

IGBT Double Pulse Test



Never stop thinking

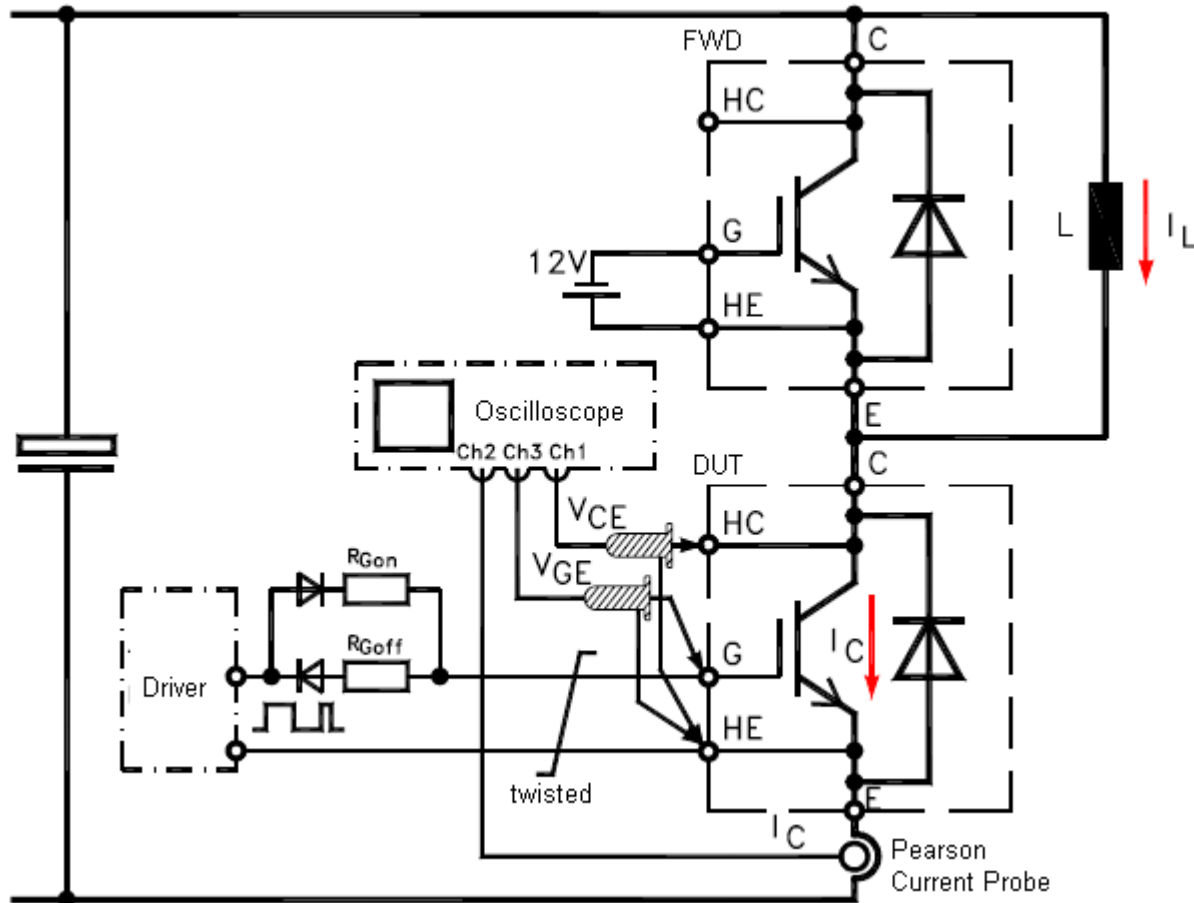
IGBT Double Pulse Test

- Basic principle of double pulse test
- Safe operation of IGBT
- What can be done with double pulse test
- Impact of R_g , C_{ge} on IGBT switching

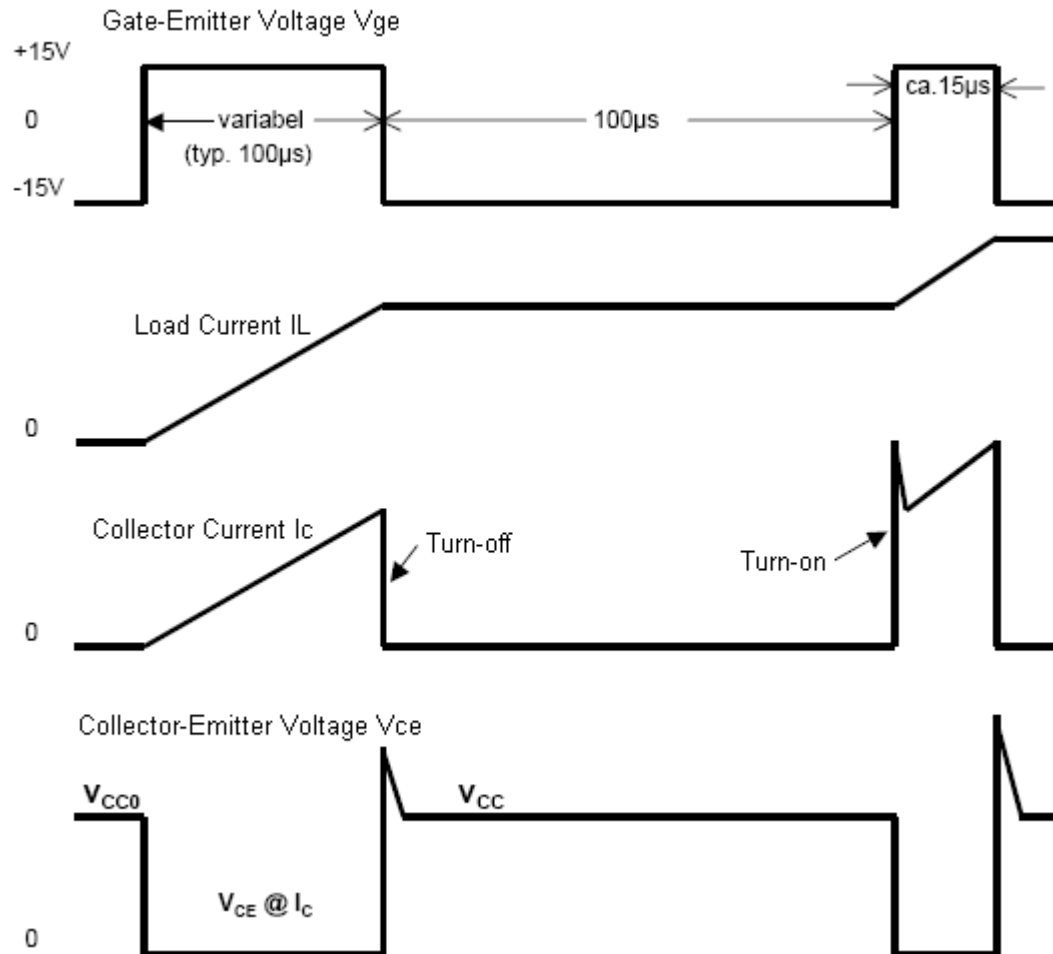
IGBT Measurements

Turn-on & Turn-off

■ Test Setup

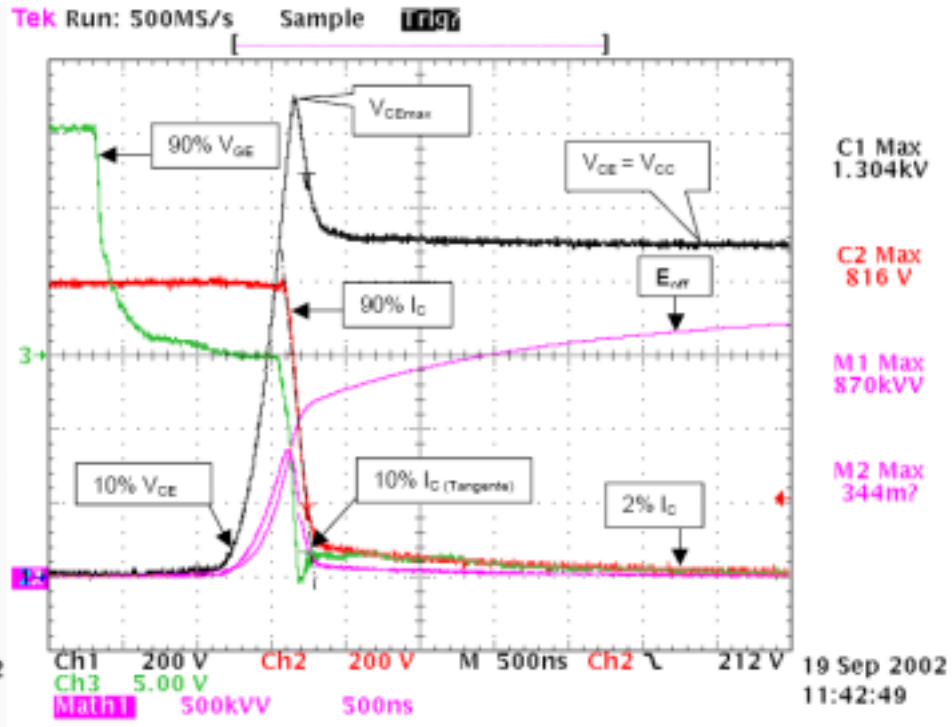
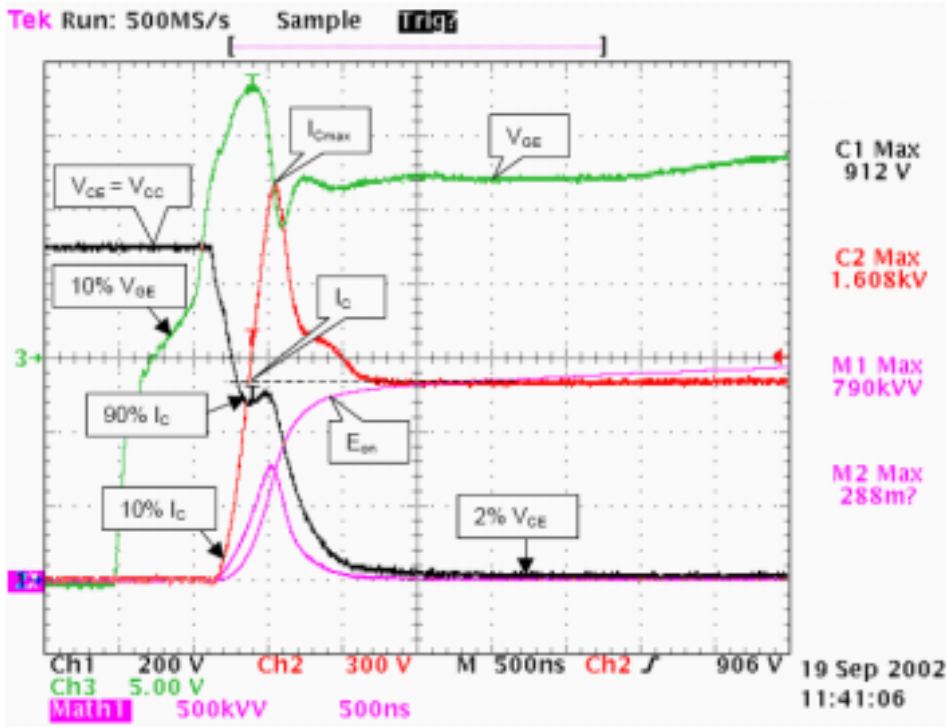


