

Power Factor Controller IC for High Power Factor

Features

- Critical conduction mode of PFC pre-regulators
- Very low startup current (<50uA)
- Under-voltage lockout with hysteresis
- Zero current detector for critical conduction mode
- Dynamic and static output over-voltage protection
- Gate output maximum voltage clamped
- Internal restart timer
- Internal leading-edge blanking
- ±500mA totem pole output with active shut down

Applications

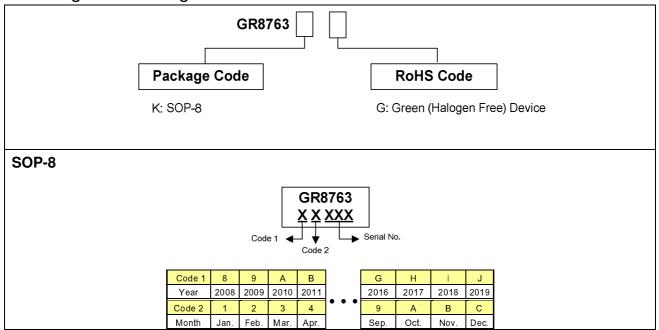
- Open-frame power supply and power adapter
- Flat TV monitor
- Desktop PC
- Electronic lamp ballast

Description

The GR8763 is a current-mode PFC controller operating in critical conduction mode. The sinusoidal current is taken from the single phase line supply and stabilized DC voltage is available at the output. The IC includes a highly liner multiplier which can reduce AC input current distortion, that allows wide-range-mains operation with an extremely low THD, even over a large load range.

The output voltage is controlled by an error amplifier and an internal voltage reference. An effective two-step (dynamic and static) OVP enables to safely handle over voltages either occurring at start-up or resulting from load disconnection. The totem-pole output stage capable of ±500mA driving capability is suitable for big MOSFET or IGBT. It is available in 8-pin SOP package.

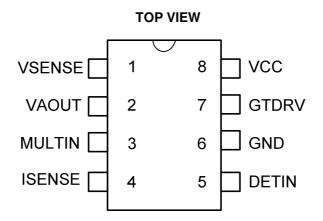
Ordering and Marking Information



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Pin Configuration



Pin Description

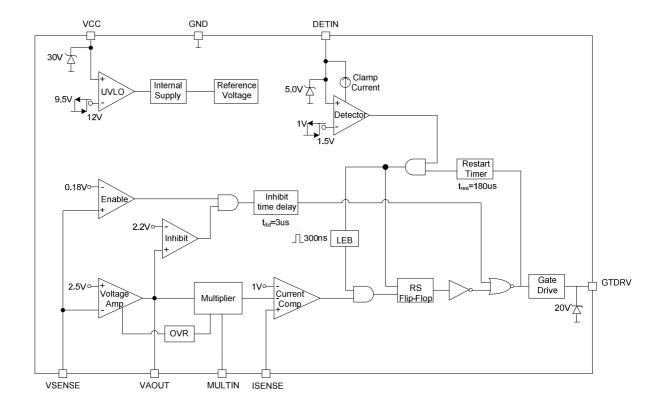
Pin	Symbol	Description
		Voltage Amplifier Inverting Input. VSENSE is connected via a resistive divider to
1	VSENSE	the boost convert output, With a capacitor connected to VAOUT, The internal
		error amplifier acts as an integrator.
		Voltage Amplifier Output. This is the first multiplier input. If the current flowing into
2	VAOUT	this pin is exceeding an internal threshold, the multiplier output voltage is reduced
		to prevent the MOSFET from over-voltage damage.
3	MULTIN	Multiplier Input, Which is the second multiplier input and is connected via a
3	MOLITIN	resistive divider to the rectifier output voltage.
		Current Sense Input, Which is connected to a sense resistor controlling the
4	ISENSE	MOSFET source current .A leading edge blanking circuitry, suppresses voltage
		spits when turning the MOSFET on. Its max voltage is internally limited at 1V.
F	DETIN	Zero Current Detector Input, Which is connected to an auxiliary winding
5	DETIN	monitoring the zero crossing of the inductor current.
6	GND	Ground
7	GTDRV	Gate Drive Output, Which is the output of a totem-pole circuitry for direct driving a
7		MOSFET.
		Positive Voltage Supply. If VCC exceeds the UVLO (on) threshold, the IC is
8	VCC	switched on. When VCC falls below the UVLO (off) threshold, it is switched off
		and power consumption is very low. VCC is internally clamped 30V.



