

### **GENERAL DESCRIPTION**

CM6901 is a SLS, SRC/LLC + SR resonant controller and it can operate at both SRC and LLC region with synchronous rectification to achieve high efficiency. Its unique features: FM + 2 PWMING modes. It is used for DC-DC conversion in offline application. Light load regulation is accomplished by transitioning the controller from frequency modulation mode into PWM mode. To have the optimal balance performance between hold-up time and efficiency, its frequency range crosses fr1, resonant frequency.

CM6901 system has a constant voltage feedback loop with precision 2.5V Vfb reference. The 1V ILIMIT can be latched when a standby converter is applied; otherwise, it is an auto-restart I limit.

### FEATURES

- 90+ Controller
- Smaller and Thinner Transformer
- 20mS with reasonable Bulk Cap
- SLS, SRC/LLC + SR resonant controller
- Supply Voltage Range : 10V to 15V.
- 2 Gate Drivers: Typical Peak Drive from 12V supply : (PMOS~200 ohm and NMOS~100 ohm).
- UVLO=11V with 1V Hystetesis.
- Reference OK Comparator.
- FM + 2 PWMING Mode Operation
  - Light Load PWMING

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- SR Ideal Diode PWMING
- GM feao, FM modulation Error Amplifier
- GM deao, PWM Error Amplifier
- Close Loop Soft start Capability with Shutdown Function.
- Auto-Restart during Current Limit .
- Precision 2.5V Vfb threshold for constant voltage feedback loop.
- Precision 1V I limit threshold.

### **CM6901** Pin Configuration

- APPLICATIONS
- DC-DC power supply.
- Operates in FM/PWM mode.





### ORDERING INFORMATION

Part Number	Temperature Range	Package
CM6901GIS*	-40℃ to 125℃	16-Pin SOP (S16)
CM6901GISTR*	-40°℃ to 125°℃	16-Pin SOP (S16)
CM6901GIP*	-40℃ to 125℃	16-Pin PDIP (P16)
CM6901XIS*	-40°℃ to 125°℃	16-Pin SOP (S16)
CM6901XISTR*	-40℃ to 125℃	16-Pin SOP (S16)
CM6901XIP*	-40℃ to 125℃	16-Pin PDIP (P16)

\*Note : G : Suffix for Pb Free Product

X : Suffix for Halogen Free Product

TR : Package is Typing Reel

## **BLOCK DIAGRAM**





### **PIN DESCRIPTION**

Din No	Cumb al	Description		Operating Voltage			
PIN NO.	Symbol	Description	Min.	Тур.	Max.	Unit	
1	RSET	External resistor which convert FEAO voltage signal into current signal for frequency modulation. SR is PWMing Control , when RSET lower than 1.5V.			5.5	V	
2	VFB	Non-inverting input into resonant error amplifier .	0	2.5	3	V	
3	FEAO	Resonant error amplifier output and compensation node for frequency modulation control.	0		5.5	V	
4	D_IN-	Inverting input into PWM error amplifier.	0		6	V	
5	D_IN+	Non-inverting input into PWM error amplifier.	0		6	V	
6	DEAO	PWM error amplifier output and compensation node for PWM control.	0		4.5	V	
7	CSS	Soft start for FM/PWM operation with 1V enable threshold. Also, use for auto-restart operation during current limit.	0		5.5	V	
8	ILIM	Input to current comparator with 1V threshold.	0	1	1.25	V	
9	RTCT	Oscillator timing components which set the minimum frequency.	1.2		3	V	
10	GND	Ground					
11	SDRVB	Synchronous MOSFET driver output.	-0.3		VCC	V	
12	SDRV	Synchronous MOSFET driver output.	-0.3		VCC	V	
13	PRIDRVB	Primary side MOSFET driver output.	-0.3		VCC	V	
14	PRIDRV	Primary side MOSFET driver output.	-0.3		VCC	V	
15	VCC	Positive supply for the IC	10	15	17.5	V	
16	VREF	Buffered output for the 7.5V voltage reference		7.5		V	



#### **ABSOULTE MAXIMUM RATINGS** (TA=25°C, unless otherwise specified.)

The following ratings designate persistent limits beyond which damage to the device may occur.

Symbol	Parameter	Value	Unit
vcc	DC Supply Voltage	-0.3 to 20	V
SRDRV	SRDRV Voltage	-0.3 to VCC+0.3	V
SRDRVB	SRDRVB Voltage	-0.3 to VCC+0.3	V
PRIDRV	PRIDRV Voltage	-0.3 to VCC+0.3	V
PRIDRVB	PRIDRVB Voltage	-0.3 to VCC+0.3	V
VREF	VREF Voltage	-0.3 to 7.5	V
RTCT	RTCT Voltage	-0.3 to VREF+0.3	V
VILIM	VILIM Voltage	-0.3 to VREF+0.3	V
CSS	CSS Voltage	-0.3 to VREF+0.3	V
DEAO	PWM Error Amplifier Output Voltage	-0.3 to VREF+0.3	V
D_IN+	Non-Inverting Input Into PWM Error Amplifier Voltage	-0.3 to VREF+0.3	V
FEAO	Resonant Error Amplifier Output Voltage	-0.3 to VREF+0.3	V
VFB	Non-Inverting Input Into Resonant Error Amplifier Voltage	-0.3 to VREF+0.3	V
RSET	V to I Voltage	-0.3 to VREF+0.3	V



## **ELECTRICAL CHARACTERISTICS**

(VCC=12V, RT=43K±1%, CT=820pF±1%, Freq. = 80 KHz, Duty Cycle=48% , Temp= -40  $\sim$  125  $\,^\circ\!C$  , unless otherwise specified.)