■ Basic Wave Forms



IGBT Measurements Turn-on & Turn-off

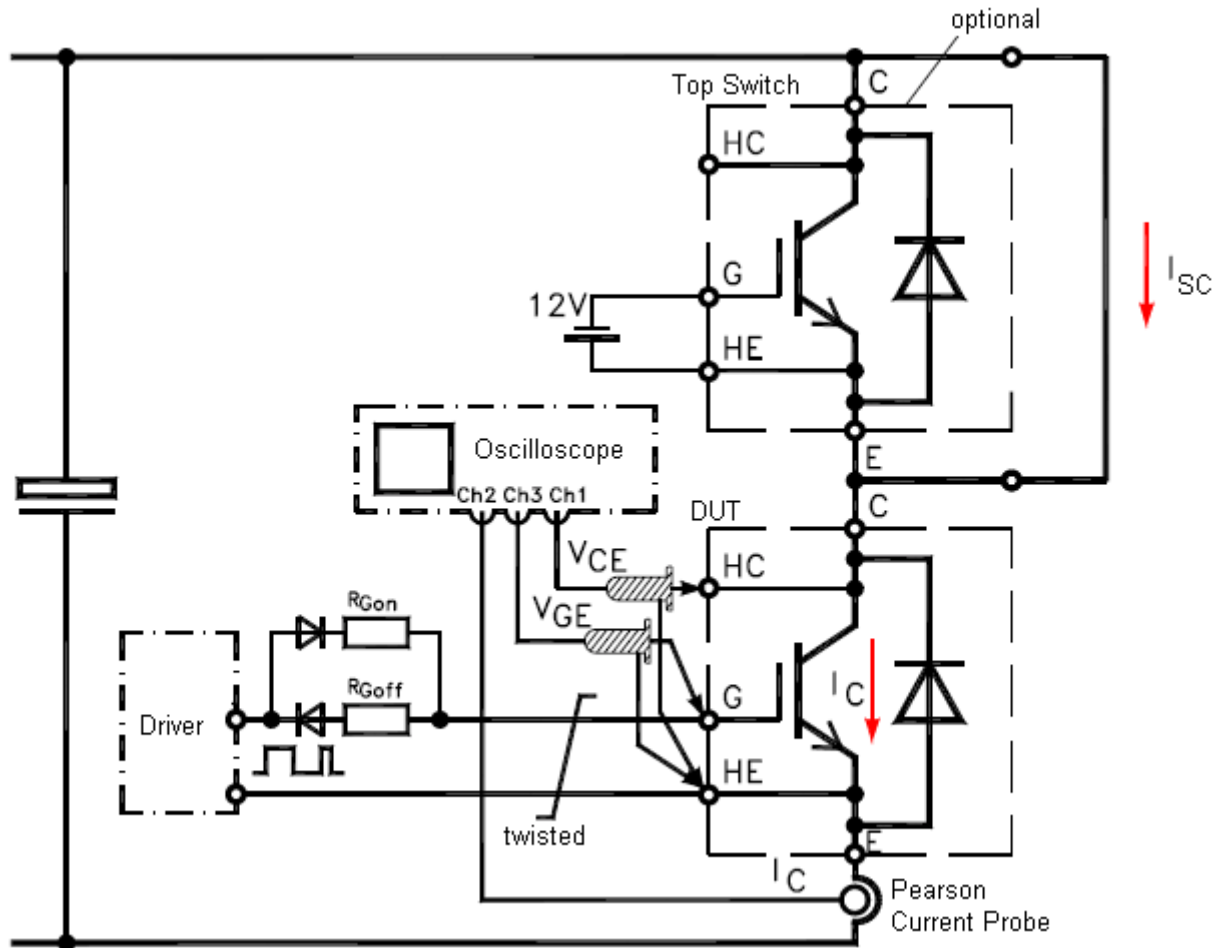
■ Sample Wave Forms



IGBT Measurements

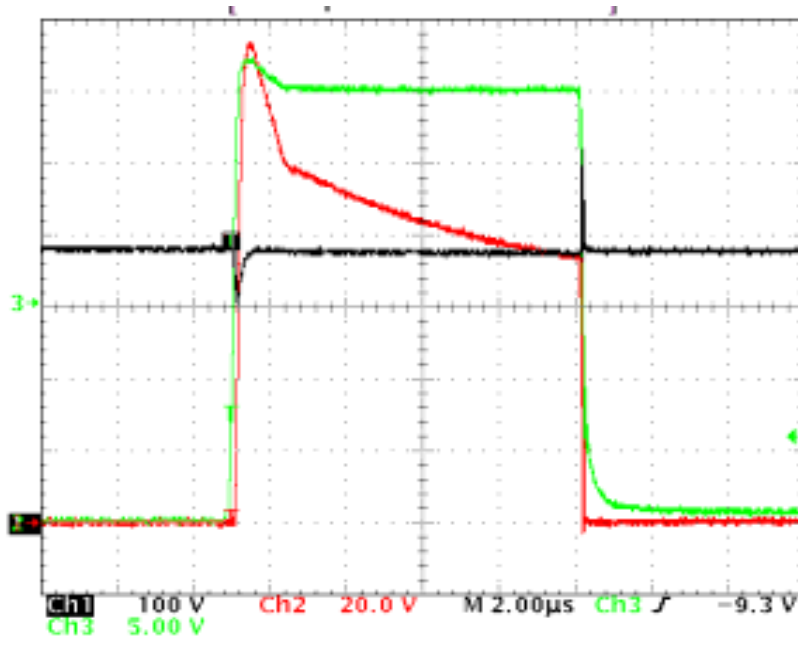
Short Circuit I & II

■ Test Setup



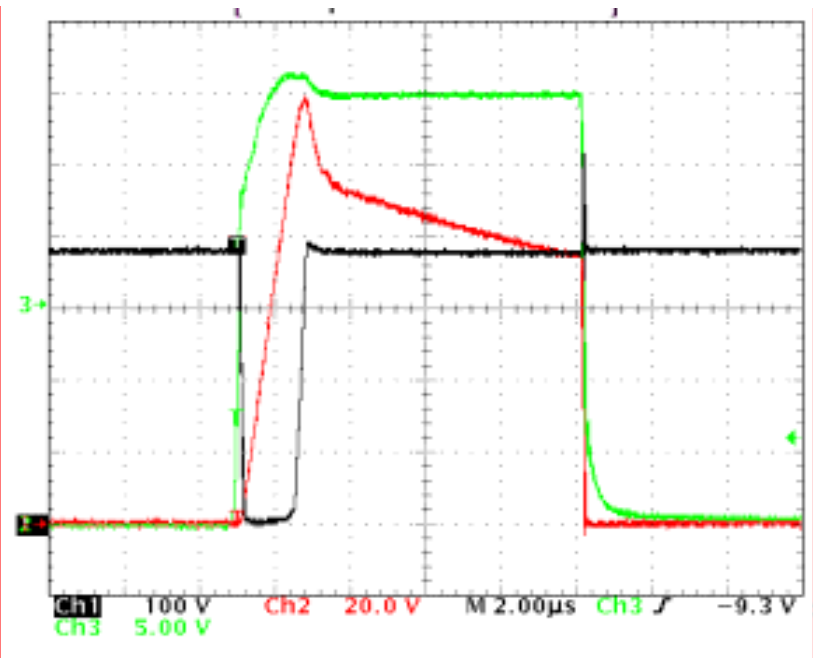
Two Types of Short-circuits

- Before short-circuit occurs, IGBT is OFF & blocks the DC-bus voltage.
- The short-circuit is created by the switch-on of the IGBT.



V_{CE} never reaches V_{CEsat} value!

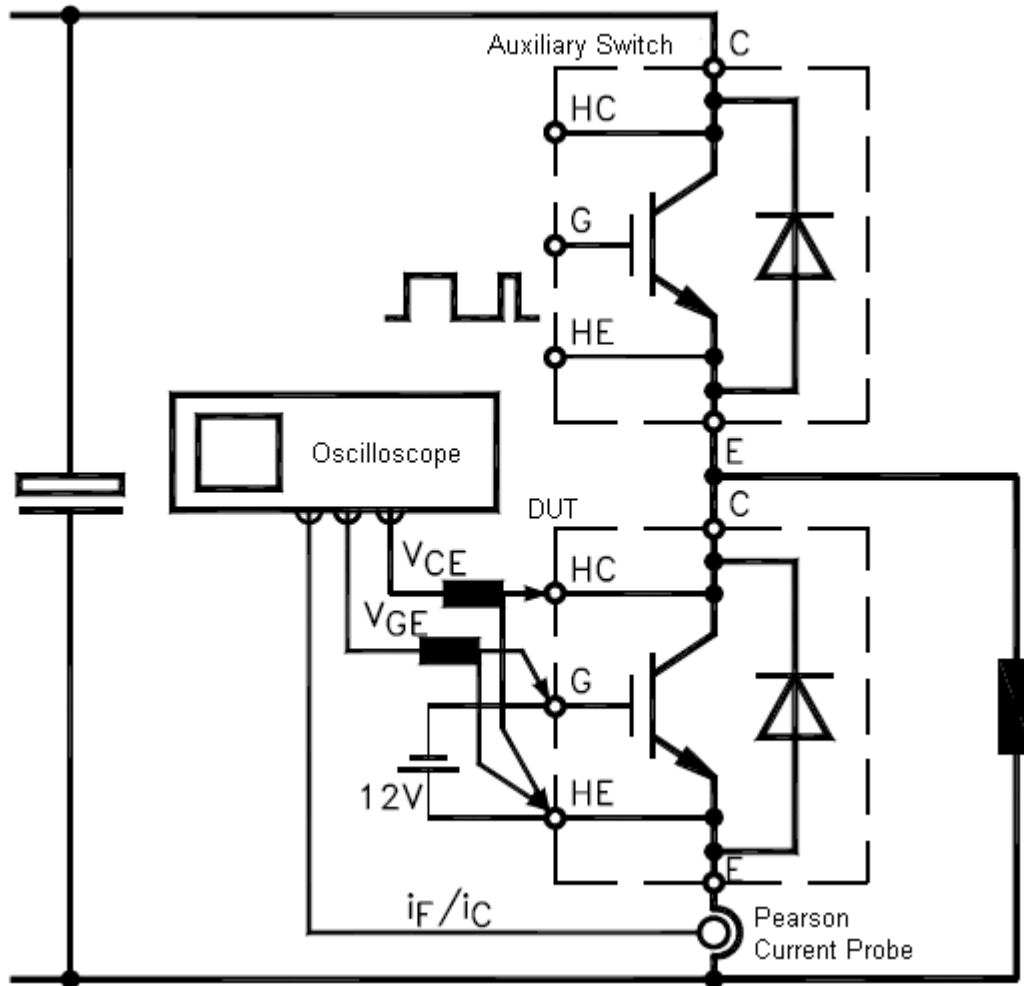
- Before the short-circuit occurs, IGBT is ON & in saturation region.
- The short-circuit is created by applying the DC-bus voltage on C-E of the IGBT.



VCE de-saturates from V_{CEsat} value!