Absolute Maximum Ratings

Supply voltage VCC	- 30V
VSENSE, MULTIN, ISENSE, DETIN	3 ~ 6.5V
Junction temperature	150℃
Operating ambient temperature	~ 85℃
Storage temperature range	~ 150℃
SOP-8 package thermal resistance	160℃/W
Power dissipation (SOP-8, at ambient temperature = 85°C)	400mW
Lead temperature (All Pb free packages, soldering, 10sec)	260 ℃
ESD voltage protection, human body model	2KV
ESD voltage protection, machine model	200V

Block Diagram





Electrical Characteristics (Ta = 25°C & Vcc = 14.5V, unless otherwise specified.)

Parameter	Symbol & Condition	Min.	Тур.	Max.	Unit		
SUPPLY VOLTAGE							
Zener Voltage	Vz; lcc+lz = 20mA	27	30	33	V		
UVLO on Voltage	Vcc-on	11	12.0	13	V		
UVLO off Voltage	Vcc-off	8.5	9.5	10.5	V		
Vcc Hysteresis	Vcc-HY		2.5		V		
SUPPLY CURRENT							
Startup Current	ICCL		30	50	uA		
Operating Current	ІССН		2.5		mA		
VOLTAGE AMPLIFIER							
Voltage feedback Input Theshold	VFB	2.45	2.5	2.55	V		
Line Regulation(VFBLR)	Vcc = 12V~25V			5	mV		
Enable Threshold	Vsense		0.18		V		
Inhibit Threshold Voltage	VVAOUT		2.2		V		
Voltage Gain	Gv		90		dB		
Gain – Bandwidth	Bw		3		MHz		
Output Current Source (IVAOUTH)	VVAOUT = 0V; Vsense = 2.3V		-4		mA		
Output Current Sink (IVAOUTL)	VVAOUT = 4V; Vsense = 2.8V		6		mA		
Upper Clamper Voltage	Vvaouth		3.5		V		
Lower Clamp Voltage	Vvaoutl		2.0		V		
MULTIPLIER INPUT							
Dynamic Voltage Range Multin	VMultin	0	3.5		V		
	Klow,		0.0				
Multiplier Gain	VAOUT<3.5,Vmultin =1V		0.2				
	KHigh, VAOUT>3.5,Vmultin =1V		0.5				
OVER-VOLTAGE REGULATOR							
Threshold Current	IOVR	25	30	35	uA		

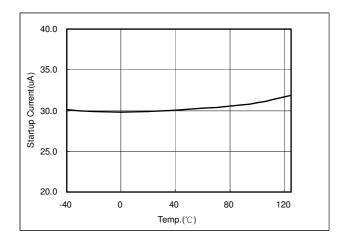


Electrical Characteristics (Cont.)

Parameter	Symbol & Condition	Min.	Тур.	Max.	Unit		
CURRENT COMPARATOR							
Max Threshold Voltage	Visense	0.95	1.0	1.05	V		
Leading Edge Blanking	TLEB	200	300	400	ns		
Shut Down Delay	Tdelay		100		ns		
DETECTOR							
Upper Threshold Voltage	VDETU		1.5		V		
Lower Threshold Voltage	VDETL		1.0		V		
Hysteresis	HYS		0.5		V		
Zener Voltage	Vz-Detin	4.5	5.0	5.5	V		
RESTART TIME							
Restart Time	TRES		180		us		
GATE DRIVER							
Output Low Voltage	Io = 20mA			1.0	V		
Output High Voltage	lo = 20mA	8.0			V		
Rising Time	Trise; Load = 1nF		50		ns		
Falling Time	TFALL; Load = 1nF		30		ns		
Clamp Vclamp			20		V		



