

GENERAL DESCRIPTION

OB2263 is a highly integrated current mode PWM control IC optimized for high performance, low standby power and cost effective offline flyback converter applications in sub 30W range.

PWM switching frequency at normal operation is externally programmable and trimmed to tight range. At no load or light load condition, the IC operates in extended 'burst mode' to minimize switching loss. Lower standby power and higher conversion efficiency is thus achieved.

VDD low startup current and low operating current contribute to a reliable power on startup design with OB2263. A large value resistor could thus be used in the startup circuit to minimize the standby power.

The internal slope compensation improves system large signal stability and reduces the possible subharmonic oscillation at high PWM duty cycle output. Leading-edge blanking on current sense(CS) input removes the signal glitch due to snubber circuit diode reverse recovery and thus greatly reduces the external component count and system cost in the design.

OB2263 offers complete protection coverage with automatic self-recovery feature including Cycle-by-Cycle current limiting (OCP), over load protection (OLP), VDD over voltage clamp and under voltage lockout (UVLO). The Gate-drive output is clamped to maximum 18V to protect the power MOSFET.

Excellent EMI performance is achieved with On-Bright proprietary frequency shuffling technique together with soft switching control at the totem pole gate drive output.

Tone energy at below 20KHZ is minimized in the design and audio noise is eliminated during operation. OB2263 is offered in SOT23-6, SOP-8 and DIP-8 packages.

FEATURES

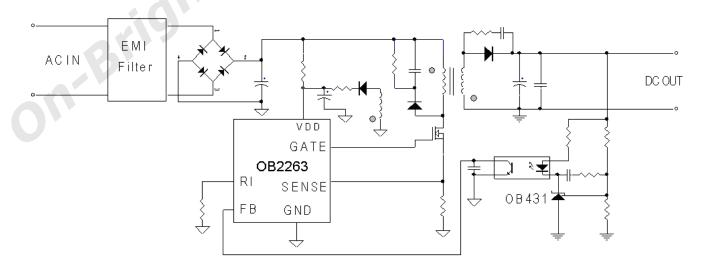
- On-Bright Proprietary Frequency Shuffling Technology for Improved EMI Performance.
- Extended Burst Mode Control For Improved Efficiency and Minimum Standby Power Design
- Audio Noise Free Operation
- External Programmable **PWM** Switching Frequency
- **Internal Synchronized Slope Compensation**
- Low VDD Startup Current and Low Operating Current (1.4mA)
- Leading Edge Blanking on Current Sense Input
- Good Protection Coverage With Auto Self-
 - VDD Over Voltage Clamp and Under Voltage Lockout with Hysteresis (UVLO)
 - Gate Output Maximum Voltage Clamp (18V)
 - On-Bright Proprietary Line Input Compensated Cycle-by-Cycle Over-current Threshold Setting For Constant Output Power Limiting Over Universal Input Voltage Range.
 - Overload Protection (OLP)

APPLICATIONS

Offline AC/DC flyback converter for

- **Battery Charger**
- Power Adaptor
- Set-Top Box Power Supplies
- Open-frame SMPS

TYPICAL APPLICATION

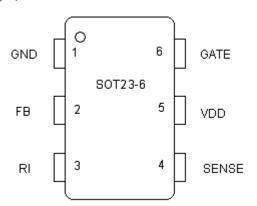


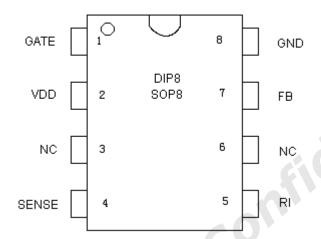


GENERAL INFORMATION

Pin Configuration

The OB2263 is offered in SOT23-6, DIP8 and SOP8 packages, shown as below.