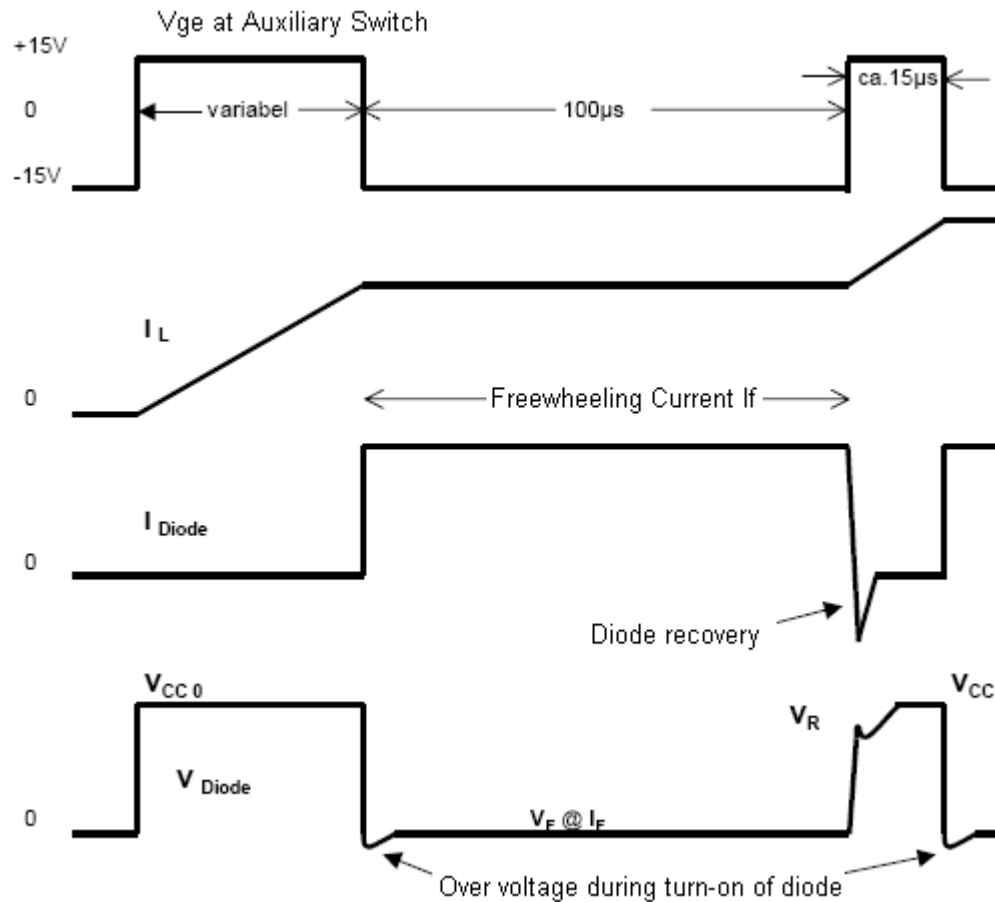
Diode Measurements

■ Test Setup



Diode Measurements Recovery

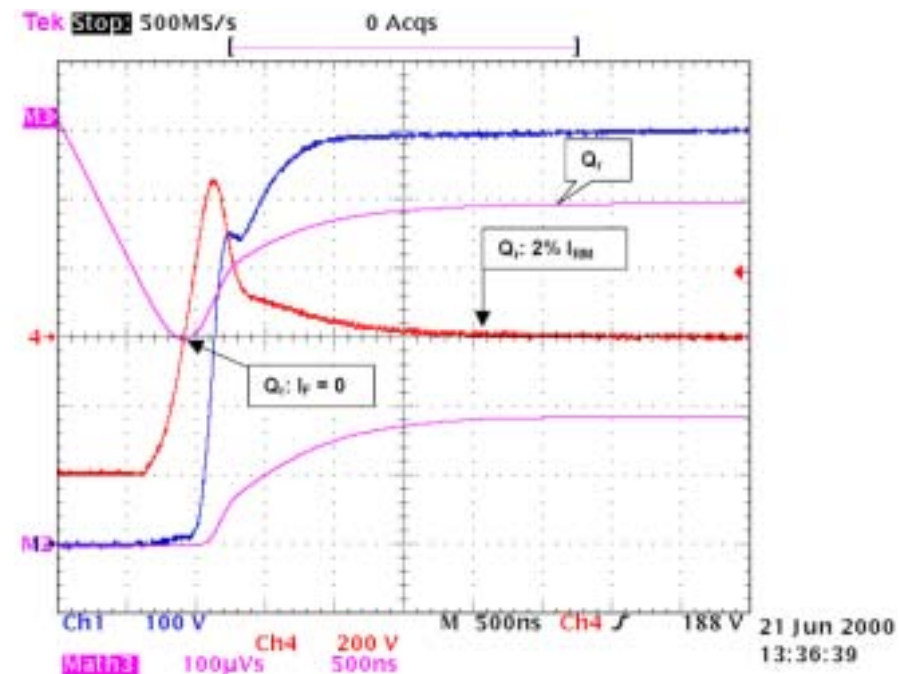
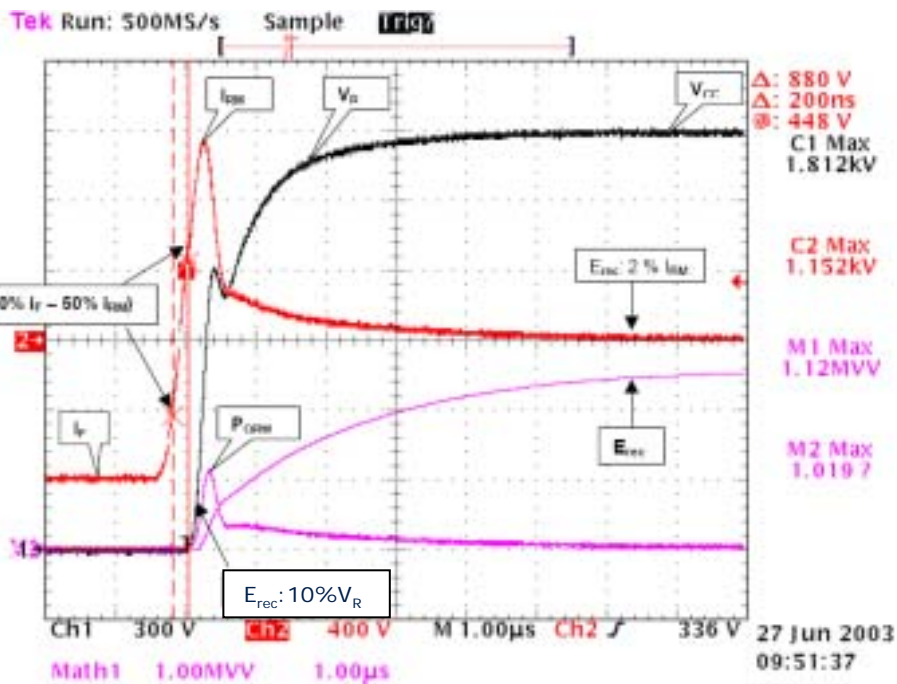
■ Basic Wave Forms



Diode Measurements Recovery

■ Sample Wave Form

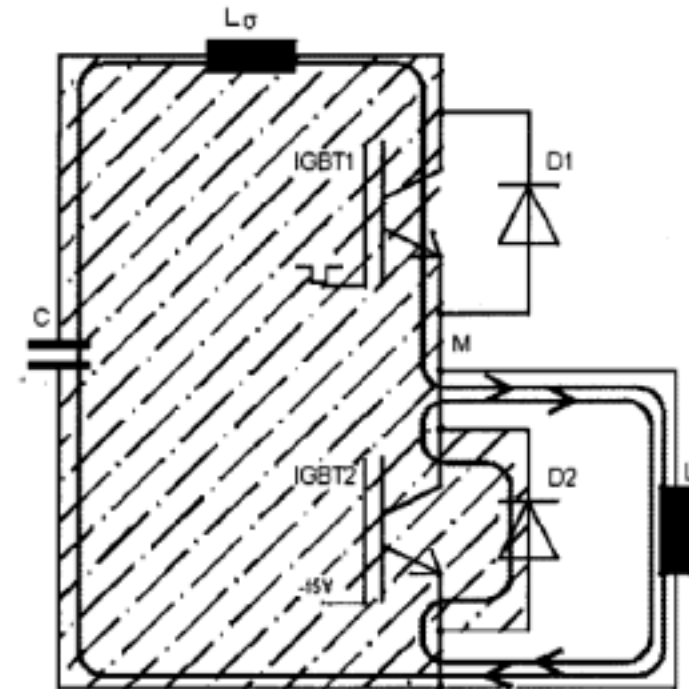
| d_{iF}/dt | E_{rec} | Q_r |
|--------------------------|-------------------------|------------------------|
| $50\% I_F - 50\% I_{RM}$ | $10\% V_R - 2\% I_{RM}$ | $I_F = 0 - 2\% I_{RM}$ |



Double Pulse Test

- **IGBT characterization**
 - Comparable test conditions as datasheet
 - Results close to Infineon datasheet expected.

- **Customer's application setup**
 - Helpful for further design
 - Different test conditions, and different results expected
 - Both IGBT in the (2-level) leg to be tested – different commutation loop & different behavior



IGBT Double Pulse Test

- Basic principle of double pulse test

- Safe operation of IGBT

 - IGBT RBSOA

 - Diode SOA

 - Short circuit

 - Vge limit

 - Others – Not tested by double pulse test

- What can be done with double pulse test

- Impact of R_g , C_{ge} on IGBT switching

IGBT Safe Operation – IGBT RBSOA

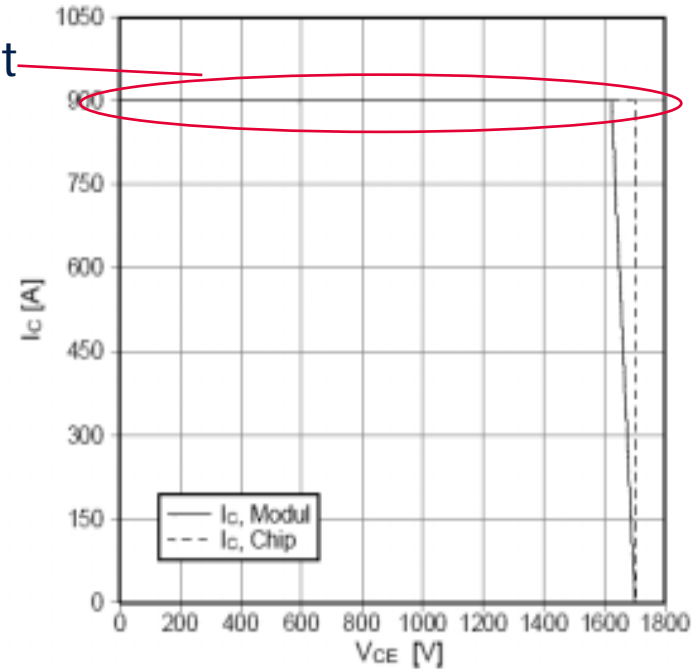
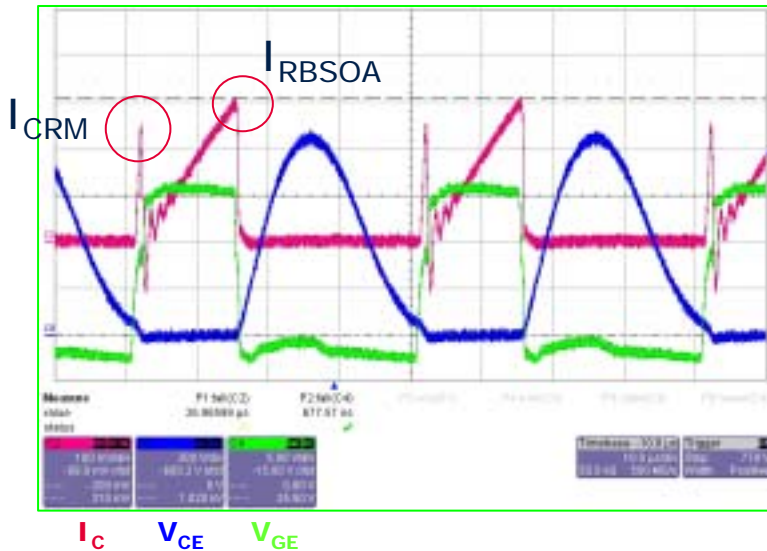
■ Pulse current (ICRM IRBSOA)

| | | | | |
|--|---------------------|-----------|-----|---|
| Periodischer Kollektor Spitzenstrom repetitive peak collector current | $t_p = 1\text{ ms}$ | I_{CRM} | 900 | A |
|--|---------------------|-----------|-----|---|

I_{CRM} is defined as repetitive turn on pulse current, related to IGBT thermal

I_{CRM} may be exceeded during turn on due to reverse recovery.