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit
VREF (pin 16)						
Reference Voltage	VREF	TA=25℃; Iref=1mA	7.41	7.5	7.59	V
Line Regulation		11.5V < Vcc < 16.5V		10	25	mV
Load Regulation		0mA < Iref < 5mA	0	10	25	mV
<b>VCC</b> (pin 15)						
VCC Start up voltage	Vstart		10.23	11	11.5	V
VCC Turn off voltage			9.3	10	10.7	V
VCC Start up current			63.5	74.5	85.5	uA
UVLO Hysteresics	Hyst		0.93	1	1.07	V
Operating Current	ICC			1		mA
SRDRVB , SRDRV , PRIDRV	B, PRIDRV	(pin 11,12,13,14)				
Output Low Voltage	Vol	Isrdrv=-6mA		0.6	1.2	V
Output High Voltage	Voh	Isrdrv=+6mA	9.6	10.8		V
lout peak source current	lout			0.12		А
lout peak sink current	lout			0.12		А
PMOS Rdson	Rout	<b>TA=25</b> ℃		200	250	Ohm
NMOS Rdson	Rout	<b>TA=25</b> ℃		100	150	Ohm
Dead Time between PRIDRV and PRIDRVB	TDead	TA=25℃		650		ns
Rising Edge Delay Between PRIDRV and SRDRV	Tdelay	TA=25℃	175	217.5	260	ns
Falling Edge Delay Between SRDRV and PRIDRV	Tdelay	TA=25℃	178	222.5	267	ns
Duty Cycle Range			0		50	%
ILIM (pin 8)		1	1	1	1	
Threshold Voltage	Vth	<b>TA=25</b> ℃	1.45	1.65	1.75	V
SS Voltage Hi		<b>TA=25</b> ℃	0.95	1	1.05	V
SS Voltage Low		<b>TA=25</b> ℃	1.31	1.64	1.97	V



### **ELECTRICAL CHARACTERISTICS**

(Vcc=12V, RT=43K±1%, CT=820pF±1%, Freq. = 80 KHz, Duty Cycle=48% , Temp=-40 ~ 125 $^{\circ}$ C , unless otherwise specified.)

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit
CSS (pin 7)						
Soft Start Current	lss	<b>TA=25</b> ℃	-8.7	-7.5	-6.3	uA
Enable Voltage	Ven	<b>TA=25</b> ℃	0.9	1.0	1.1	V
Enable Hysteresis	Vhyst	<b>TA=25</b> ℃		100		mV
Auto-restart Upper Threshold	Vup		4.85	5.0	5.15	V
Auto-restart Lower Threshold	Vlw			1-Vhyst		V
FEAO Resonant Error Amplif	i <b>ier</b> (pin 3)					
Input Voltage Range					6	V
Transconductance	Gm	VFB±50mV, TA=25℃	37	46.5	56	umho
Feedback Reference Voltage		<b>TA=25</b> ℃	2.475	2.5	2.525	V
Input Bias Current			-1	-0.5		uA
Output High Voltage				5.5+VBE		V
Output Low Voltage				0.1	0.4	V
Sink Current		VFB =250mV , TA=25℃	-20.4	-17	-13.6	uA
Source Current		VFB = -250mV , TA=25℃	12.4	15.6	18.8	uA
Open Loop Gain				60		dB
DEAO PWM Error Amplifier (	pin 6)					
Input Voltage Range					6	V
Reference Voltage			0.96	1.01	1.06	V
Transconductance	Gm	<b>TA=25</b> ℃	95	120	145	umho
Input Bias Current			-1.0	-0.5		uA
Output Voltage High				3+VBE		V
Output Voltage Low				0.1	0.4	V
Sink Current				-5		uA
Source Current				7		uA
Open Loop Gain				60		dB



### **ELECTRICAL CHARACTERISTICS**

(Vcc=12V, RT=43K±1%, CT=820pF±1%, Freq. = 80 KHz, Duty Cycle=48% , Temp=-40 ~ 125  $^\circ\!C$  , unless otherwise specified.)