Ordering Information

Part Number	Description
OB2263MP	SOT23-6, Pb-free in T&R
OB2263AP	DIP8, Pb-free in Tube
OB2263CP	SOP8, Pb-free in Tube
OB2263CPA	SOP8, Pb-free in T&R

Package Dissipation Rating

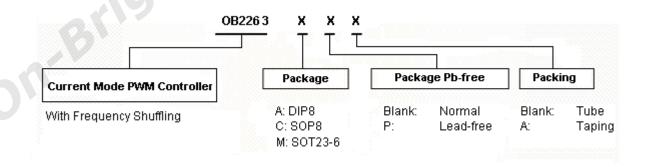
Package	RθJA (°C/W)	
DIP8	90	
SOP8	150	
SOT23-6	200	

Absolute Maximum Ratings

Parameter	Value
VDD DC Supply Voltage	30 V
VDD Zener Clamp	VDD_Clamp+0.1V
Voltage ^{Note}	
VDD DC Clamp Current	10 mA
FB Input Voltage	-0.3 to 7V
Sense Input Voltage	-0.3 to 7V
RI Input Voltage	-0.3 to 7V
Min/Max Operating	-20 to 150 °C
Junction Temperature T _J	
Min/Max Storage	-55 to 160 °C
Temperature T _{stg}	

Note: VDD_Clamp has a nominal value of 34V.

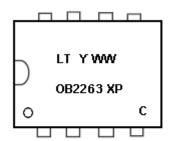
Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum-rated conditions for extended periods may affect device reliability.





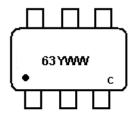
Marking Information

DIP8 SOP8



X: A for DIP 8 C for SOP8 P: Pb-free Package

Y: Year Cod e(0-9) WW: Week Code(1-52) C: Optional Internal Code SOT23-6



ue Y: Year Code(0-9) WW: Week Code(1-52) C: Optional Internal Code

TERMINAL ASSIGNMENTS

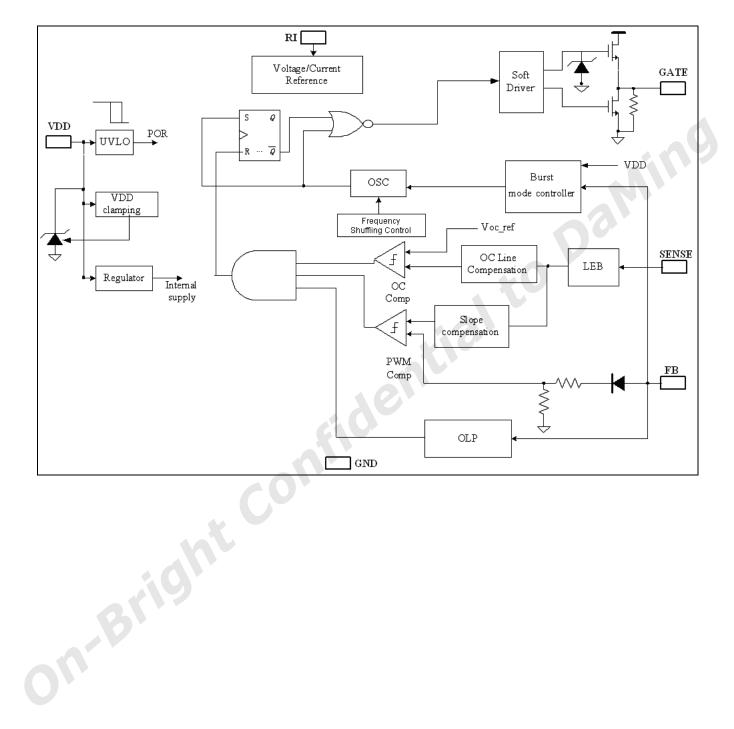
Pin Name	I/O	Description
GND	P	Ground
FB	I	Feedback input pin. The PWM duty cycle is determined by voltage level into this pin and
		SENSE pin input.
RI	I	Internal Oscillator frequency setting pin. A resistor connected between RI and GND sets
		the PWM frequency.
SENSE	I	Current sense input pin. Connected to MOSFET current sensing resistor node.
VDD	P	Chip DC power supply pin.
GATE	О	Totem-pole gate drive output for the power MOSFET.