I_{RBSOA} is defined as maximum turn off current



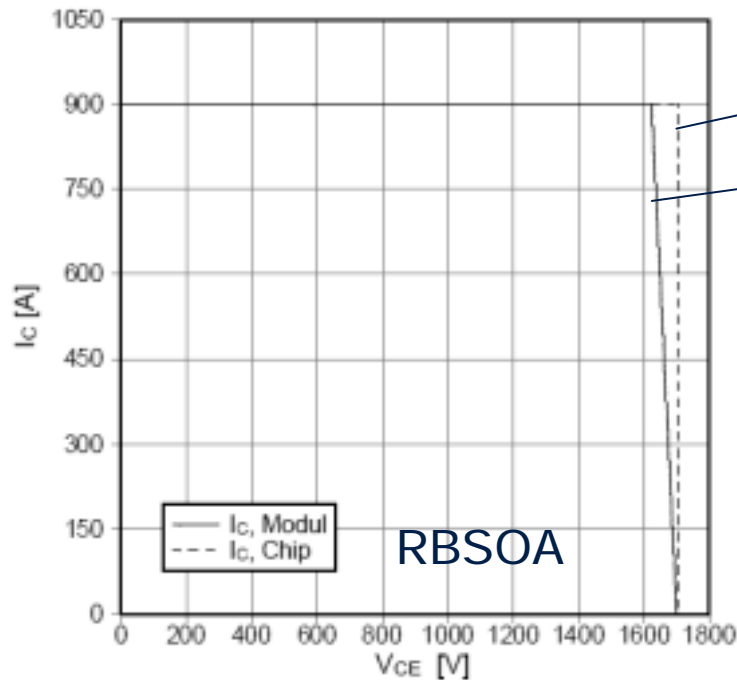
1ms is just test condition, real pulse width is depend on thermal

IGBT Safe Operation – IGBT RBSOA

■ Blocking voltage (V_{CES})

| | | | | |
|--|------------------------|------------------|------|---|
| Kollektor-Emitter-Sperrspannung collector-emitter voltage | T _{vj} = 25°C | V _{CES} | 1700 | V |
|--|------------------------|------------------|------|---|

V_{CES} specified at T_j=25°C. Higher T_j, higher blocking voltage



Chip level

Module level

Due to stray inductance inside module

$$\Delta V = di / dt * L_{\delta}$$

V_{CES} is easiest to be exceeded during turn off, due to external and internal stray inductance

V_{CES} can not be violated at any condition, otherwise IGBT would break though



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IGBT Safe Operation – Diode SOA

■ Blocking voltage (VRRM)

| | | | | |
|---|-------------------------------|------|------|---|
| Periodische Spitzensperrspannung repetitive peak reverse voltage | $T_{vj} = 25^{\circ}\text{C}$ | VRRM | 1700 | V |
|---|-------------------------------|------|------|---|

Similar definition of V_{CES} at $T_j 25^{\circ}\text{C}$

■ Pulse current (ICRM)

| | | | | |
|--|----------------------|------|-----|---|
| Periodischer Spitzenstrom repetitive peak forward current | $t_p = 1 \text{ ms}$ | IFRM | 900 | A |
|--|----------------------|------|-----|---|

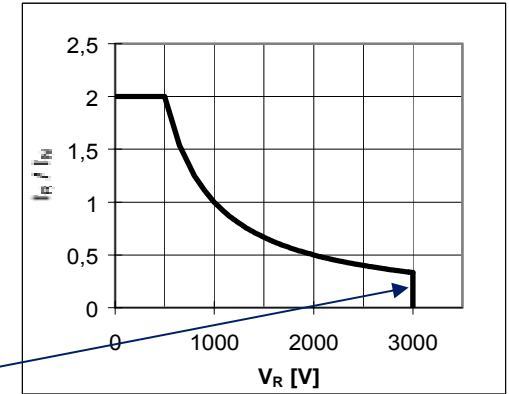
Similar definition of I_{CRM} , two time of I_F .

IGBT Safe Operation – Diode SOA

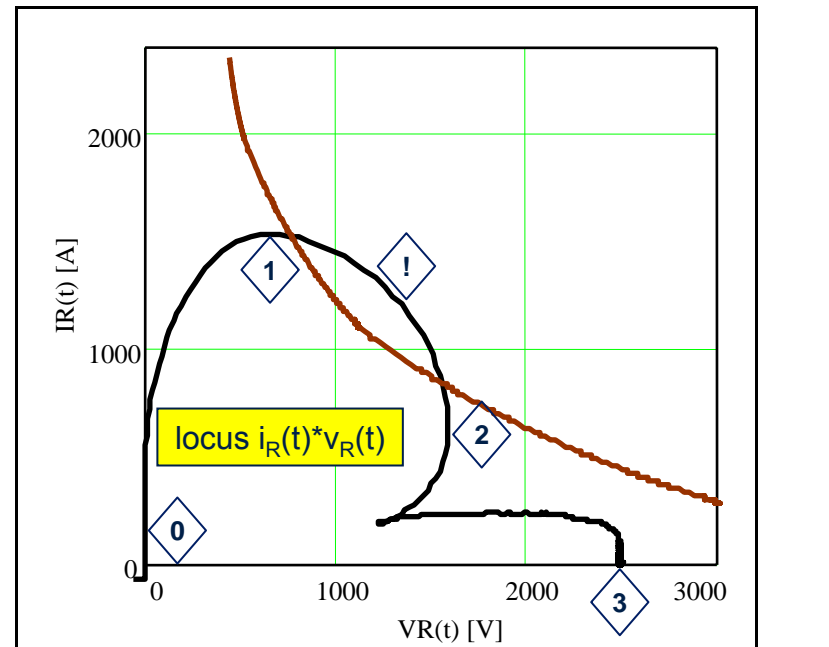
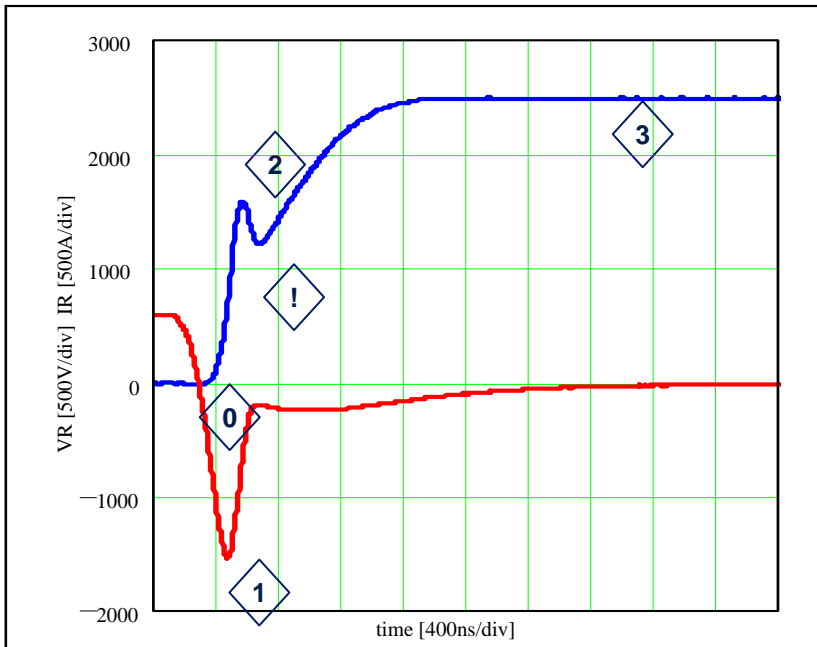
■ Diode SOA

High voltage module specify the SOA of diode. Not only peak current and voltage is limited, peak power also is restricted.

The instantaneous peak power should never exceed the limit for the max. power given in the SOA diagram.



More severe with small current at low temperature due to snap off and oscillation



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 - IGBT RBSOA

 - Diode SOA

 - **Short circuit**

 - Vge limit

 - Others – Not tested by double pulse test

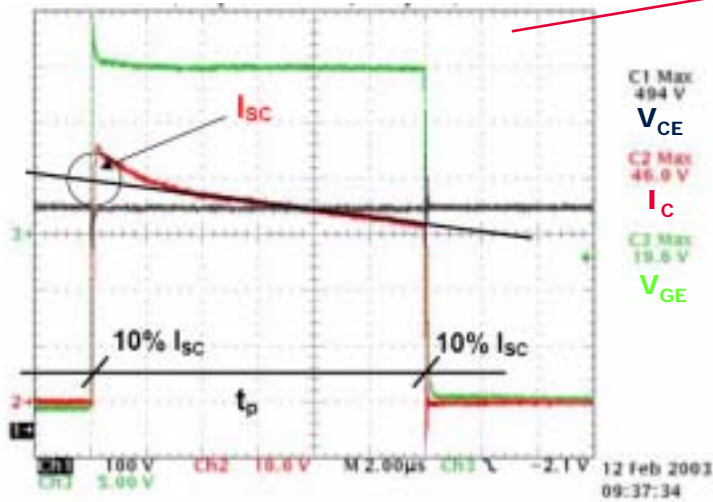
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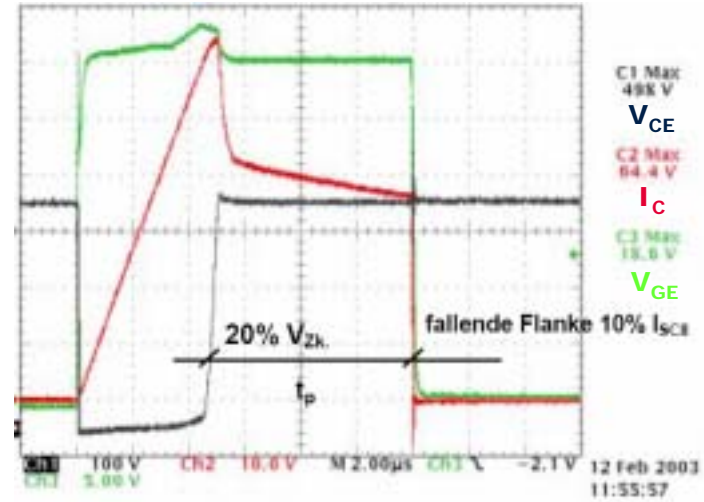
IGBT Safe Operation – short circuit

■ Short circuit current (ISC)

| | | | | | |
|---------------------------------|--|---|----------|------|---|
| Kurzschlussverhalten SC data | $V_{GE} \leq 15 \text{ V}, V_{CC} = 1000 \text{ V}$ $V_{CEmax} = V_{CES} - L_{sCE} \cdot di/dt$ | $t_p \leq 10 \mu\text{s}, T_{vj} = 125^\circ\text{C}$ | I_{sc} | 1800 | A |
|---------------------------------|--|---|----------|------|---|



SC1: Short before Switch On



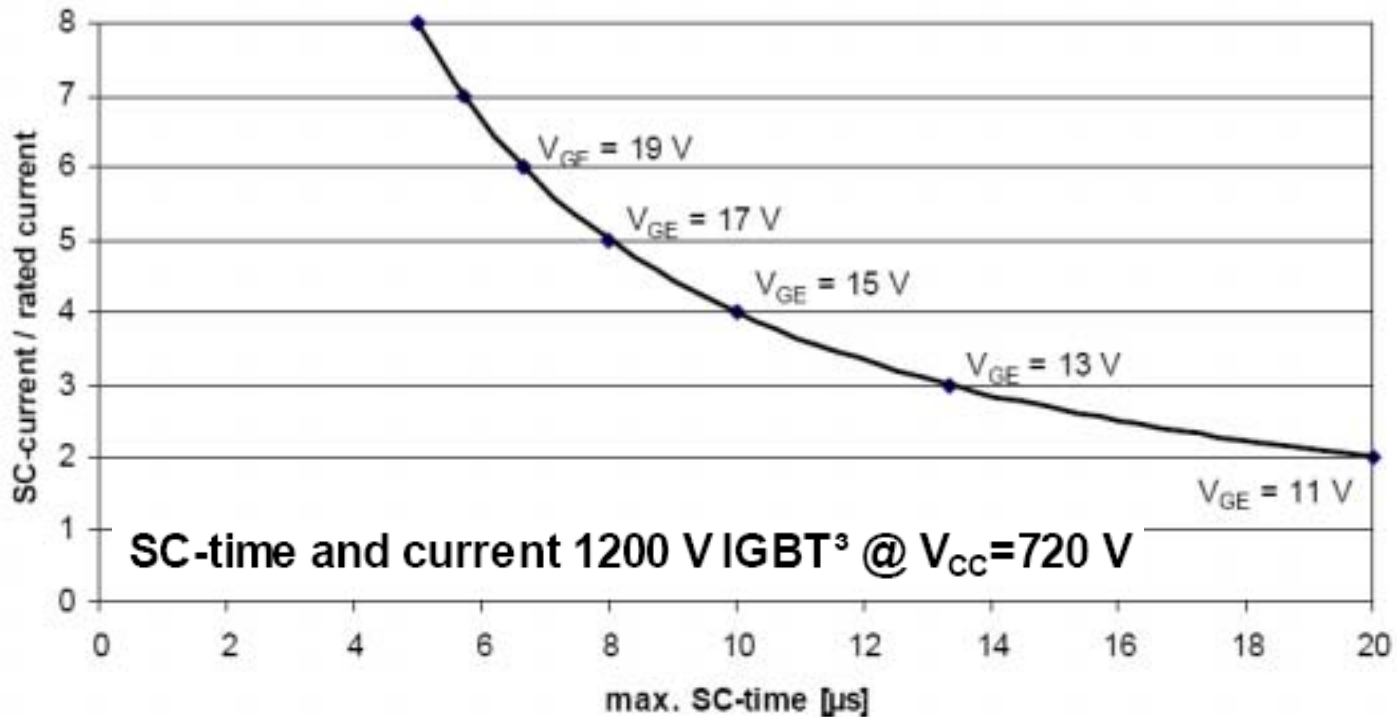
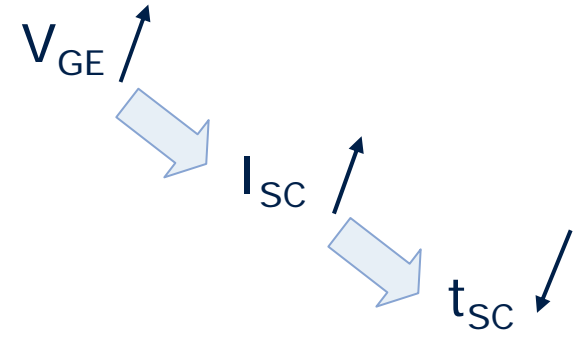
SC2: Short after Switch On

The short circuit current value is a typical value. In applications, SC1 and SC2 can only be safely turned off when desaturated, the short circuit time should not exceed 10us.