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Unit
RTCT Oscillator;RT=43K±1%	;CT=820pF±	<b>:1%,</b> (pin 9)				
Initial Accuracy		<b>TA=25</b> ℃	72	80	88	Khz
Voltage Stability		11.5V < Vcc < 16.5V		1		%
Temperature Stability				2		%
Ramp Valley to Peak Voltage				2		V
CT Discharge Time				350		nS
Maximum Duty cycle		DEAO > 3.0V		48		%
Minimum Duty Cycle		DEAO < 1.2V		0		%
SR Ideal Diode (pin12)						
Maximum Duty cycle		<b>TA=25</b> ℃	35	40	45	%
Middle Duty cycle		<b>TA=25</b> ℃	22	24.5	27	%
Minimum Duty cycle		<b>TA=25</b> ℃	2.5	4.65	6.8	%
Frequency at Minimum Duty cycle		TA=25℃	36	39	42	KHz



### **Timing Diagram**



### Oscillator

The oscillator frequency is determined by the values of RT and CT.

fosc = 1 / (tRAMP + t DEADTIME)

tramp = RT \* CT \* In((VREF + ICHG\*RT -1.25)/(VREF + ICHG\*RT -3)) where ICHG = 4\*(FEAO-VBE)/RSET

tdeadtime = 2.5V/2.5mA \* CT = 850 \* CT

# Resonant Section

Frequency modulation of the resonant controller section is accomplished by controlling the charging current of the oscillator through resonant error amplifier. The frequency of the resonant section is  $\frac{1}{2}$  of the oscillator frequency. Compensation is accomplished by connecting R and C in series to the FEAO pin.

ICHG = 4\*(FEAO-VBE)/RSET

ICHG(max) = 20V/RSET





### 2 PWMings:

## SR Ideal Diode PWMing (Synchronous Outputs)

SR Ideal Diode PWMing for synchronous drivers is accomplished by comparing the voltage signal at the RSET pin to RTCT ramp. The pulse-width reduction happens when the voltage at the RSET is lower than 1.5V. This allows safe operation of the power converter with synchronous rectification when the switching frequency is below the highest resonant point frequency fr1.





### Light Load PWMING

Light Load PWMING control is used in voltage mode. In Applicaton1, when FEAO is equal to the external reference voltage at D\_IN+ due to light load, the frequency of the oscillator is fixed at a value determined by the voltage at FEAO, and the controller enters into PWM mode. In this mode, both frequency modulator and pulse width modulator are active.

In Application 2 ,the offset voltage Vos is externally generated between the  $V_{FB}$  pin and D\_IN- through a resistor network  $R_{f1}$ ,  $R_{f2}$  and  $R_{f3}$  in order to separate the interaction between FM mode and PWM mode.

 $Vos = 2.5VXR_{f2}/R_{f3}$ 

When the output load is reduced to a point where FEAO is railed to its maximum value of 5V+ VBE, the frequency is fixed at a value determined by 5V+VBE. Further reduction in load will cause the controller to go into PWM mode when the voltage at the VFB pin is 2.5V+Vos. During this light load operation, FM error amplifier is in open-loop mode while PWM is in closed-loop mode. Compensation is accomplished by connecting R and C in series to the DEAO pin.

In Application 3 , low gain configuration accomplished by connecting  $R_{f4}$  and  $R_{f5}$  in the closed loop configuration . The gain for the PWM is determined by  $R_{f4}$  and  $R_{f5}$  where the gain is equal to  $-R_{f5}/R_{f4}$ . The voltage of VR<sub>set</sub> at which the controller goes into FM and PWM simultaneously is equal to :

VRset = D\_IN<sub>+</sub>  $\times$  (1+R<sub>f4</sub>/R<sub>f5</sub>) - (R<sub>f4</sub>/R<sub>f5</sub>)  $\times$  3 where 3 is the peak voltage of RTCT

Varyin R<sub>set</sub> from VR<sub>set</sub> to VR<sub>set</sub> + 1.75  $\times$  R<sub>f4</sub>/R<sub>f5</sub> will cause the duty cycle to vary from 50% to 0% while the frequency will vary proportionally according to 1.75\*R<sub>f4</sub>/R<sub>f5</sub> where 1.75 is the peak-to-peak voltage of the RTCT ramp. For proper operation, select R<sub>f5</sub> value so that the current through R<sub>f5</sub> {((D\_IN+) - 1.2)/R<sub>f5</sub>} should be less than 5uA.



SLS (SRC/LLC+SR) Controller with 1 FM+2 PWMs





### Soft Start and Enable

Soft start of the FM and PWM is controlled by the selection of the external capacitor at CSS pin. A current source of 7.5uA supplies the charging current for the capacitor. Soft start of the FM and PWM begins at 1.25V. The soft start pin CSS also serves as an enable function. The output drivers are enabled when CSS pin reached 1V.

### **Auto-retry Mode**

During normal operation, CSS pin will be charged to 5.5V + VBE. When VFB exceeds 2.93V due to ILIM exceeds 1V due to over current condition, output drivers are immediate set to low, and CSS begins to discharge with 0.75uA current source. When CSS pin is below 0.9V, 7.5uA current source start charging the CSS pin. When CSS pin reaches 1V, the output drivers are re-enable and the controller goes into soft start mode. If over current condition still exists then, the outputs are immediately disabled, but CSS continues to charge toward 5V. When CSS reaches 5V, 7.5uA charging current is disabled and 0.75uA current source begins to discharge the CSS, and the cycle repeat until over current condition is removed.



### **OCP Setting**



# TYPICAL APPLCATION CIRCUIT





### PACKAGE DIMENSION





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