Typical Performance Characteristics



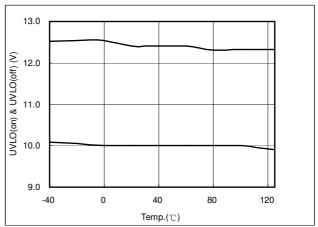


Fig. 1

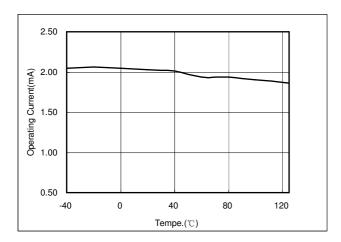


Fig. 2

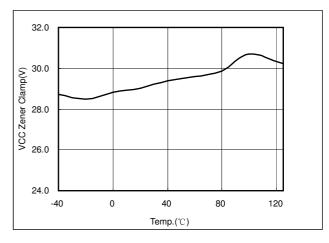


Fig. 3

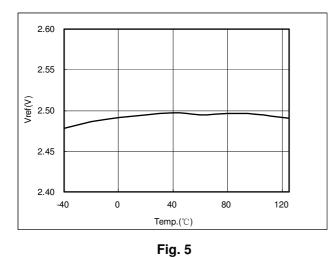


Fig. 4

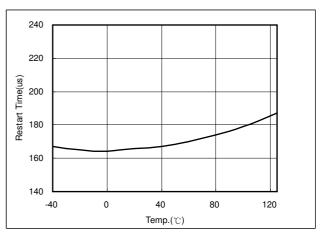
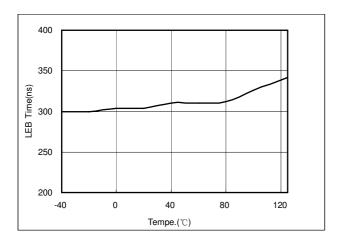


Fig. 6



Typical Performance Characteristics (Cont.)



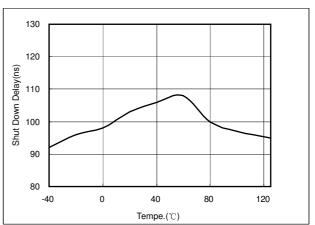


Fig. 7



Functional Description

Overview

The GR8763 contains an one quadrant multiplier with a wide linear operating range, a voltage amplifier used in a feedback loop, an overvoltage regulator, a current sense comparator, a zero current detector, a PWM and logic circuit, a totem pole MOSFET driver, an internal voltage reference, a restart timer and under voltage lockout circuit.

Multiplier

The one quadrant multiplier has two inputs M1 and M2. The input M1 (MULTIN Pin) provides the information of the waveform of the input voltage. Its range is from 0 to 3.5V being referenced to ground. The input M2 (VAOUT Pin) is the voltage amplifier output, which provides the information of the DC output voltage. Both inputs are designed to achieve good linearity over a wide dynamic range to represent an AC line free from distortion.

Voltage Amplifier

With an external capacitor between the pins VSENSE and VAOUT the voltage amplifier forms an integrator. The integrator monitors the average output voltage over several line cycles. Typically the integrator's bandwidth is set below 20Hz in order to suppress the 100Hz ripple of the rectified line voltage. The non-inverting input is biased internally at 2.5V. The output is directly connected to the multiplier input.

The gate drive is disabled when VSENSE voltage is less than 0.18V or VAOUT voltage is less than 2.2V.

Overvoltage Regulator

Because of the integrator's low bandwidth fast changes of the output voltage cant's be regulated within an adequate time. Fast output changes occur during initial start-up, sudden load removal. While the integrator's differential input voltage remains zero.