RECOMMENDED OPERATING CONDITION

Symbol	Parameter	Min	Unit
		Max	
VDD	VDD Supply Voltage	10 to 30	V
RI	RI Resistor Value	100	Kohm
T_A	Operating Ambient Temperature	-20 to 85	°C



BLOCK DIAGRAM





ELECTRICAL CHARACTERISTICS

 $(T_A = 25^{\circ}C \text{ if not otherwise noted})$

$(T_A = 25^{\circ}C)$ if not of	Ź	Test Conditions	N/2	Т	Mar	TIm:4
Symbol Symbol Voltage (V	Parameter	Test Conditions	Min	Тур	Max	Unit
Supply Voltage (V		VDD-12 5V DI-100V	1	1 2	20	1 . A
I_VDD_Startup	VDD Start up	VDD=12.5V, RI=100K		3	20	uA
	Current	Measure Leakage current				
I VDD O	0 1 0 1	into VDD		1.4		A
I_VDD_Ops	Operation Current	VDD=16V,		1.4		mA
THAT O(ON)	VDD II 1 V 1	RI=100Kohm, V _{FB} =3V	7.0	0.0	0.0	3.7
UVLO(ON)	VDD Under Voltage		7.8	8.8	9.8	V
IIII O(OFF)	Lockout Enter		12	1.4	1.5	V
UVLO(OFF)	VDD Under Voltage		13	14	15	V
	Lockout Exit					
VDD Clama	(Recovery)	I — 5 m A		34		V
VDD_Clamp	VDD Zener Clamp	$I_{VDD} = 5 \text{ mA}$		34		V
Foodbook Innut C	Voltage				70.	
Feedback Input S		A \$7 / A \$7	1	2.0		17/17
A _{VCS}	PWM Input Gain	$\Delta V_{FB}/\Delta V_{cs}$				V/V
V _{FB} _Open	V _{FB} Open Loop		X	4.8		V
7 01	Voltage	gt . TD : GVD 1		1.0		
I _{FB} _Short	FB pin short circuit	Short FB pin to GND and		1.2		mA
** °D	current	measure current			0.7.5	
V_{TH}_0D	Zero Duty Cycle FB	VDD = 16V,			0.75	V
	Threshold Voltage	RI=100Kohm				
V_{TH}_PL	Power Limiting FB	4.0		3.7		V
	Threshold Voltage			10.5		
T _D _PL	Power limiting	23 (0)		35		mSec
7 DI	Debounce Time					TZ 1
Z _{FB} _IN	Input Impedance	1/00 101/		6		Kohm
DC_MAX	Maximum Duty	VDD=18V,		75		%
	Cycle	RI=100Kohm, FB=3V,				
C		CS=0				
Current Sense In		DI = 100 W -1	1	200		T
T_blanking	Leading edge	RI = 100 Kohm		300		ns
7 DI	blanking time			10		17. 1
Z _{SENSE} _IN	Input Impedance	VDD 16V		40		Kohm
T _D _OC	Over Current	VDD = 16V,		75		nSec
	Detection and	$CS>V_{TH}_OC$, $FB=3.3V$				
V OC	Control Delay	FD-2 2W DI-100 W-1	0.70	0.75	0.00	V
V_{TH} OC	Over Current	FB=3.3V, RI=100 Kohm	0.70	0.75	0.80	V
	Threshold Voltage at					
Oggillatar	zero Duty Cycle			1		
Oscillator	Narmal Ossill-4:	DI = 100 V - h	60	65	70	VII7
F_{OSC}	Normal Oscillation	RI = 100 Kohm	60	65	70	KHZ
	Frequency					
Af Tomp	Eraguanav	VDD = 16V,		5	+	%
Δf_Temp	Frequency Temperature Stability	VDD = 16V, RI=100Kohm, T_A -20°C] 3		70
	remperature Stability	to 100 °C				
Af VDD	Eraguanay Valtaga	VDD = 12-25V,		5	1	%
Δf_{VDD}	Frequency Voltage Stability			3		70
DI range	-	RI=100Kohm	50	100	150	Vohm
RI_range	Operating RI Range		30	100	150	Kohm
V_RI_open	RI open load voltage	VDD = 16V DI =		2	1	V
F _{osc} _BM	Burst Mode Base	VDD = 16V, RI =		22		KHZ



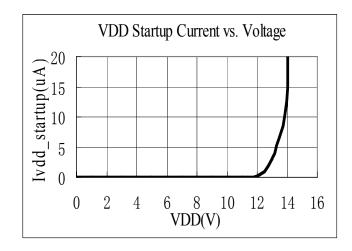
Current Mode PWM Controller Frequency Shuffling

