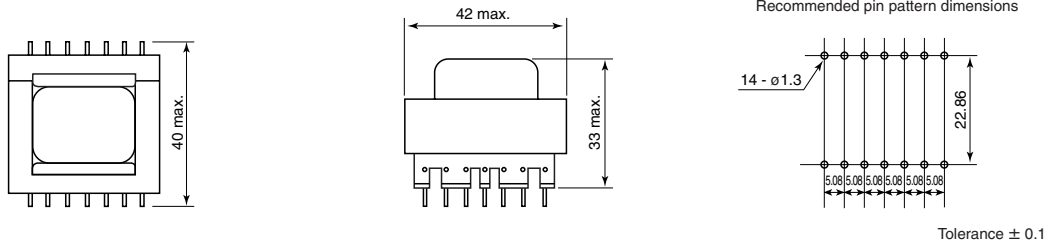
article name	Core Ae	W	Bobbin
FEI35S	1.2	41.5	EB35S-P1212-F



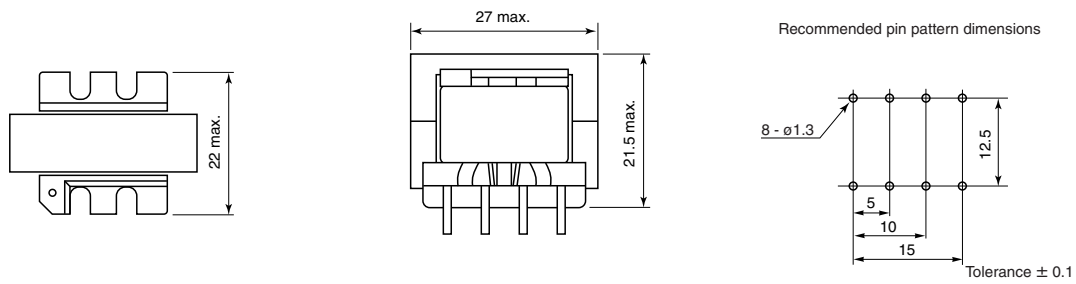
article name	Core Ae	W	Bobbin
FEI40	1.48	61.0	EB40-P1212-F
FEE40	1.28	51.1	



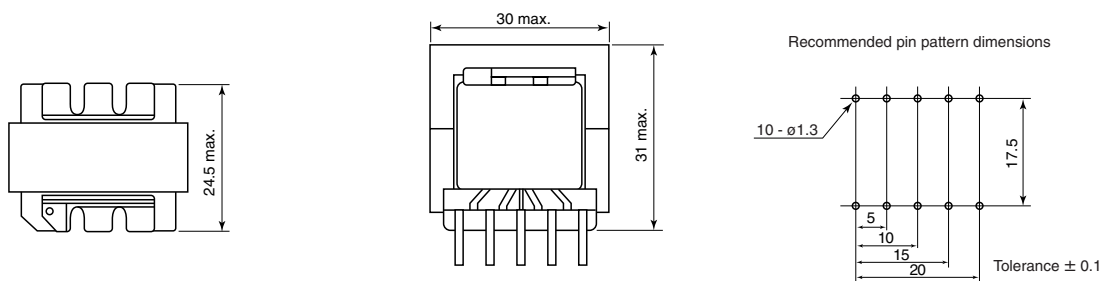
article name	Core Ae	W	Bobbin
FEI40	1.48	61.0	EB40-P1114-FA
FEE40	1.28	51.1	



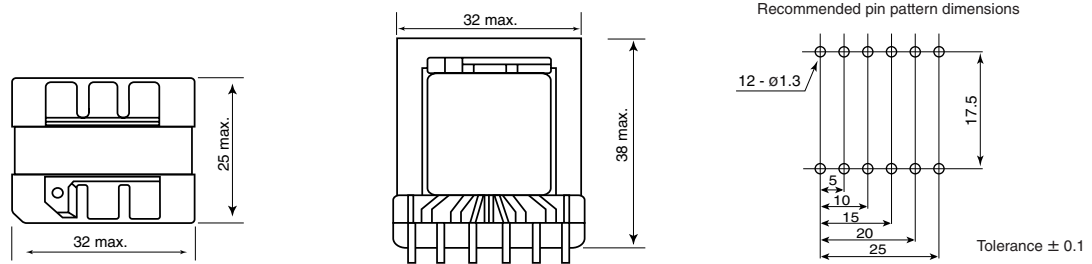
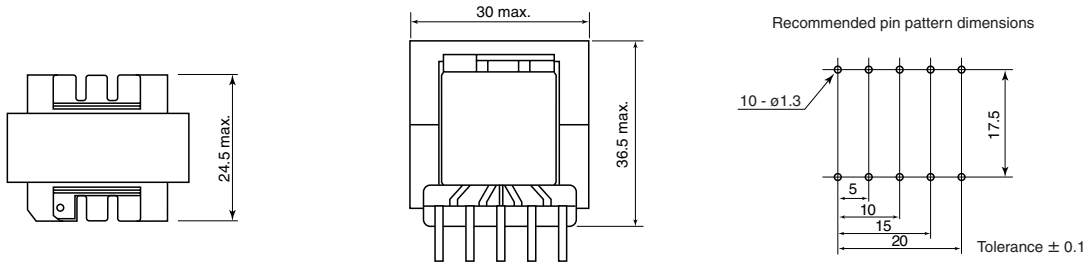
article name	Core Ae	W	Bobbin
FEER25.5	0.43	10.8	ERB25.5-P1208-F



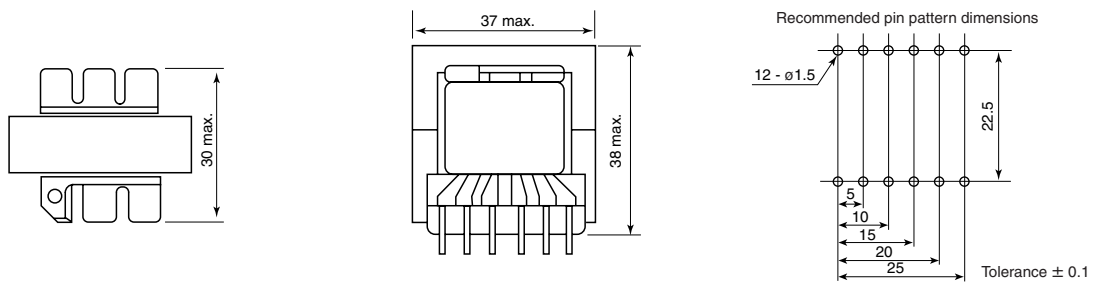
article name	Core Ae	W	Bobbin
FEER28	0.85	28.5	ERB28-P1210-F