During this fast changes, a peak current is flowing through the external capacitor into pin VAOUT. If this current exceeds an internal defined margin the overvoltage regulator circuit reduces the multiplier output voltage. As a result the on time of the MOSFET is reduced.

Current Sense Comparator

An external sense resistor transfers the source current of the MOSFET into a sense voltage. The multiplier output is compared with this sense voltage. The switch-on peak current of the MOSFET is blanked out via a leading edge blanking circuit with a blanking time of typically 300ns.



Functional Description (Cont.)

Zero Current Detector

The zero current detector senses the inductor current via an auxiliary winding and ensures that the next on-time of the MOSFET is initiated immediately when the inductor current has reached zero. This diminishes the reverse recovery losses of the boost converter diode. The MOSFET is switched off when the voltage drop of the sense resistor reaches the voltage level of the multiplier output. So the boost current waveform has a triangular shape and there are no dead-time gaps between the cycles. This leads to a continuous AC line current limiting the peak current to twice of the average current.

To prevent false tripping the zero current detector is designed as a Schmitt-Trigger with a hysteresis of 0.5V. An internal 5.0V clamp protects the input from overvoltage breakdown. An external resistor has to

be used in series with the auxiliary winding to limit the current through the clamps.

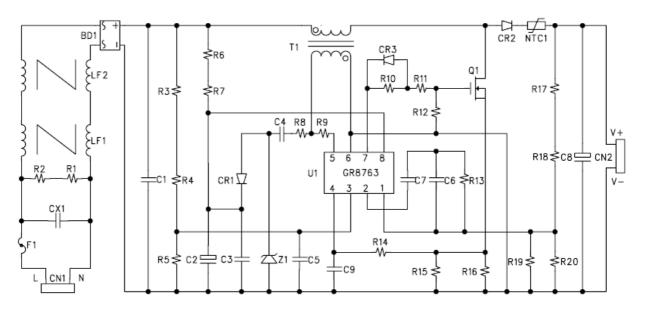
Restart Timer

The restart timer function eliminates the need of a oscillator. The timer starts or restarts the GR8763 when the drive output has been off for more than 180us after the inductor current reaches zero.

Under-voltage Lockout

An under voltage lockout circuit switches the IC on when Vcc reaches the UVLO(ON) threshold and switches the IC off when Vcc is falling below the UVLO(OFF) threshold. During star up the supply During star up the supply current is less than 50uA. An internal voltage clamp has been added to protect the IC from Vcc overvoltage.

Typical Application Circuit





BOM List

No.	Part	Description	Quan.
1	BD1	PBL405 4A 600V	1
2	CN1, CN2	2/3 pin 3.96 180°, white	1
3	C1	474K 400V Pitch=15mm	1
4	C2	47uF 35V 6*12mm 105℃	1
5	C3	0805 104K 50V X7R	1
6	C4	103K 100V X7R	1
7	C5, C6	0805 103K 50V X7R	2
8	C7	0805 225K 50V X7R	1
9	C8	47uF/450V 22*26(max) 105°C	1
10	C9	NC	1
11	CX1	474K/275V	1
12	CR1	FR104 1A 400V	1
13	CR2	ER506 5A/800V DO-201	1
14	CR3	LL4148 DO-80	1
15	F1	4*8mm T3.15A/250V, fuse	1
16	HS1	15*10.5*25mm heatsink, Alminium	1
17	LF1	T12*10*8c 0.7mH	1
18	LF2	T16*12*8c 15mH	1
19	NTC1	SCK2R55A	1
20	Q1	2SK2842 12A/500V TO-220	1
21	R1, R2, R3, R4	1206 1M 5%	4
22	R5	20K 0805 5%	1
23	R6, R7	1206 240K 5%	2
24	R8	100R 1206 5%	1
25	R9	39K 1206 5%	1
26	R10	33R 1206 5%	1
27	R11	0R 1206 5%	1
28	R12	100K 0805 5%	1
29	R13	4.7K 0805 5%	1
30	R14	OR 0805 5%	1
31	R15	0.27R 2WS 5%	1
32	R16	NC	0