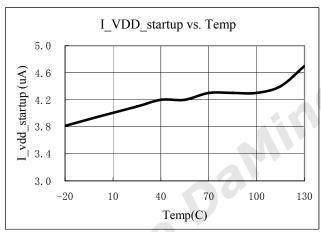
	Frequency	100Kohm				
Gate Drive Outp				_		
VOL	Output Low Level	VDD = 16V, Io = -20 mA			0.8	V
VOH	Output High Level	VDD = 16V, $Io = 20 mA$	10			V
V_Clamp	Output Clamp			18		V
	Voltage Level					
T_r	Output Rising Time	VDD = 16V, CL = 1nf		220		nSec
T_f	Output Falling Time	VDD = 16V, $CL = 1nf$		70		nSe
Frequency Shuff			1	1		
Δf_{OSC}	Frequency	RI=100K	-3		3	%
	Modulation range					
	/Base frequency					
f_shuffling	Shuffling Frequency	RI=100K		64		HZ

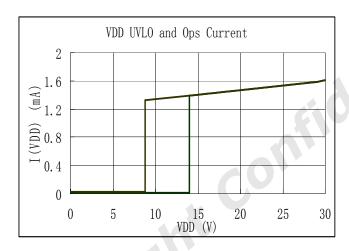


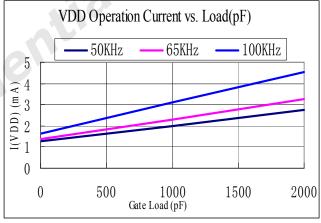
CHARACTERIZATION PLOTS

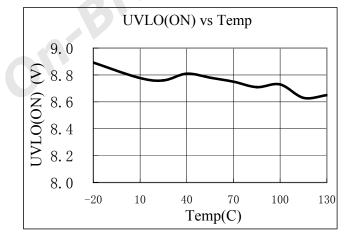
VDD = 16V, RI = 100 Kohm, $T_A = 25^{\circ}$ C condition applies if not otherwise noted.

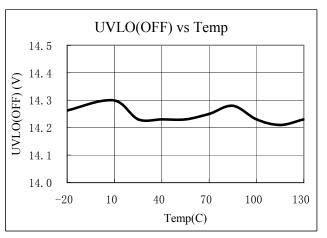




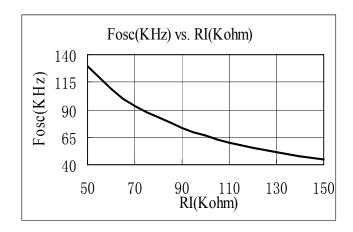


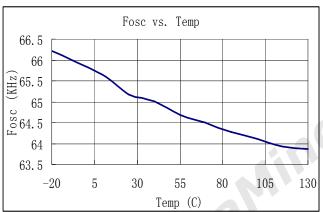


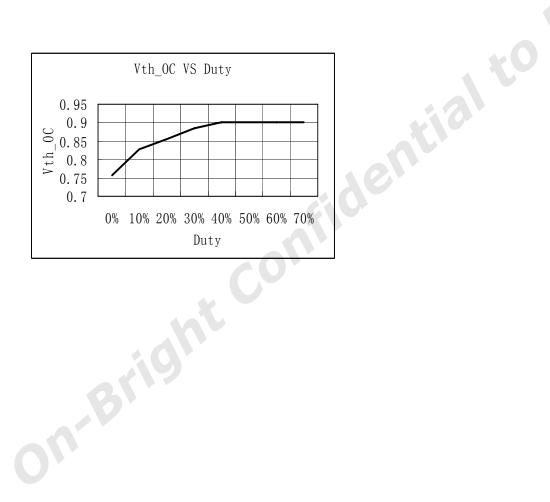














OPERATION DESCRIPTION

The OB2263 is a highly integrated PWM controller optimized for offline flyback converter applications in sub 30W power range. The extended burst mode control greatly reduces the standby power consumption and helps the design easily international meet the power conservation requirements.