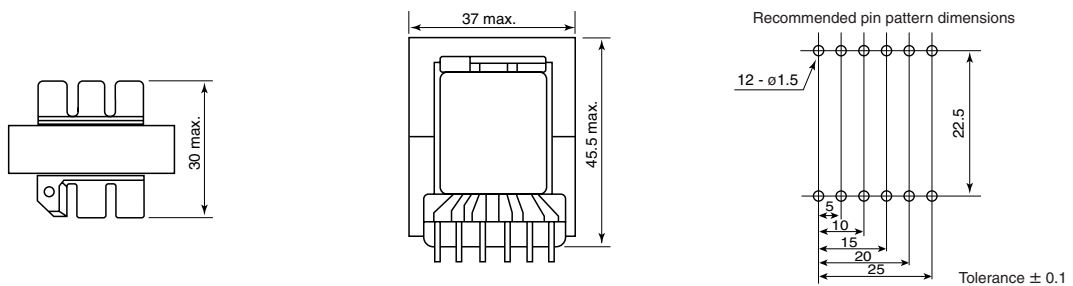
article name	Core Ae	W	Bobbin
FEER28L	0.85	33.2	ERB28L-P1210-F ERB28L-P1212-F



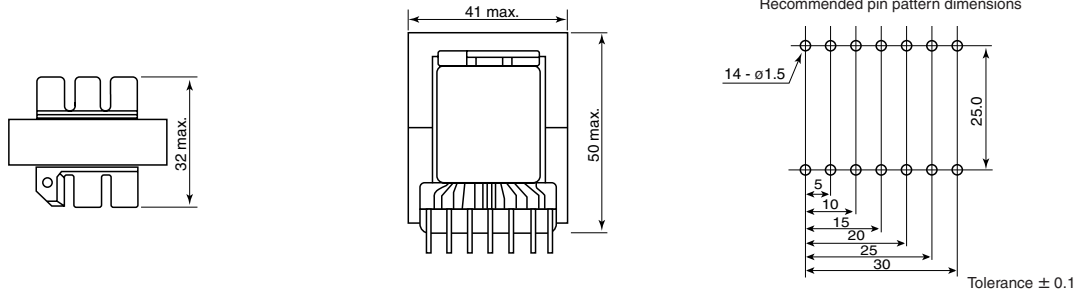
article name	Core Ae	W	Bobbin
FEER35	1.10	45.0	ERB35-P1212-F



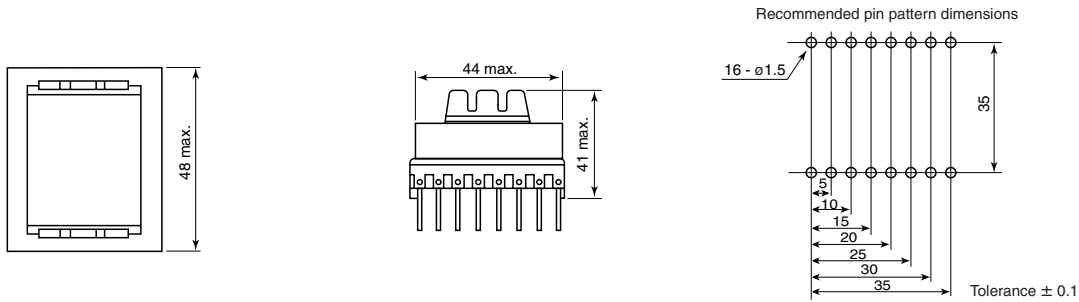
article name	Core Ae	W	Bobbin
FEER35L	1.08	50.7	ERB35L-P1212-F



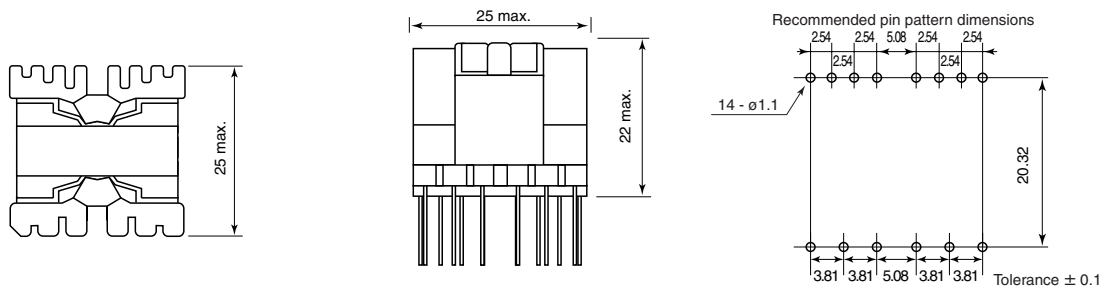
article name	Core Ae	W	Bobbin
FEER39L	1.32	70.0	ERB39L-P1214-F



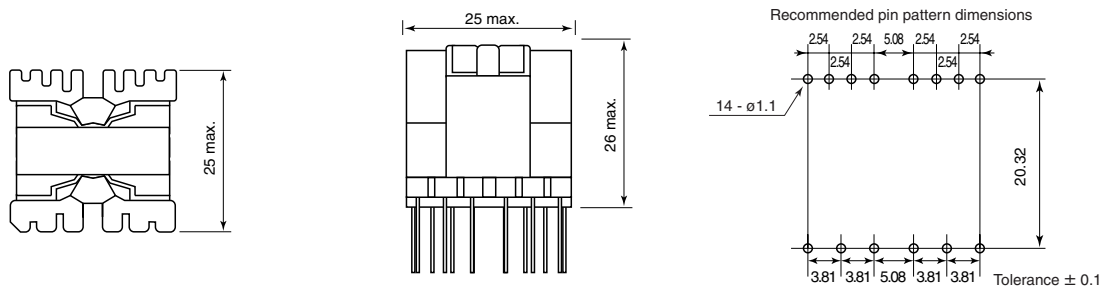
article name	Core Ae	W	Bobbin
FEER39L	1.32	70.0	ERB39L-P1116-F



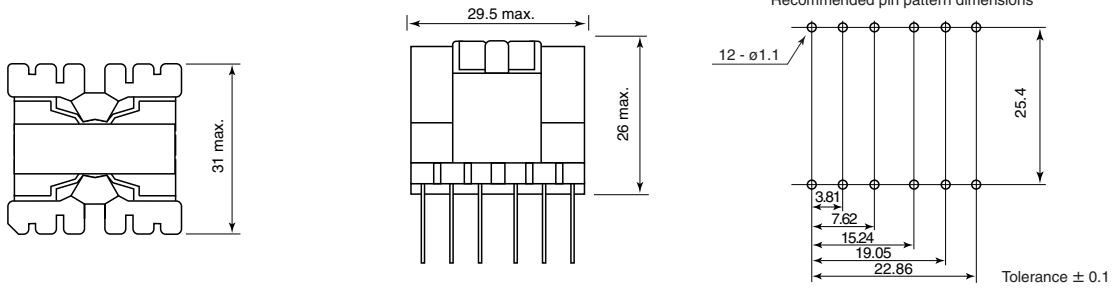
article name	Core Ae	W	Bobbin
FPQ2016-T-22	0.62	13.0	PQB2016-P1214