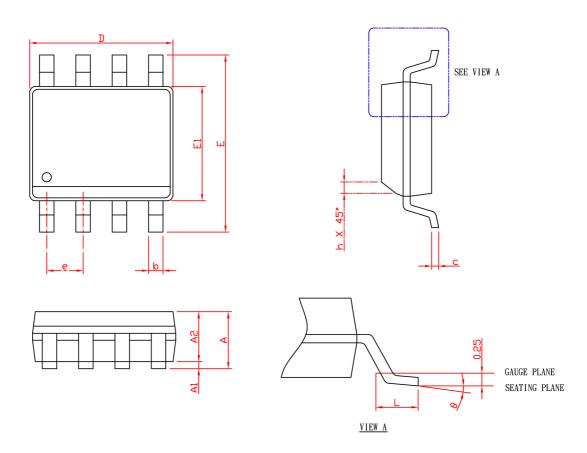


BOM List (Cont.)

No.	Part	Description	Quan.
33	R17, R18	330K 1206 1%	2
34	R19	4.3K 0805 1%	1
35	R20	91K 0805 1%	1
36	T1	PQ2620 0.7mH	1
37	U1	SOP-8 GR8763	1
38	Z1	18V 1W	1
39	PCB	CEM-1 2oz 1mm 110*56mm	1



Package Information



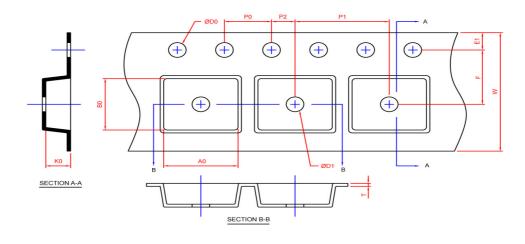
	SOP-8						
SYMBOL	MILLIM	ETERS	INCHES				
	MIN.	MAX.	MIN.	MAX.			
Α		1.75		0.069			
A1	0.10	0.25	0.004	0.010			
A2	1.25		0.049				
b	0.31	0.51	0.012	0.020			
С	0.17	0.25	0.007	0.010			
D	4.80	5.00	0.189	0.197			
E	5.80	6.20	0.228	0.244			
E1	3.80	4.00	0.150	0.157			
е	1.27	BSC	0.050	BSC			
h	0.25	0.50	0.010	0.020			
L	0.40	1.27	0.016	0.050			
θ	0°	8°	0°	8°			

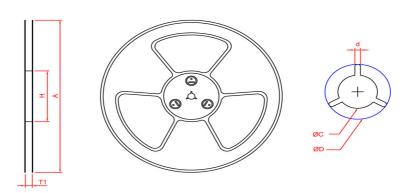
Note: 1. Followed from JEDEC MS-012 AA.

- 2. Dimension "D" does not include mold flash, protrusions or gate burrs. Mold flash, protrusion or gate burrs shall not exceed 6 mil per side.
- 3. Dimension "E1" does not include inter-lead flash or protrusions. Inter-lead flash and protrusions shall not exceed 10 mil per side.



Carrier Tape & Reel Dimensions





Application	Α	Н	T1	С	d	D	W	E1	F
SOP-8	330.0 <u>+</u> 2.0	50 MIN.	12.4+2.00 -0.00	13.0+0.50 -0.20	1.5 MIN.	20.2 MIN.	12.0±0.30	1.75 <u>+</u> 0.10	5.5±0.05
	P0	P1	P2	D0	D1	Т	A0	В0	K0
	4.0±0.10	8.0 <u>±</u> 0.10	2.0 <u>±</u> 0.05	1.5+0.10 -0.00	1.5 MIN.	0.6+0.00 -0.40	6.40±0.20	5.20±0.20	2.10±0.20

(mm)

Devices Per Unit

Application	Carrier Width	Cover Tape Width	Devices Per Reel
SOP-8	12	-	2500

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