Startup Current and Start up Control

Startup current of OB2263 is designed to be very low so that VDD could be charged up above UVLO threshold level and device starts up quickly. A large value startup resistor can therefore be used to minimize the power loss yet provides reliable startup in application. For AC/DC adaptor with universal input range design, a 2 M Ω , 1/8 W startup resistor could be used together with a VDD capacitor to provide a fast startup and low power dissipation solution.

Operating Current

The Operating current of OB2263 is low at 1.4mA. Good efficiency is achieved with OB2263 low operating current together with extended burst mode control features.

Frequency shuffling for EMI improvement

The frequency Shuffling/jittering (switching frequency modulation) is implemented in OB2263. The oscillation frequency is modulated with a random source so that the tone energy is spread out. The spread spectrum minimizes the conduction band EMI and therefore reduces system design challenge.

Extended Burst Mode Operation

At zero load or light load condition, majority of the power dissipation in a switching mode power supply is from switching loss on the MOSFET transistor, the core loss of the transformer and the loss on the snubber circuit. The magnitude of power loss is in proportion to the number of switching events within a fixed period of time. Reducing switching events leads to the reduction on the power loss and thus conserves the energy. OB2263 self adjusts the switching mode according to the loading condition. At from no load to light/medium load condition, the FB input drops below burst mode threshold level. Device enters Burst Mode control. The Gate drive output switches only when VDD voltage drops below a preset level and FB input is active to output an on state. Otherwise the gate drive remains at off state to

minimize the switching loss and reduces the standby power consumption to the greatest extend. The frequency control also eliminates the audio noise at any loading conditions.

Oscillator Operation

A resistor connected between RI and GND sets the constant current source to charge/discharge the internal cap and thus the PWM oscillator frequency is determined. The relationship between RI and switching frequency follows the below equation within the specified RI in Kohm range at nominal loading operational condition.

$$F_{OSC} = \frac{6500}{RI(Kohm)}(Khz)$$

Current Sensing and Leading Edge Blanking

Cycle-by-Cycle current limiting is offered in OB2263 current mode PWM control. The switch current is detected by a sense resistor into the sense pin. An internal leading edge blanking circuit chops off the sense voltage spike at initial MOSFET on state due to Snubber diode reverse recovery so that the external RC filtering on sense input is no longer required. The current limit comparator is disabled and thus cannot turn off the external MOSFET during the blanking period. PWM duty cycle is determined by the current sense input voltage and the FB input voltage.

Internal Synchronized Slope Compensation

Built-in slope compensation circuit adds voltage ramp onto the current sense input voltage for PWM generation. This greatly improves the close loop stability at CCM and prevents the sub-harmonic oscillation and thus reduces the output ripple voltage.

Gate Drive

OB2263 Gate is connected to an external MOSFET gate for power switch control. Too weak the gate drive strength results in higher conduction and switch loss of MOSFET while too strong gate drive output compromises the EMI.

A good tradeoff is achieved through the built-in totem pole gate design with right output strength and dead time control. The low idle loss and good EMI system design is easier to achieve with this dedicated control scheme. An internal 18V clamp is added for MOSFET gate protection at higher than expected VDD input.



Protection Controls

Good power supply system reliability is achieved with its rich protection features including Cycle-by-Cycle current limiting (OCP), Over Load Protection (OLP) and over voltage clamp, Under Voltage Lockout on VDD (UVLO).

With On-Bright Proprietary technology, the OCP Awn device der.

On-Bright: Confidential to Daniel Confidential to D threshold tracks PWM Duty cycles and is line voltage compensated to achieve constant output

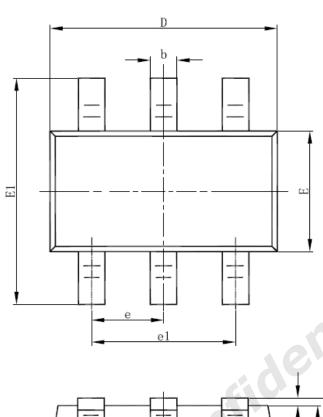
At overload condition when FB input voltage exceeds power limit threshold value for more than TD PL, control circuit reacts to shut down the output power MOSFET. Device restarts when VDD voltage drops below UVLO limit.