IGBT Safe Operation – short circuit

■ Short circuit condition:

- V_{GE}: gate voltage (15V)
- V_{CC}: DC bus voltage
- T_{vj}: short circuit start temperature



It is important to clamp gate voltage during short circuit



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 - **V_{ge} limit**

 - Others – Not tested by double pulse test

- What can be done with double pulse test

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IGBT Safe Operation – V_{ge} limit

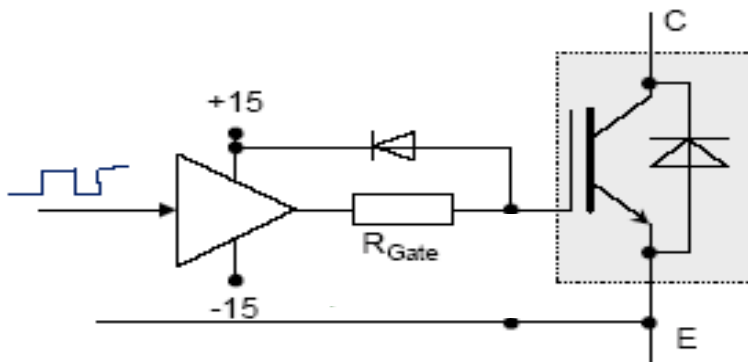
Gate-emitter voltage (V_{ge})

| | | | | |
|---|--|------------------|-------|---|
| Gate-Emitter-Spitzenspannung gate-emitter peak voltage | | V _{GES} | +/-20 | V |
|---|--|------------------|-------|---|

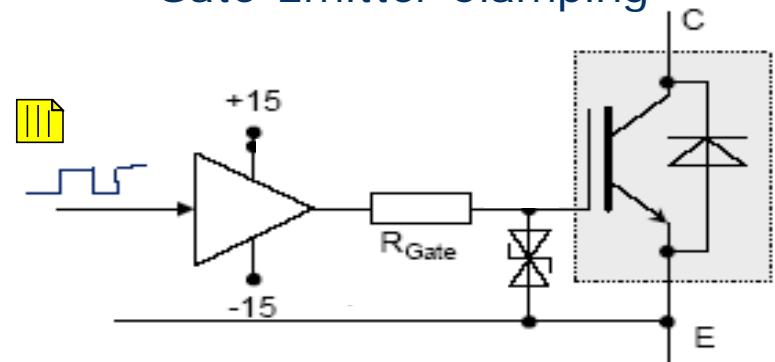
Gate Clamping:

- Limitation of increase of gate voltage due to positive feedback over C_{GC}
- An issue with long durations regarding gate oxide break down
- Limitation of short circuit currents

Method 1
Gate-Supply Clamping



Method 2
Gate-Emitter Clamping



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- **Safe operation of IGBT**

- IGBT RBSOA
- Diode SOA
- Short circuit
- Vge limit
- Others – Not tested by double pulse test**

- What can be done with double pulse test

- Impact of R_g , C_{ge} on IGBT switching

IGBT Safe Operation – Others, not tested by two-pulse test

| | | | | | |
|--|---|---------------|-----|-----|----|
| Höchstzulässige Sperrschichttemperatur maximum junction temperature | Wechselrichter, Brems-Chopper / Inverter, Brake-Chopper | $T_{vj\ max}$ | | 175 | °C |
| Temperatur im Schaltbetrieb temperature under switching conditions | Wechselrichter, Brems-Chopper / Inverter, Brake-Chopper | $T_{vj\ op}$ | -40 | 150 | °C |

■ Maximum junction temperature

- IGBT & Diode loss estimation
- Thermal impedance

■ Reliability

- DC stability

| | | | | |
|---|--|-------------|------|---|
| Kollektor-Emitter-Gleichsperrspannung DC stability | $T_{vj} = 25^{\circ}\text{C}, 100\ \text{fit}$ | $V_{CE\ D}$ | 2100 | V |
|---|--|-------------|------|---|

- Thermal cycling & power cycling

■ Min switching time

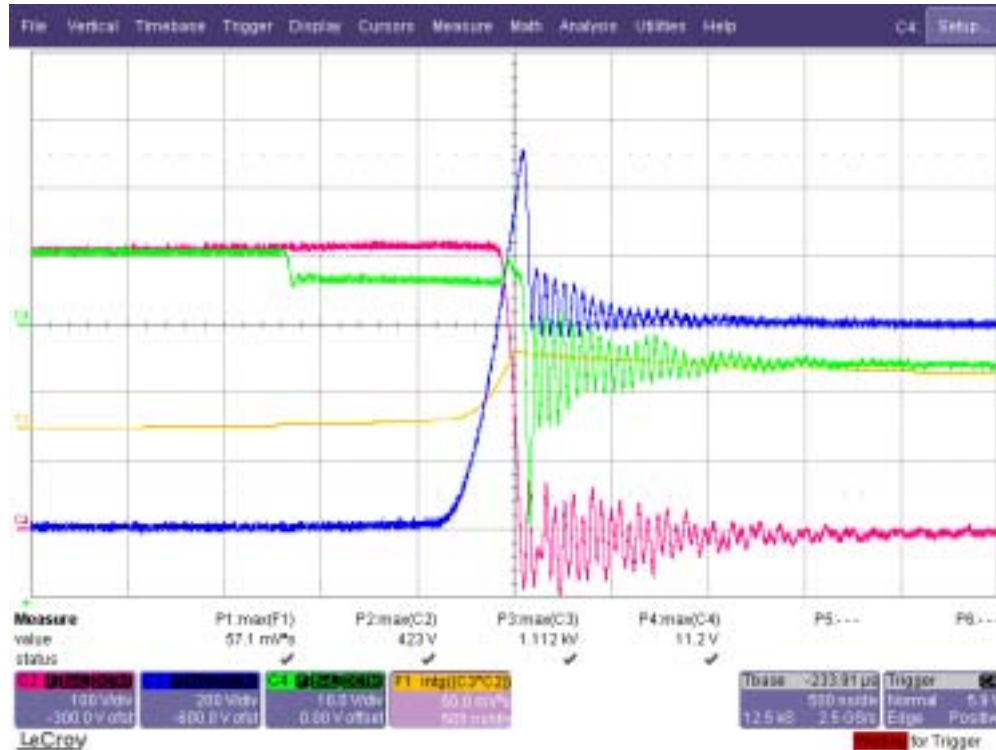
| | | | | |
|---|--|----------------|------|----|
| Mindesteinschaltdauer minimum turn-on time | | $t_{Fon\ min}$ | 10,0 | µs |
|---|--|----------------|------|----|

■

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 - Measurement of loss, switching time, stray inductance
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 - Optimize Driver design
- Impact of R_g , C_{ge} on IGBT switching

Double pulse test – waveforms



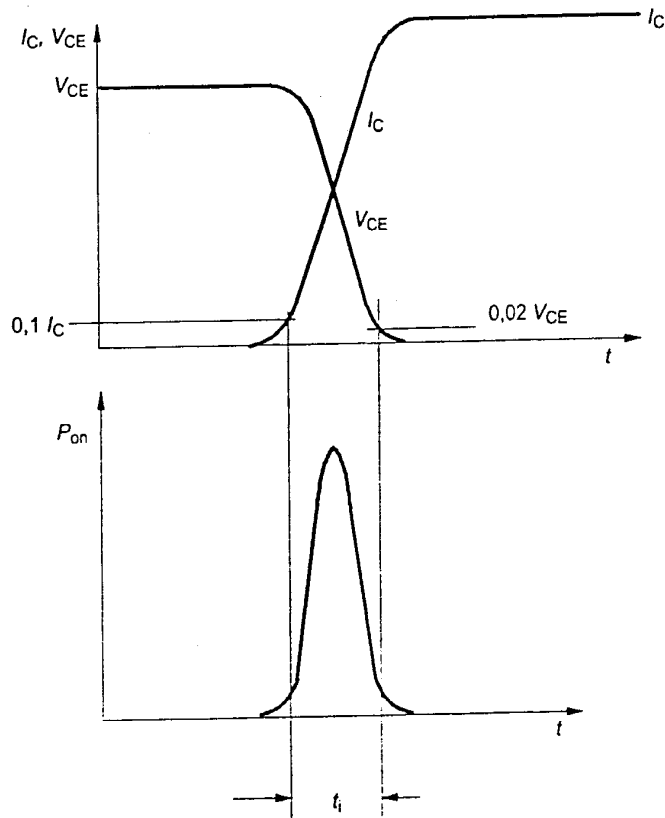
Both too small R_g and too large R_g can lead to oscillations.

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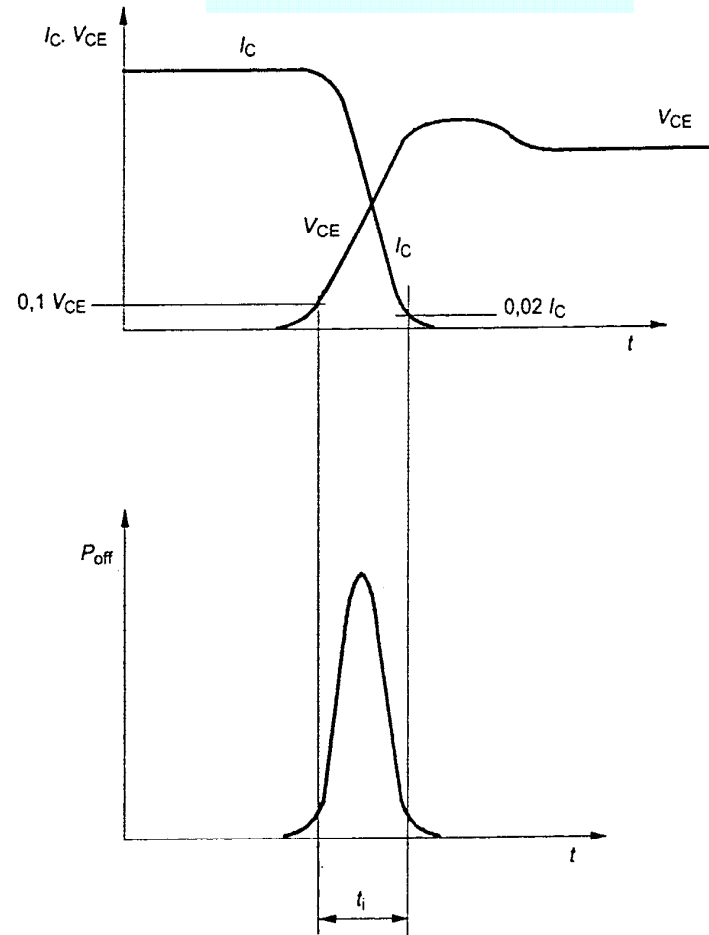
Definition of Eon and Eoff_IEC60747-9