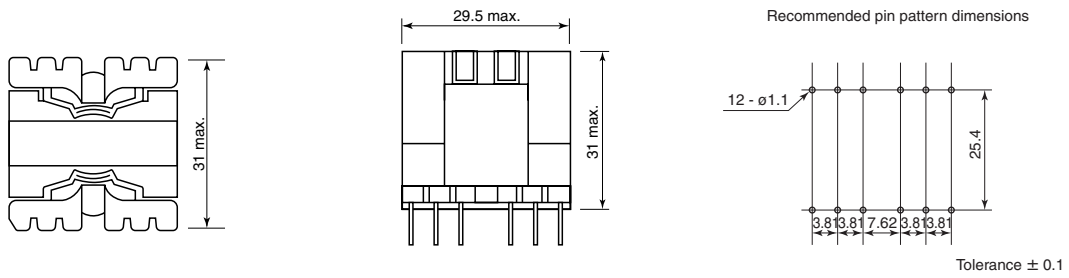
article name	Core Ae	W	Bobbin
FPQ2020-T-22	0.62	5.0	PQB2020-P1214



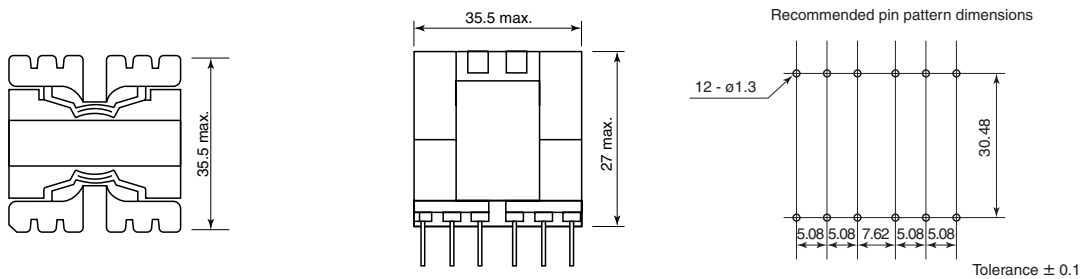
article name	Core Ae	W	Bobbin
FPQ2620-T-22	1.19	31.0	PQB2620-P1212



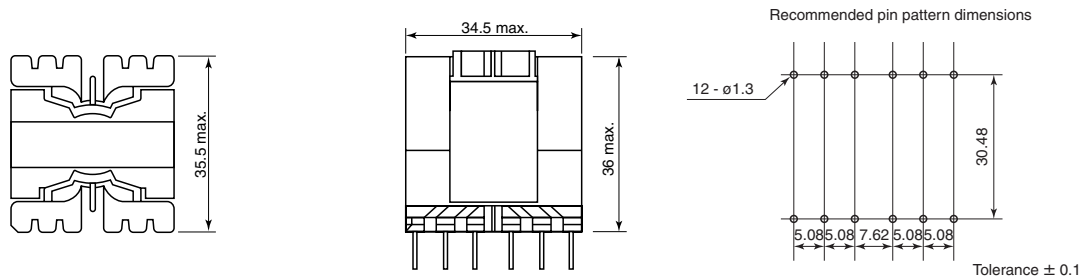
article name	Core Ae	W	Bobbin
FEQ2625-7-22	1.19	36.0	PQB2625-P1212-F



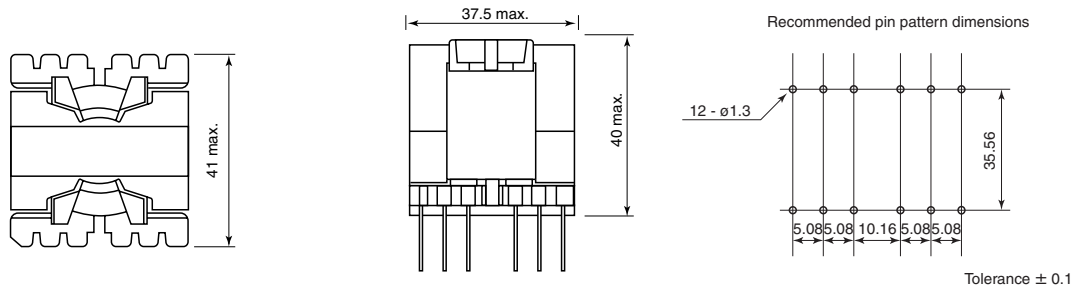
article name	Core Ae	W	Bobbin
FPQ3220-T-22	1.70	42.0	PQB3220-P1212



article name	Core Ae	W	Bobbin
PQB3220-T-22	1.61	55.0	PQB3230-P1212-F



article name	Core Ae	W	Bobbin
FPQ3535-T-22	1.96	73.0	PQB3535-P1212-FA



### Notes for Handling

- Confirm the required operating conditions for each product before using it.  
(Applying excess load may damage the transformer.)
- Avoid subjecting the terminals of the transformer to excessive stress.  
(This can damage the wiring.)
- Avoid handling the product while holding the transformer part after it is mounted on the board.  
(This can loosen the core or damage the wiring.)
- Never use any product that has been dropped .  
(A cracked core can cause unsatisfactory characteristics .)

## Custom transformer series (Through-hole type)

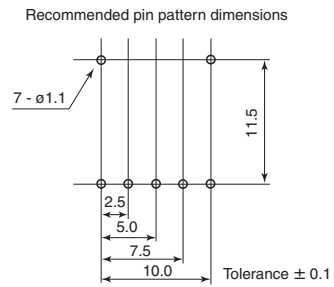
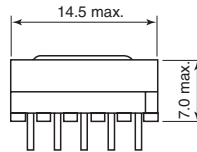
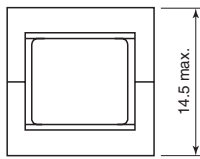
Output Wattage	Core Size	Bobbin
4W	FEY13.5 / 6.75 / 3.0	EYB13.5-P1107-F
5W	FEY16/14.5/5	EYB16-P1107-F1
6W	FEE18V	EB18V-P2109-F(Cover)
7W	FEE19V	EB19V-P1108-F
15W	FEI22S	EB22S-P1211-F
15W	FEI22S	EB22S-P1210-F,EB22S-C
15W	FEE23V	EB23V-P2111-F,EB23V-CK
20W	FEE23	EB23-P1209-FA
20W	FEY25/22	EYB25-P1111-F,EYB5-C
40W	FEEH28	ERB28S-P1216-FD
40W	FEEH28	EHB28-P1214-F2
40W	FEEH28	ERB28S-P5216-F
40W	FEE28V	EB28V-P2111-F,EB28V-CK
50W	FEY28D	EYB28-P1112-F
70W	FEY28DH	EYB28DH-P1112-F
150W	FEER40	ERB40-P1218-F

Circuit condition: Flyback at= 100 kHz

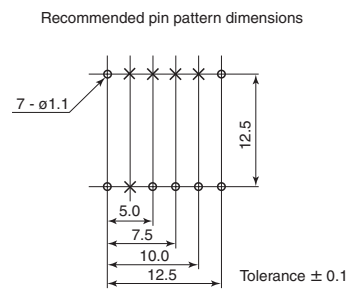
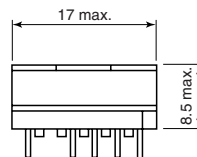
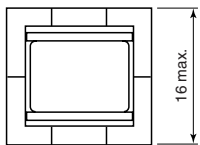
\*Contact us individually regarding safety standard complied models.