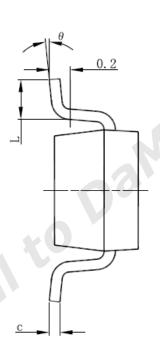
VDD is supplied by transformer auxiliary winding output. It is clamped when VDD is higher than threshold value. The power MOSFET is shut down when VDD drops below UVLO limit and device

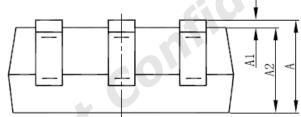


PACKAGE MECHANICAL DATA

SOT-23-6L PACKAGE OUTLINE DIMENSIONS



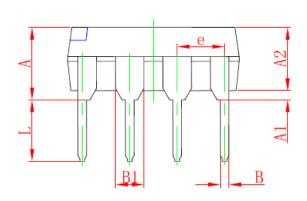


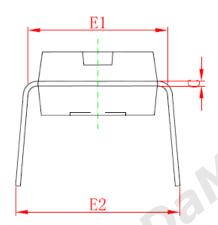


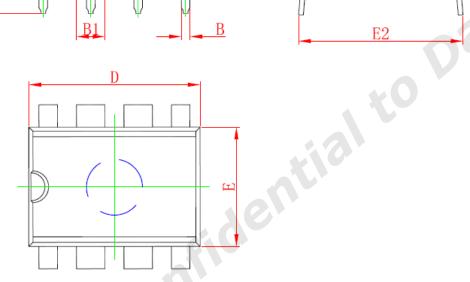
Symbol	Dimensions In Millimeters		Dimension	s In Inches
Symbol	Min	Max	Min	Max
A	1.000	1.300	0.039	0.051
A1	0.000	0.150	0.000	0.006
A2	1.000	1.200	0.039	0.047
b	0.300	0.500	0.012	0.020
c	0.100	0.200	0.004	0.008
D	2.800	3.020	0.110	0.119
Е	1.500	1.700	0.059	0.067
E1	2.600	3.000	0.102	0.118
e	0.950 (BSC)		0.037	(BSC)
e1	1.800	2.000	0.071	0.079
L	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°



DIP8 PACKAGE OUTLINE DIMENSIONS



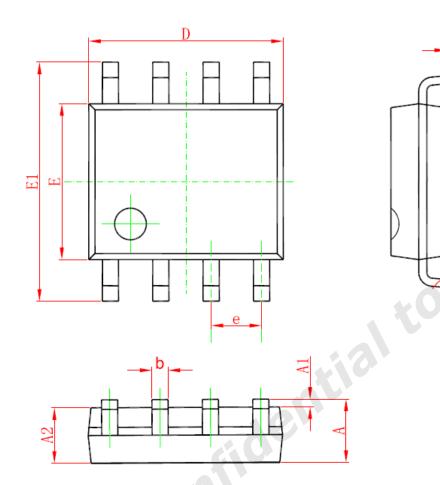




Cymbol	Dimensions In Millimeters		Dimension	s In Inches
Symbol	Min	Max	Min	Max
A	3.710	5.334	0.146	0.210
A1	0.381		0.015	
A2	3.175	3.600	0.125	0.142
В	0.350	0.650	0.014	0.026
B1	1.524	(BSC)	0.06 (BSC)
C	0.200	0.360	0.008	0.014
D	9.000	10.160	0.354	0.400
Е	6.200	6.600	0.244	0.260
E1	7.320	7.920	0.288	0.312
e	2.540	(BSC)	0.1 (BSC)	
L	2.921	3.810	0.115	0.150
E2	8.200	9.525	0.323	0.375



SOP8 PACKAGE OUTLINE DIMENSIONS



Symbol	Dimensions In Millimeters		Dimensions	s In Inches
Symbol	Min	Max	Min	Max
A	1.350	1.750	0.053	0.069
A1	0.050	0.250	0.002	0.010
A2	1.250	1.650	0.049	0.065
b	0.310	0.510	0.012	0.020
c	0.170	0.250	0.006	0.010
D	4.700	5.150	0.185	0.203
Е	3.800	4.000	0.150	0.157
E1	5.800	6.200	0.228	0.244
e	1.270 (BSC)		0.05 (BSC)
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°



IMPORTANT NOTICE

RIGHT TO MAKE CHANGES

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