Eon definition



10% I_c to 2% V_{ce}

Eoff definition



10% V_{ce} to 2% I_c

Definition of Eon and Eoff

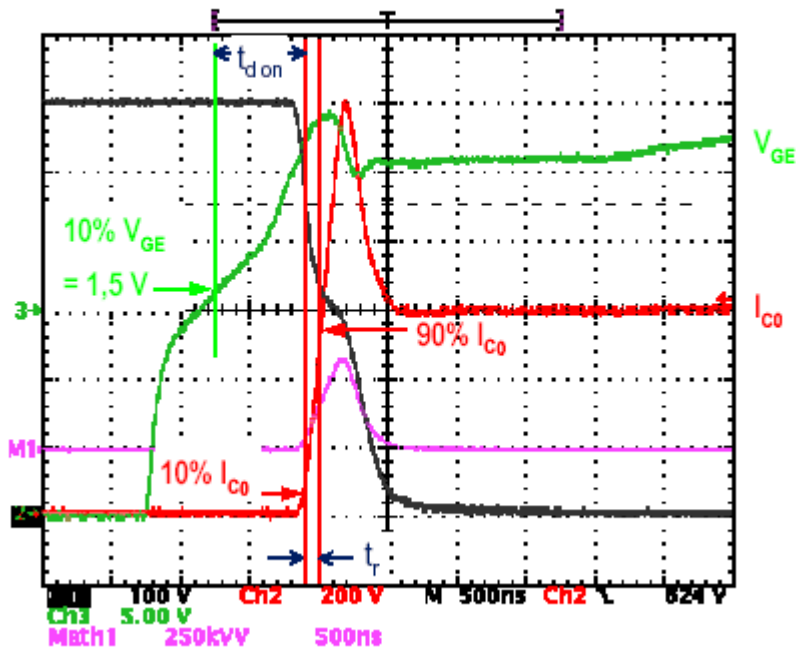
| | Eon | Eoff | Notes |
|--------------------------------------|----------------------|--------------------|-----------------------------------|
| IEC60747-9 | 10%Ic-2%Vce | 10%Vce-2%Ic | |
| IEC60747-9 Ed2 2005-06-10 | 10% Vge-2%Vce | 90%Vge-2%Ic | |
| Infineon Module | 10%Ic-2%Vce | 10%Vce-2%Ic | |
| Infineon Discrete | 10%Vge-3%Vce | 90%Vge-1%Ic | Incl. 0-10% Ic at full Vce |
| IR Discrete | 10%Ic-5%Vce | 10%Vce-5%Ic | |
| Fairchild Discrete | 10%Vge-0%Vce | 90%Vge-0%Ic | |
| Tyco Module | 10%Vge-3%Vce | 90%Vge-1%Ic | |

- IGBT module datasheets give two/three *Eon* & *Eoff* values at $T_j = 25^\circ\text{C}$ & $T_j = 125^\circ\text{C}/150^\circ\text{C}$, respectively, all for $I_{C,nom}$ & around $\frac{1}{2} V_{ces}$, $V_{GE} = \pm 15\text{V}$.
- *Eon* & *Eoff*, especially *Eon* increases with increase of R_G
- *Eon* & *Eoff* increase with the rising of T_j

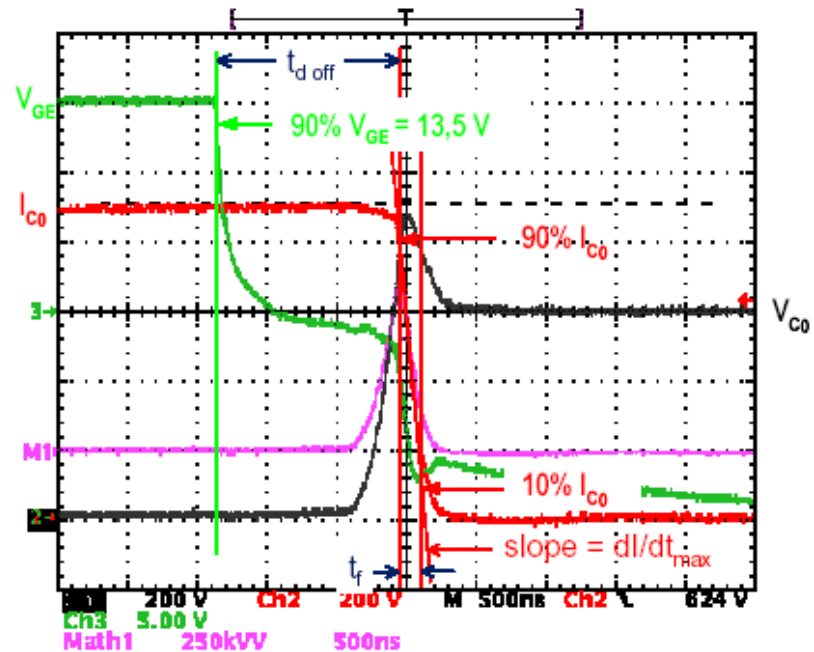
Definition of t_{don} , t_r , t_{doff} and t_f

| t_{don} | t_r | t_{doff} | t_f |
|-------------------------|----------------------|-------------------------|----------------------|
| 10% V_{ge} -10% I_c | 10% I_c -90% I_c | 90% V_{ge} -90% I_c | 90% I_c -10% I_c |

IGBT Turn on

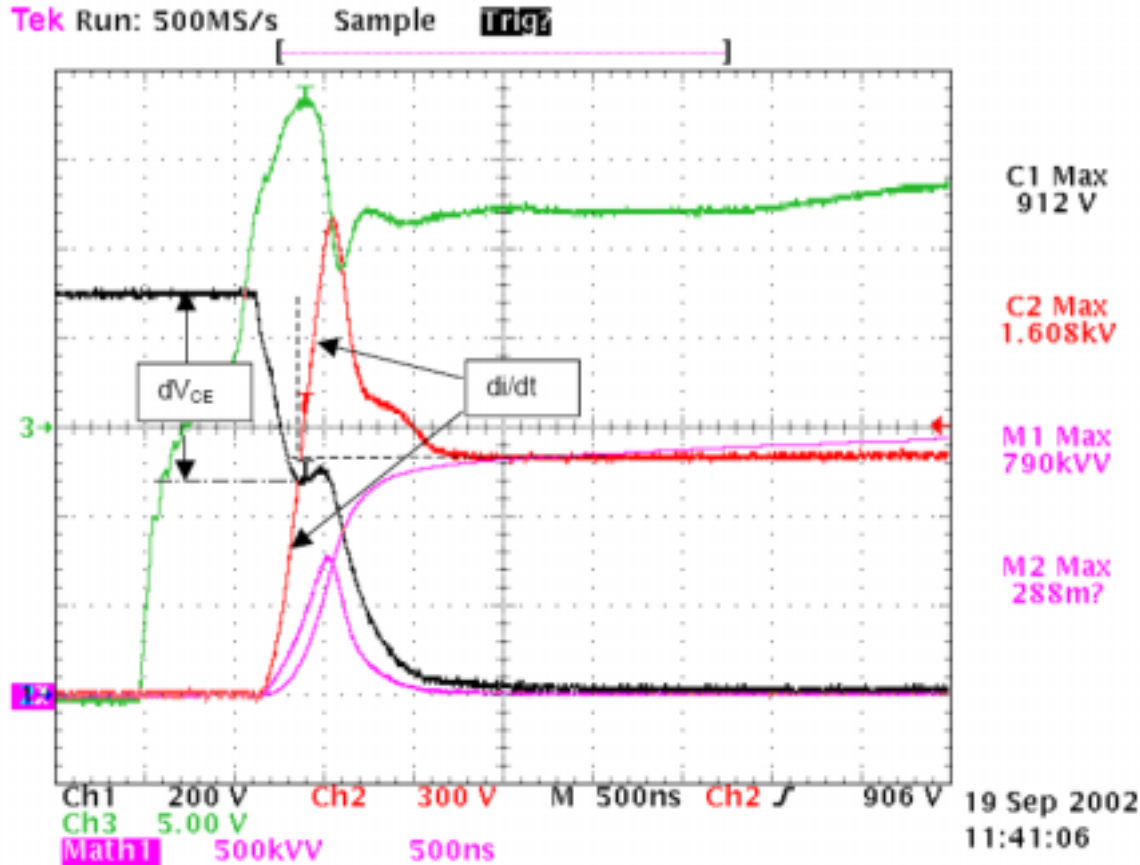


IGBT Turn off



Stray Inductance System Stray Inductance

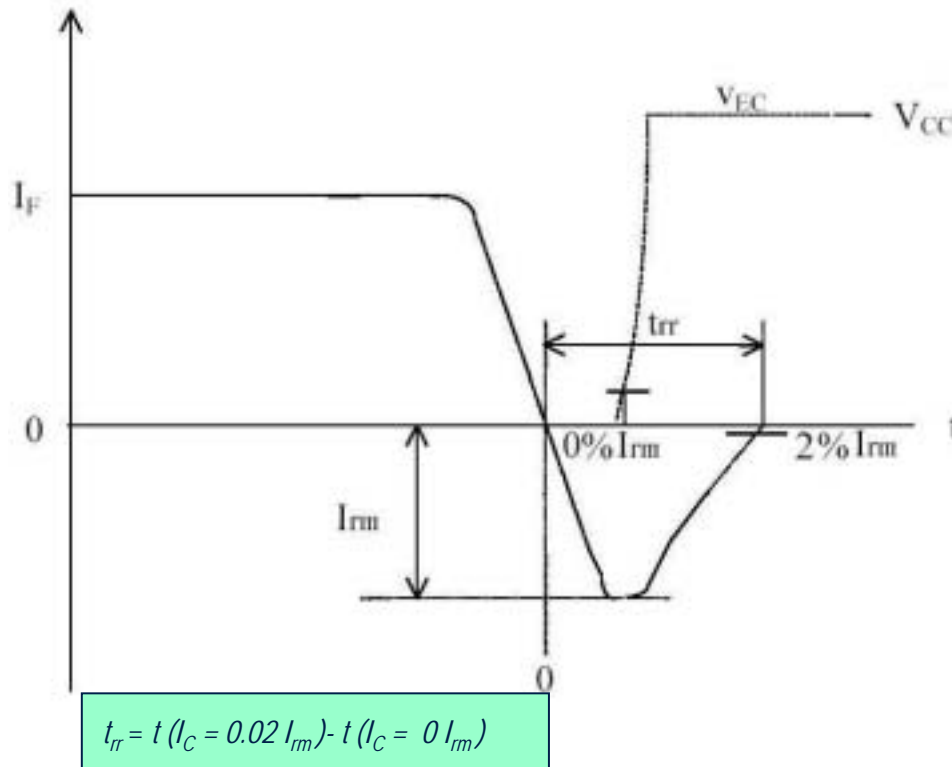
■ Sample Wave Form



$$L_{\sigma} = dV_{CE} \cdot \frac{dt}{di}$$

Note: connect to power terminal when measuring external busbar stray inductance.

Diode switching measurement



$$E_{rec} = - \int_{V_{EC}=0.1 \cdot V_{CC}}^{I_F=0.02 \cdot I_{rm}} V_{EC}(t) \cdot I_F(t) dt$$

$$Q_{rr} = - \int_{I_F=0.02 \cdot I_{rm}}^{I_F=0 \cdot I_{rm}} I_F(t) dt$$

- IGBT module datasheets give two/three Erec values under Tj = 25°C & Tj = 125°C/150°C, all under IF,nom & ½ VCES, VGE = -15V
- Erec decreases with the increase of RG
- Erec increases with the rising of Tj

Dead Time Estimation

$$t_{DT} = \underbrace{[(t_{doff}(\max) + t_f(\max) - t_{don}(\min))]}_{\text{IGBT}} + \underbrace{(t_{PHL\max} - t_{PLH\min})}_{\text{Driver}} \underbrace{]}_{\text{Margin}} * 1.2$$

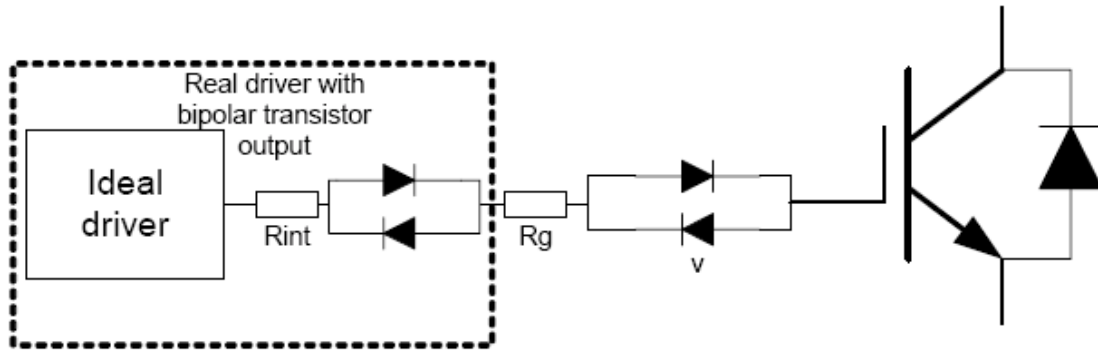
- T_j , I_c , and R_g has small impact on T_{don}
 - Min T_{don} typically at small I_c , small R_g

- T_j , I_c , R_g , and V_{ge} has significant impact on T_{doff}
 - Max T_{doff} at: small current I_c
 - High temperature T_j
 - Variation of V_{th} should be considered
 - Big R_g
 - Unipolar V_{ge}