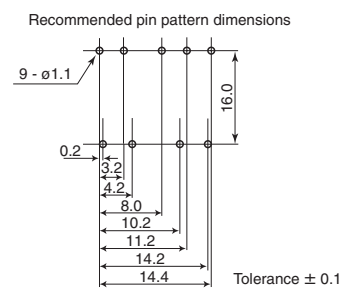
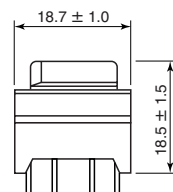
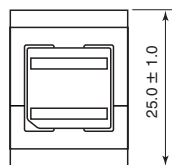
article name	Core Ae	W	Bobbin
FYB13.5/6.75/3.0	0.097	1.7	EYB13.5-P1107-F



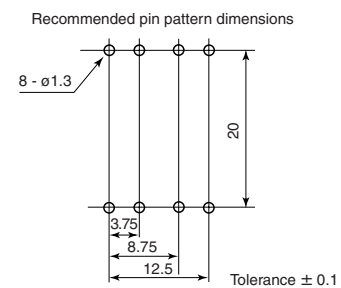
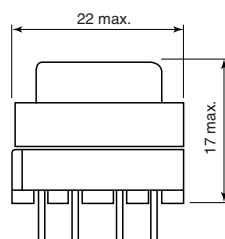
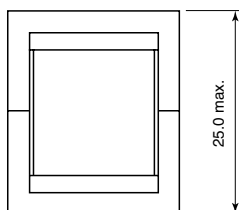
article name	Core Ae	W	Bobbin
FEY16/14.5/5	0.1593	3.0	EYB16-P1107-F1



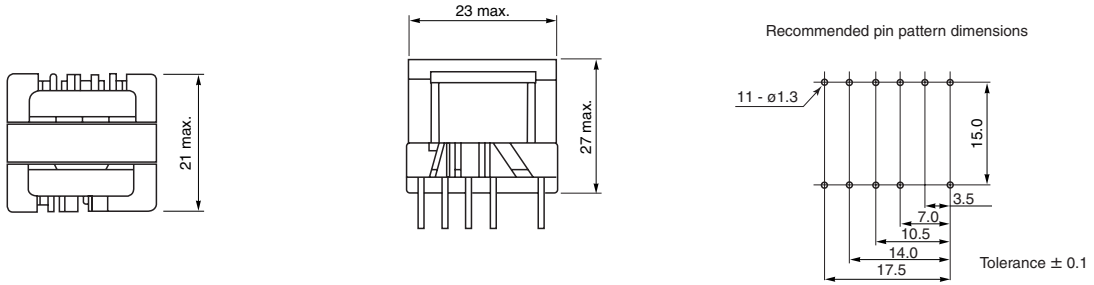
article name	Core Ae	W	Bobbin
FEE18V	0.193	5.16	EB18V-P2109-F,Cover



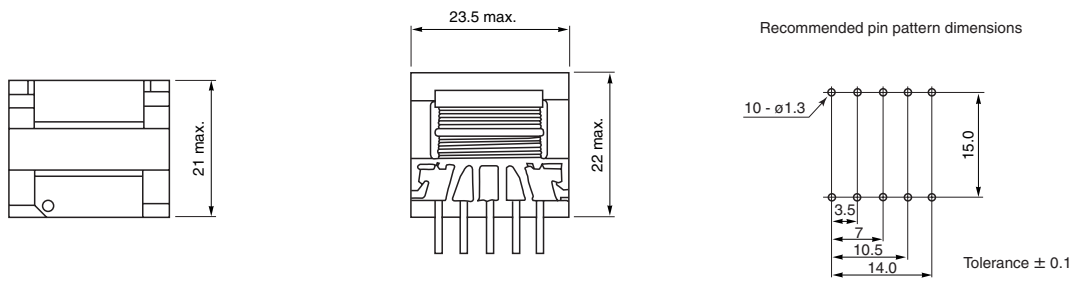
article name	Core Ae	W	Bobbin
FEE19V	0.23	6.7	EB19V-P1108-F



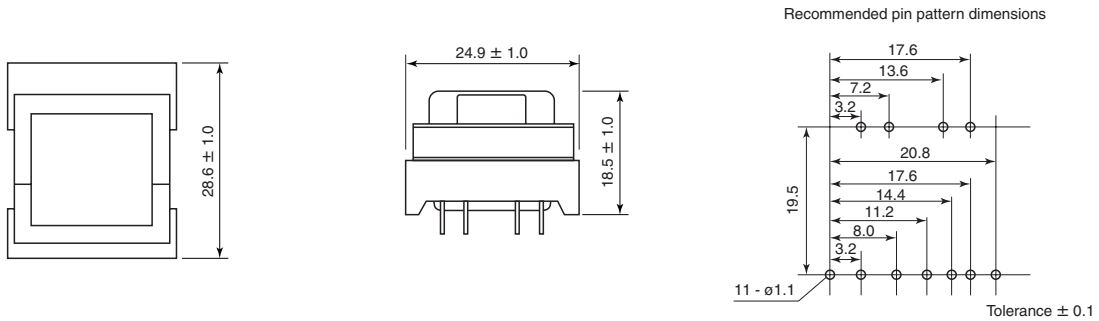
article name	Core Ae	W	Bobbin
FEI22S	0.36	7.7	EB22S-P1211-F



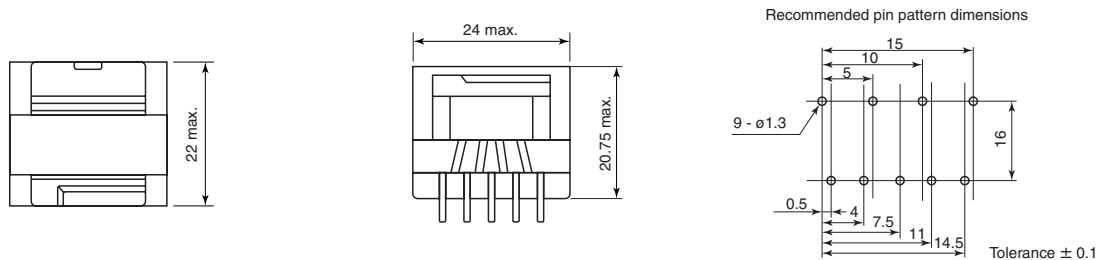
article name	Core Ae	W	Bobbin
FEI22S	0.36	7.7	EB22S-P1210-F, EB22S-C