- T_f can be neglected because it is typically very small

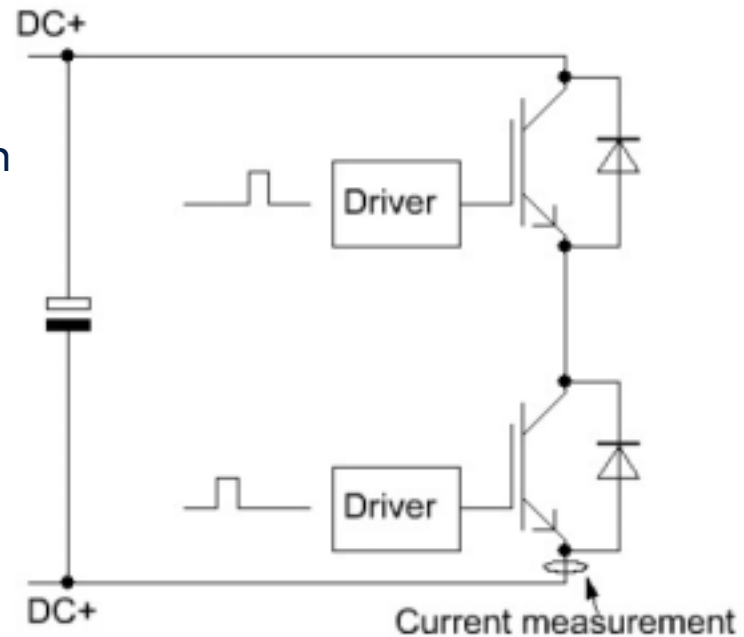
- Propagation delay time of driver must be considered to calculate t_{DT}

Dead Time Estimation



Block diagram of test to simulate variation of v_{th} and driver with bipolar output

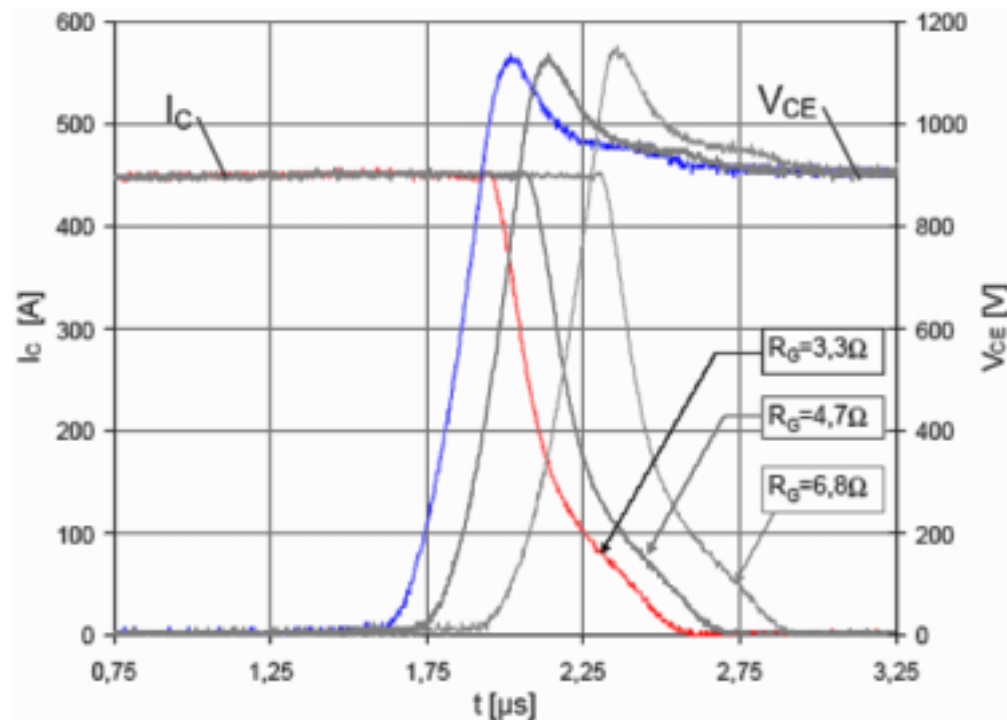
Verification of Dead Time, both high temperature and low temperature should be tested



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- **What can be done with double pulse test**
 - Check switching waveforms – oscillation?
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 - **Verify IGBT RBSOA, diode SOA**
 - Verify short-circuit protection
 - IGBT Paralleling
 - Optimize Driver design
- Impact of R_g , C_{ge} on IGBT switching

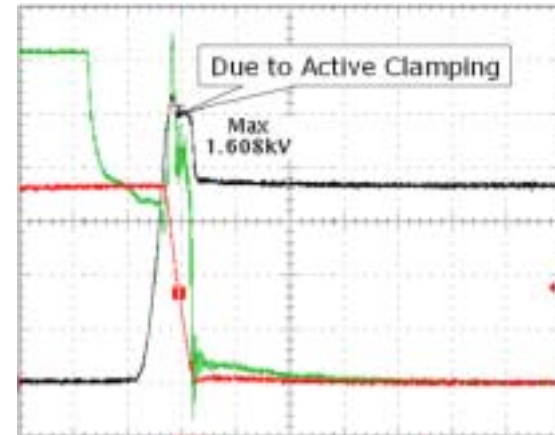
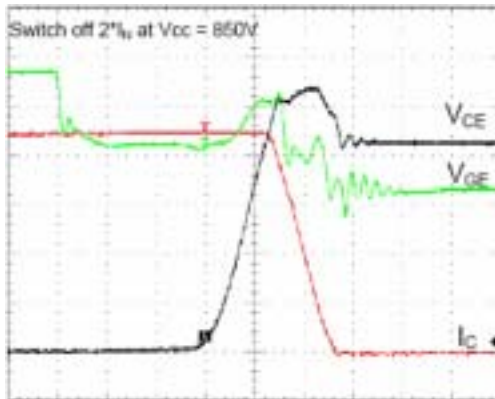
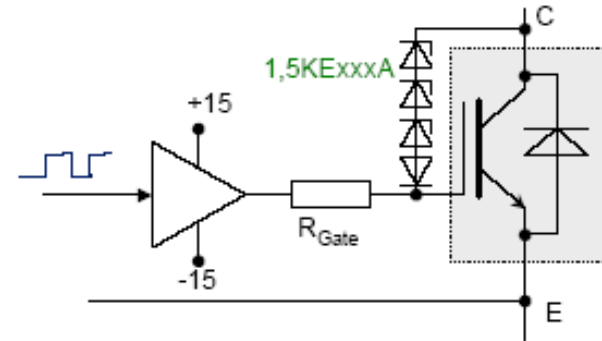
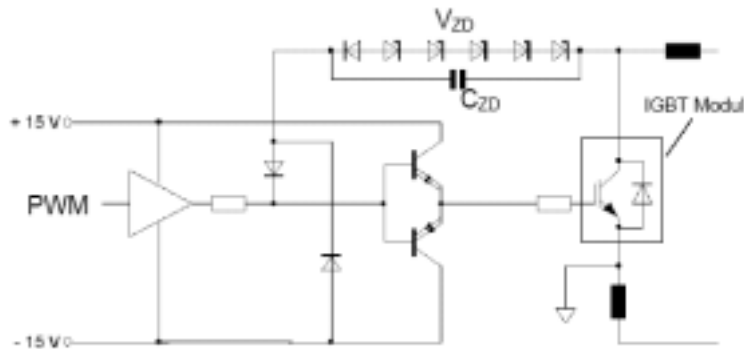
Verify IGBT RBSOA



- For 3rd and 4th IGBT, R_{goff} has little impact on E_{off} , dv/dt , and di/dt
- di/dt is only controllable if the gate voltage doesn't drop below the Miller Plateau level before I_C starts to decrease
- dv/dt and di/dt are controllable by the gate resistor when R_{goff} is very large
- A larger resistor will result in a smaller dv/dt and di/dt

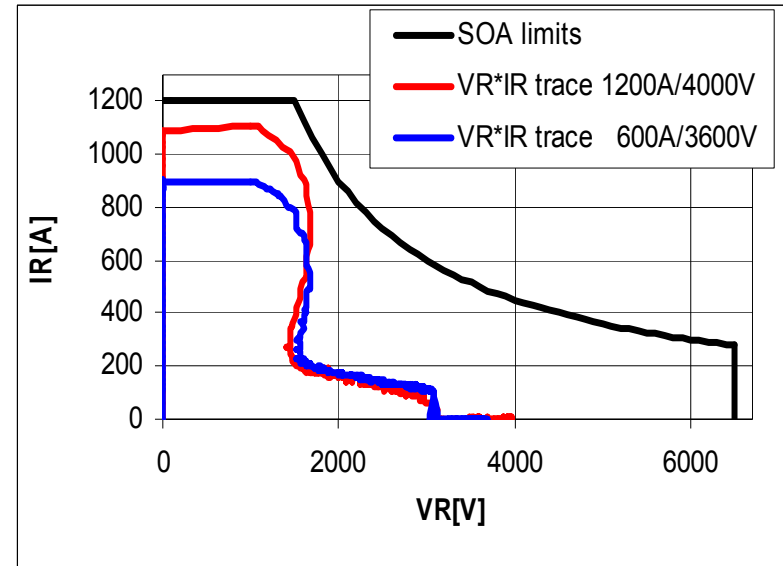
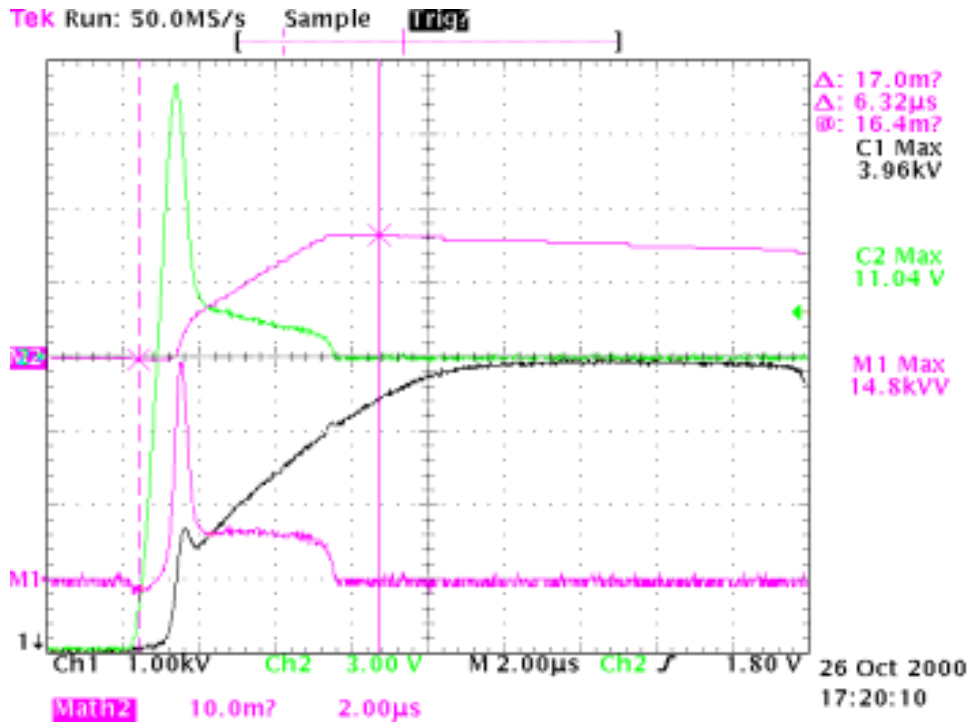
Verify IGBT RBSOA

- In no case shall over voltages exceed the maximum breakdown voltage of the IGBT
- To control the IGBT it is necessary that the gate voltage hasn't dropped below the Miller Plateau level