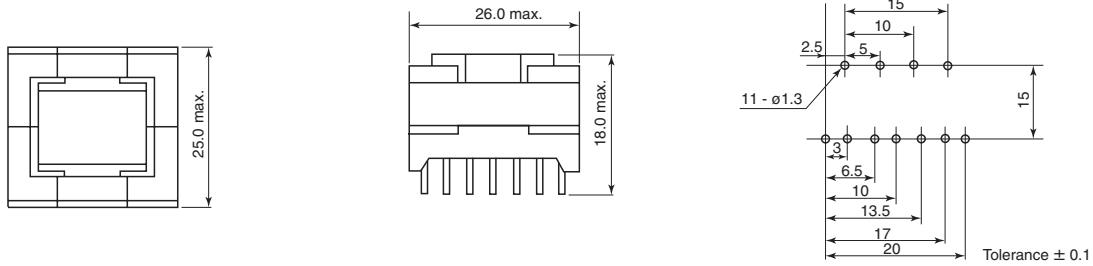
article name	Core Ae	W	Bobbin
FEE23V	0.225	7.3	EB23V-P2111-F, EB23V-CK



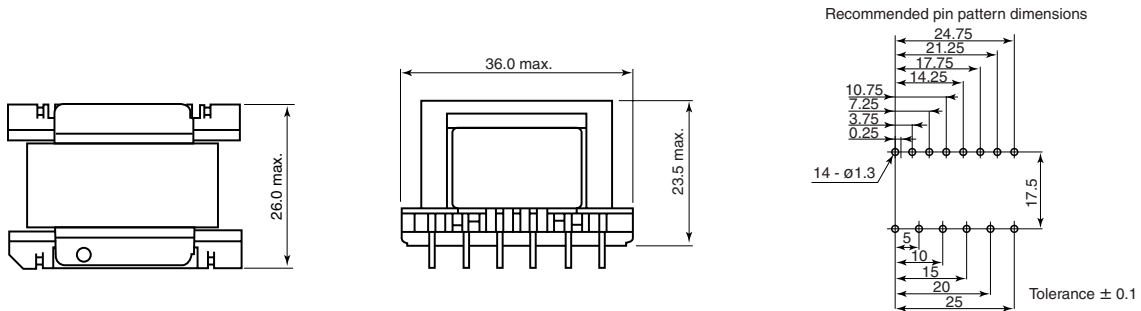
article name	Core Ae	W	Bobbin
FEE23	0.42	10.9	EB23-P1209-EA



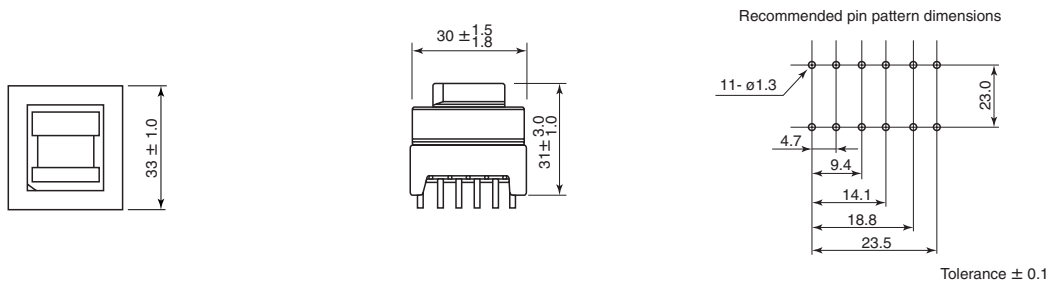
article name	Core Ae	W	Bobbin
FEY25/22	0.55	16.9	EYB25-P1111-F, EYB25-C



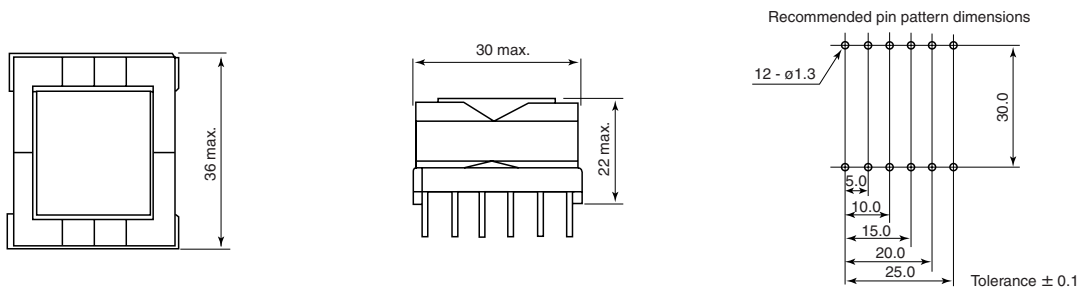
article name	Core Ae	W	Bobbin
FEEH28	0.844	23.8	EHB28-P1214-F2



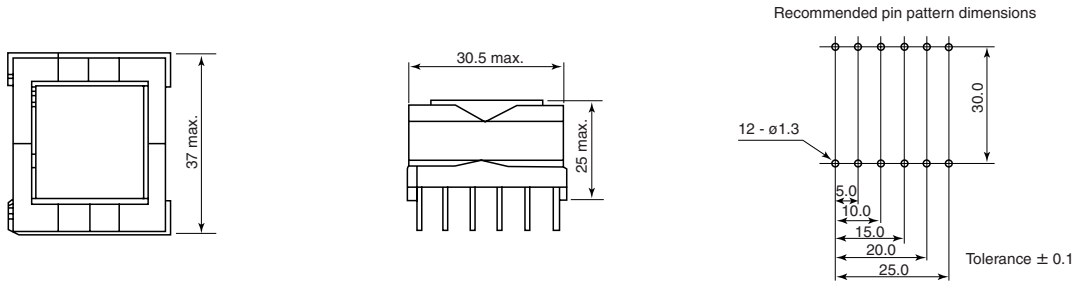
article name	Core Ae	W	Bobbin
FEE28V	0.8784	32.6	EB28V-P2111-F, EB28V-CK



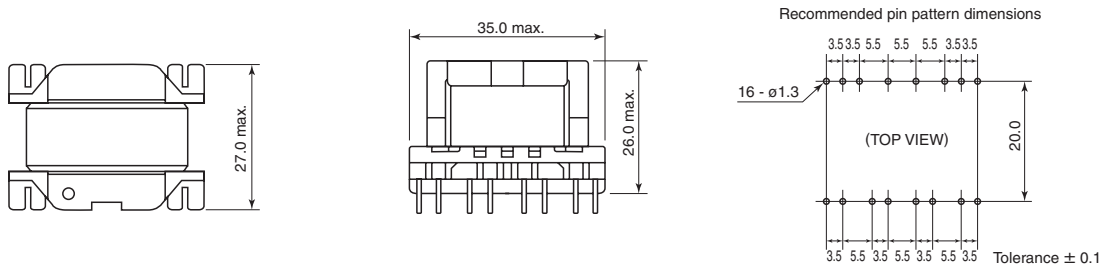
article name	Core Ae	W	Bobbin
FEY28D	0.75	32.0	EYB28-P1112-F



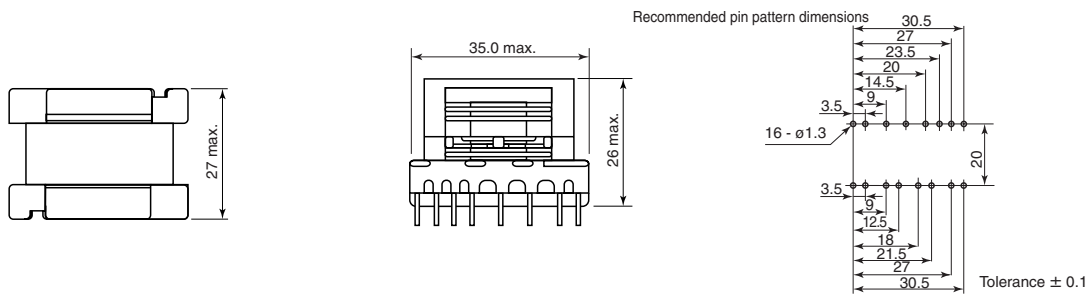
article name	Core Ae	W	Bobbin
FEY28DH	1.15	44.8	EYB28DH-P1112-F



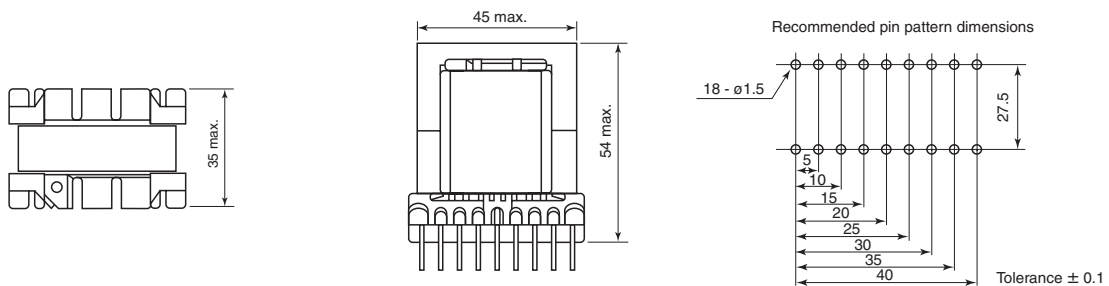
article name	Core Ae	W	Bobbin
FEEH28	0.844	23.8	EBR28S-P1216-FD



article name	Core Ae	W	Bobbin
FEEH28	0.844	23.8	ERB28S-P5216-F



article name	Core Ae	W	Bobbin
FEER40	1.525	87.5	ERB40-P1218-F



**Notes for Handling**

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- Confirm the required operating conditions for each product before using it.  
(Applying excess load may damage the transformer.)
- Avoid subjecting the terminals of the transformer to excessive stress.  
(This can damage the wiring.)
- Avoid handling the product while holding the transformer part after it is mounted on the board.  
(This can loosen the core or damage the wiring.)
- Never use any product that has been dropped .  
(A cracked core can cause unsatisfactory characteristics.)

# Transformers for Miniature Cold-cathode Fluorescent Lamps



## Outline

The widespread use of liquid-crystal displays in car dashboards, audio equipment, and notebook computers has resulted in greater demand for transformers for back-lighting purposes. With our integrated manufacturing capability that enables us to produce not only devices but also materials used in them, we have produced a series of high-quality transformers that features small-size, low-height construction. (Specification depends upon customize request)

## Features

- Usable in high-voltage applications
- Small-size, low profile
- Through-hole type ideal for board mounted applications (16W, R22SW)

## Applications

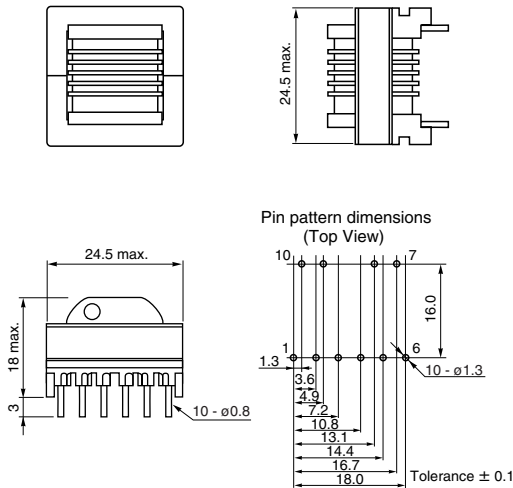
- LCD back light for personal computers
- Switching for cold-cathode fluorescent lamps of facsimile machines

Model	Maximum Power* (W)	Insulation AC Withstand Voltage [primary to secondary for 1minute] (VAC)	Maximum Output Voltage (V)
<b>BLT-16W</b>	5	1500	1200
<b>BLT-R22SW</b>	12	2000	1500

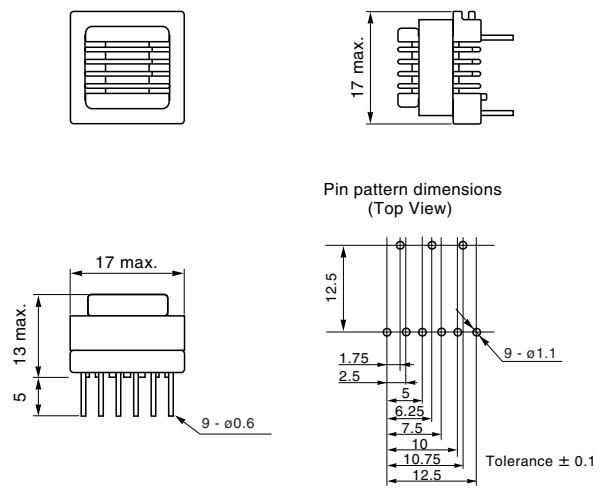
\* Maximum power: (f = 30 kHz)

## Shape and Dimensions

### ● BLT-R22SW

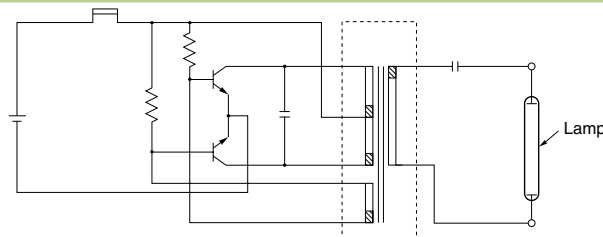


### ● BLT-16W



[mm]

## Circuit Example



# SMD Type Common Mode Choke Coils for DC-DC Converters



## Features

- Accomodates automatic mounting
- Compact & thin profile
- Wide band
- High inductance

## Application

- On-board power supply

## Operating Temperature Range

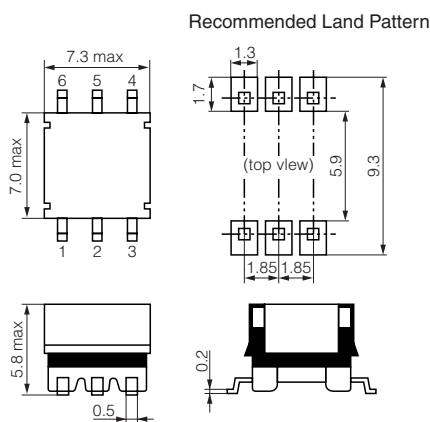
-20°C~110°C (including temperature of parts itself)

Model	Rated Voltage (V)	Rated Current (A)	Inductance ( $\mu$ H) min.	DC Resistance (m $\Omega$ ) max.	Withstanding Voltage 1MIN.,Coil to Coil DC (V)
DCM5-01	50	0.2	2540	3180	250
DCM5-02	50	0.5	520	510	250
DCM5-03	50	1.0	210	180	250
DCM5-04	50	2.0	30	30	250
DCM11-01	50	0.8	1350	240	250
DCM11-02	50	1.3	630	105	250

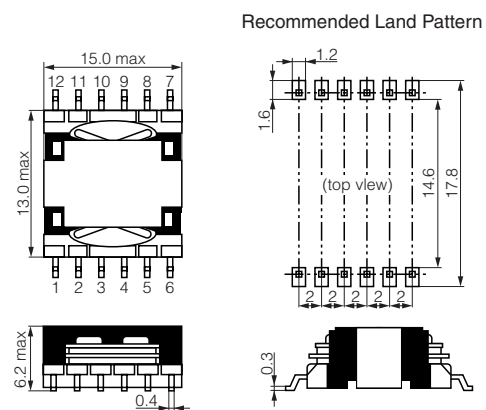
· Inductance measurement condition : f=10kHz, 0.1V · Insulation resistance : over 100m $\Omega$  (DC250V, 1MIN between each line)  
 Note:Applicable to other uses besides examples above by adjusting number of turns. contact us for detail.

## Shapes and Dimensions

### ● DCM5 Series



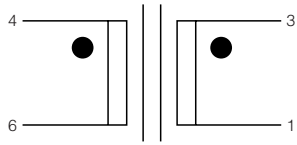
### ● DCM11 Series



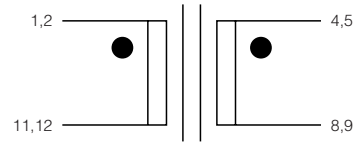
[mm]

Typical Circuit

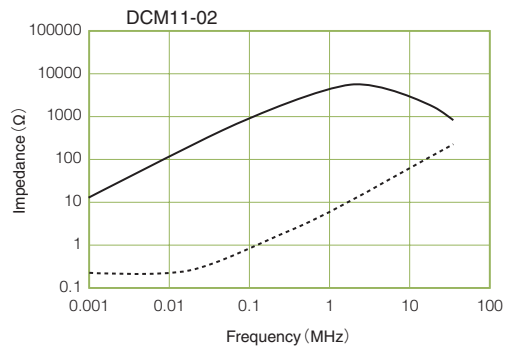
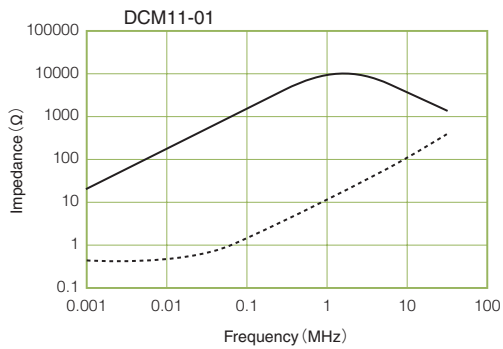
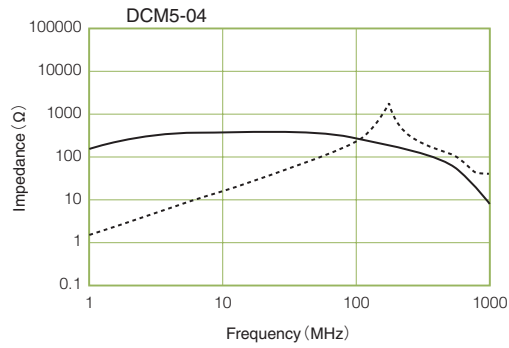
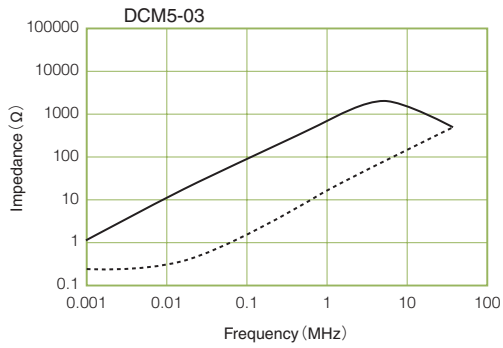
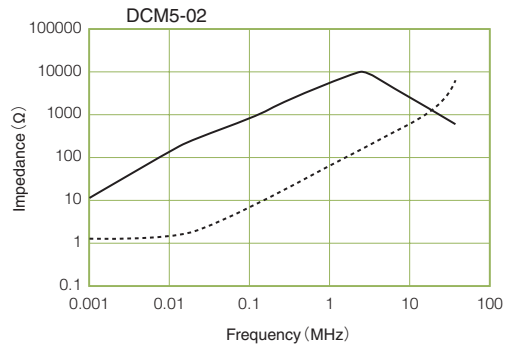
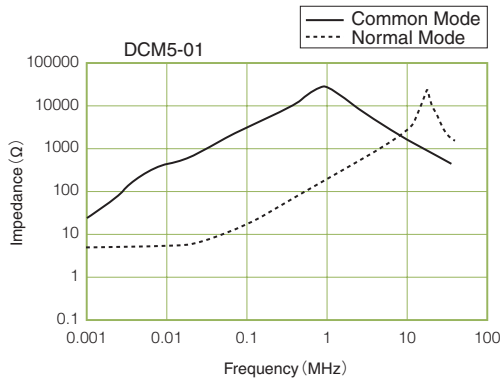
● DCM5 Series



● DCM11 Series

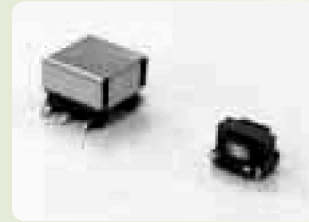


Impedance Characteristics





# SMD Type Current Transformers



## Features

- Accomodates automatic mounting
- Compact & thin profile

## Operating Temperature Range

-20°C~110°C (including temperature of parts itself)

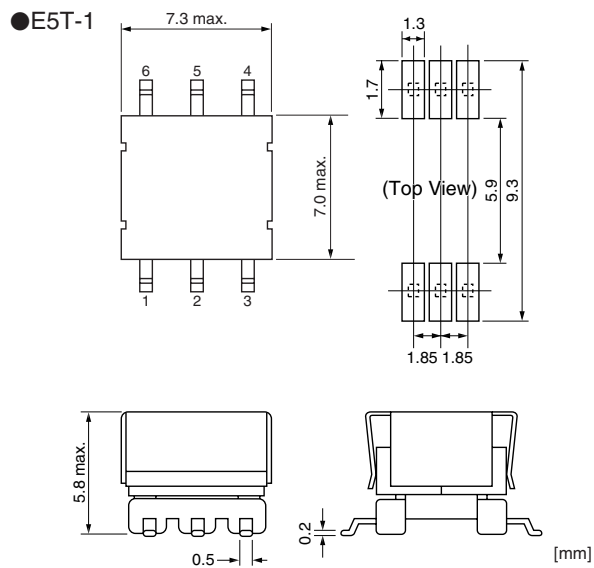
Model	Rated Voltage Primary (V)	Winding Ratio P : S	Inductance Secondary min. (mH)	DC Resistance Primary (mΩ) max.	DC Resistance Secondary (mΩ) max.
<b>E5T-1</b>	3	1 : 100	2.0	13.5	6.55
<b>E5T-2</b>	1.5	4 : 20	0.08	35	310

· Inductance measurement condition : f=10kHz, 1mA

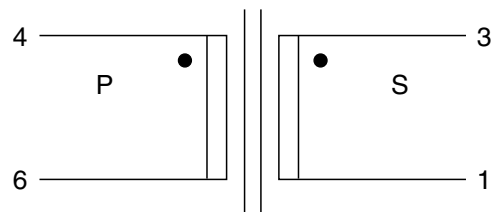
Note:Applicable to other uses besides examples above by adjusting number of turns. contact us for detail.

## Shapes and Dimensions

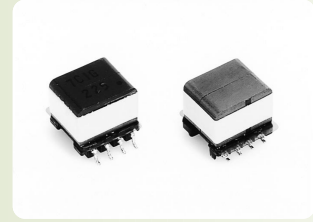
Recommended pin pattern dimensions



## Typical Circuit



# ADSL Line Transformers DSL7 Series



### Features

- Accomodates automatic mounting
- Space saving compact design
- Complies with various safety standards

### Applications

- Digital Subscriber Line Access Multiplexer (ADSL modem at the exchange)

### Operating Temperature Range

-20°C~110°C (including temperature of parts itself)

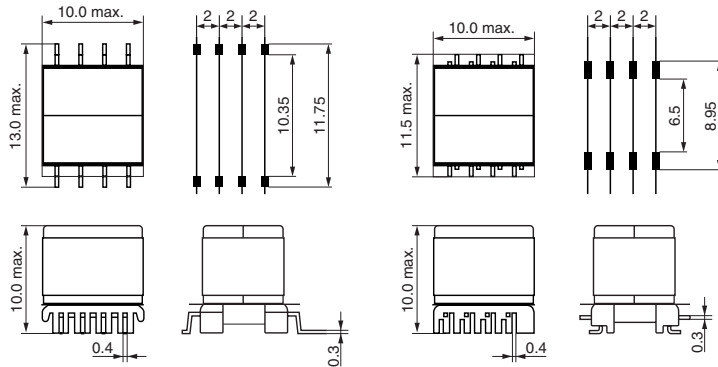
Model	Winding Ratio 1-4 : 8-5 (tie 2-3, 6-7)	Inductance [ $\mu$ H] 1 : 4 (tie 2-3) 10kHz, 0.1V	Leakage Inductance [ $\mu$ H] 1 : 4 (tie 2-3, 8-6, 7-5) 100kHz, 0.1V	DC Resistance [ $\Omega$ ]				Line Capacitance Primary-Secondary (tie 2-3, 6-7) 10kHz, 0.1V	Withstanding Voltage Primary-Secondary (tie 2-3, 6-7)	Total Harmonic Distortion 1-4 : 8-5 50kHz, 4V
				1-3	2-4	8-6	7-5			
<b>DSL7-C1</b>	1:1	850 $\pm$ 10%	5max.	1.05max.	1.05max.	0.85max.	1.05max.	—	1875V, 1s(0.5mA)	
<b>DSL7-G1</b>	1.15:1	475 $\pm$ 5%	10max.	0.85max.	0.85max.	1.05max.	1.05max.	50max.	1875V, 1s(0.5mA)	-80max.

Note:Applicable to other used besides examples above by adjusting number of turns. Contact us for detail.

### Shapes and Dimensions

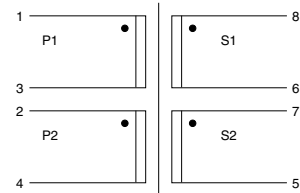
●type1

●type2

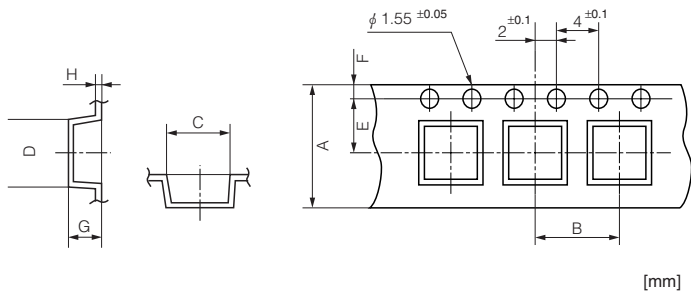


[mm]

### Circuit Diagram



Tape Dimensions / Packing Quantity



Model	Dimension (mm)			Typical Tolerance $\pm 0.1$					Packing Quantity [pcs./reel]
	A	B	C	D	E	F	G	H	
<b>DCM5 Series</b>	16 $\pm$ 0.3	8	6.3	8.3	7.5	1.75	5.2	0.38 $\pm$ 0.05	2000
<b>E5T Series</b>	16 $\pm$ 0.3	8	6.3	8.3	7.5	1.75	5.2	0.38 $\pm$ 0.05	2000
<b>DCM11 Series</b>	24 $\pm$ 0.3	20	14.5	17.2	11.5	1.75	6.8	0.38 $\pm$ 0.05	500
<b>DSL7(type2)※</b>	24 $\pm$ 0.3	20	10.9	10.2	11.5	1.75	9.2	0.5 $\pm$ 0.05	500

Minimum order quantity=quantity in per reel.  
 ※Packing material for DSL7(type1) is to be decided. Embossed carrier tape is planned.