Verify Diode SOA

example: 600A / 6500V module

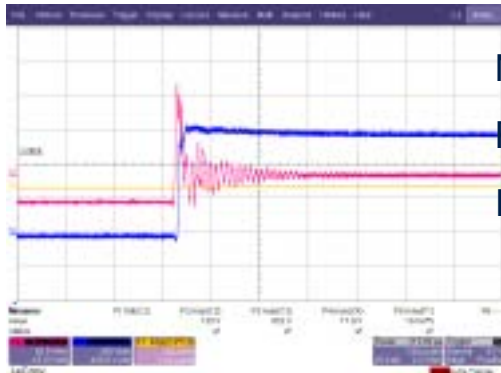


measured: $i_R(t)$ and $v_R(t)$

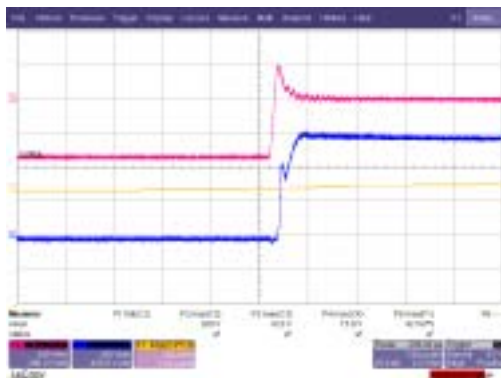
calculated: $p_{\text{peak}}(t) = i_R(t) \cdot v_R(t)$

Verify Diode SOA

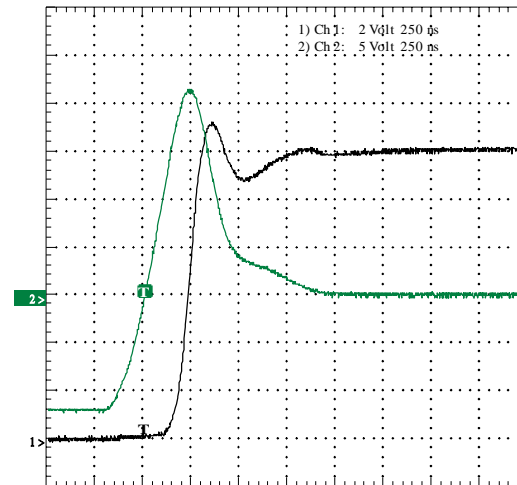
- Erec, Irr, dv/dt, and di/dt will be decreased with increasing Rg
- Erec, Irr, dv/dt, and di/dt will be decreased with increasing Cg
- Higher Tj lead to decreased dv/dt and di/dt
- Diode tends to snap off with small current



No Cge, t/div=1us
 Ron=Roff=1.8 Ω
 Ic= 1/10 Inom

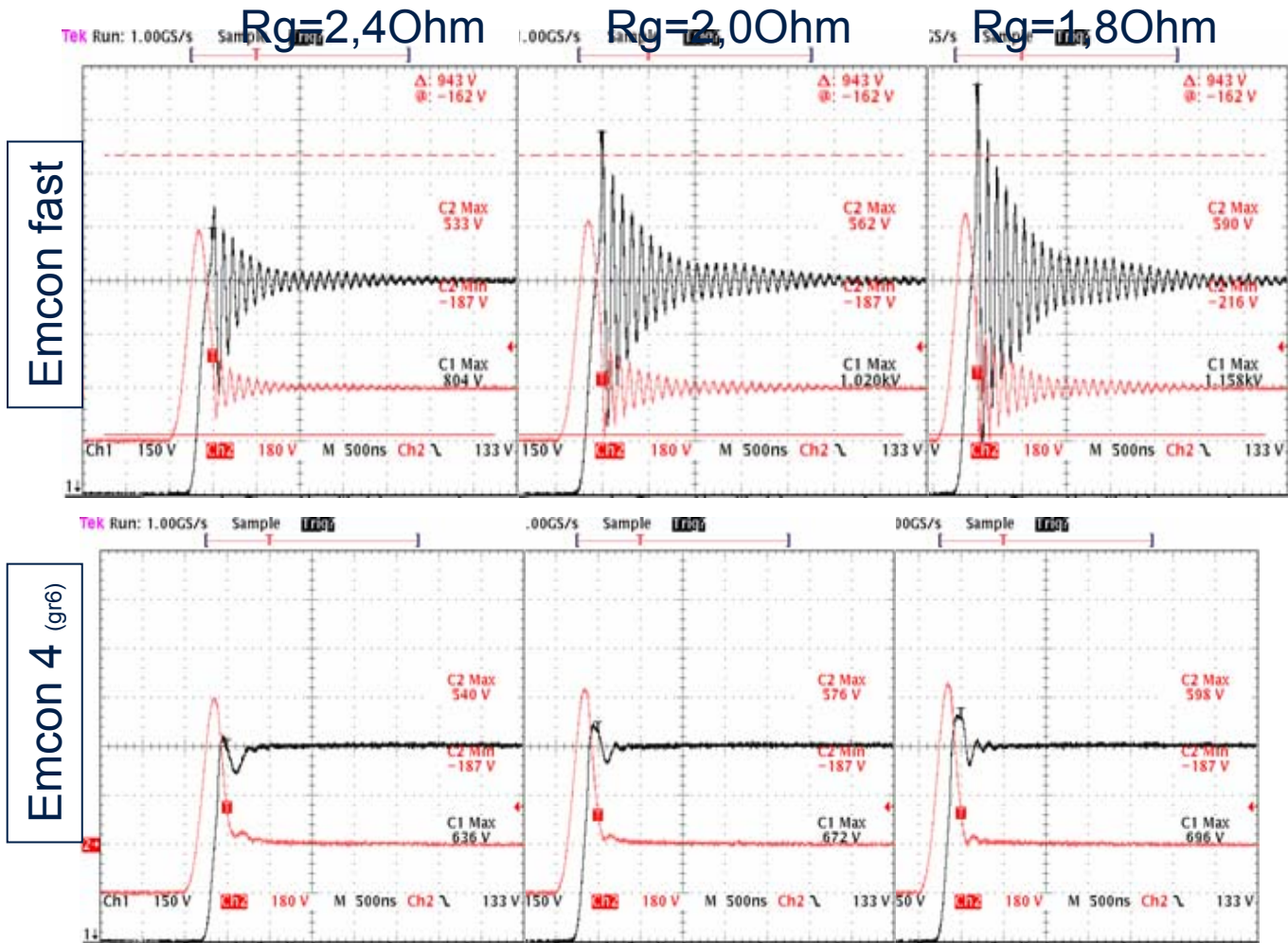


No Cge, t/div=1us
 Ron=Roff=1.8 Ω
 Ic= Inom



Soft recovery behavior

Verify Diode SOA – diode snap off

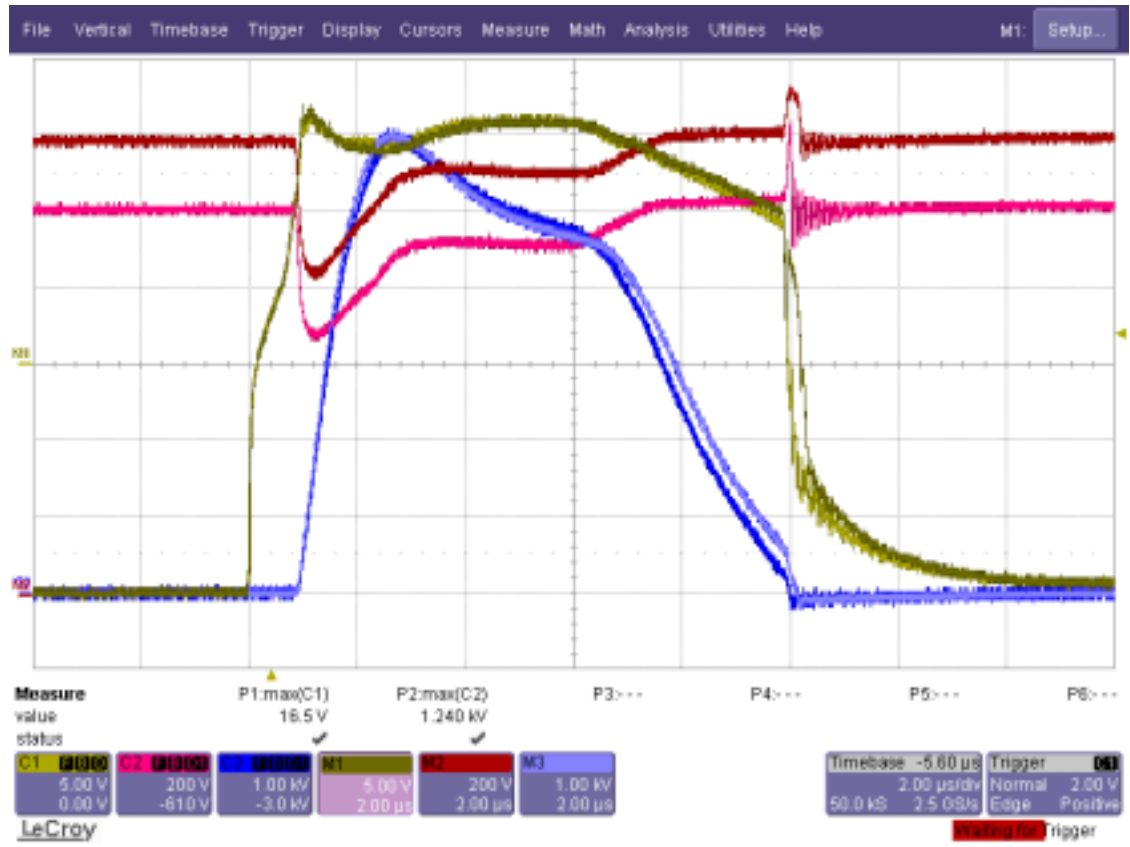


Can lead to high voltage with small current at low temperature

IGBT Double Pulse Test

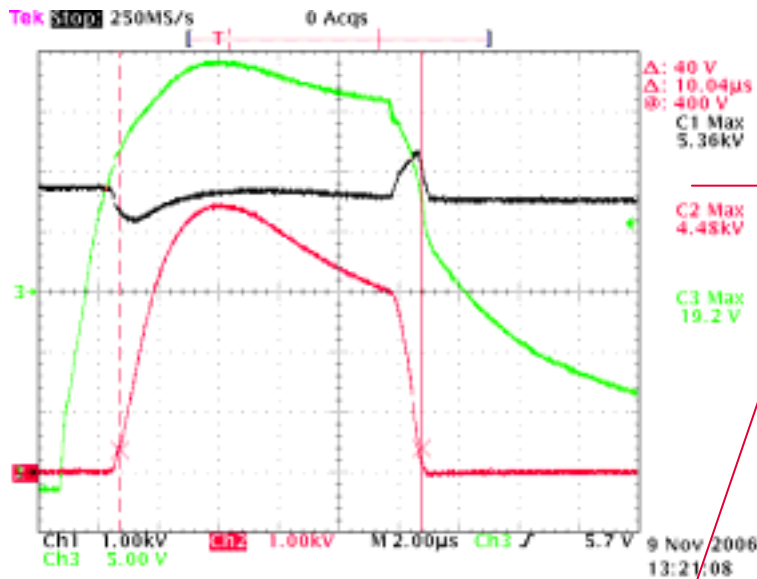
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 - **Verify short-circuit protection**
 - IGBT Paralleling
 - Optimize Driver design
- Impact of R_g , C_{ge} on IGBT switching

IGBT short circuit protection – turn off voltage overshoot

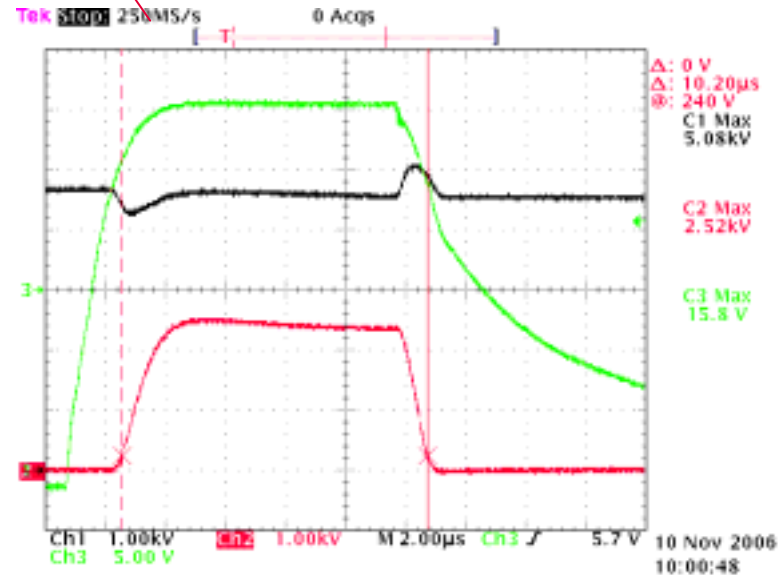
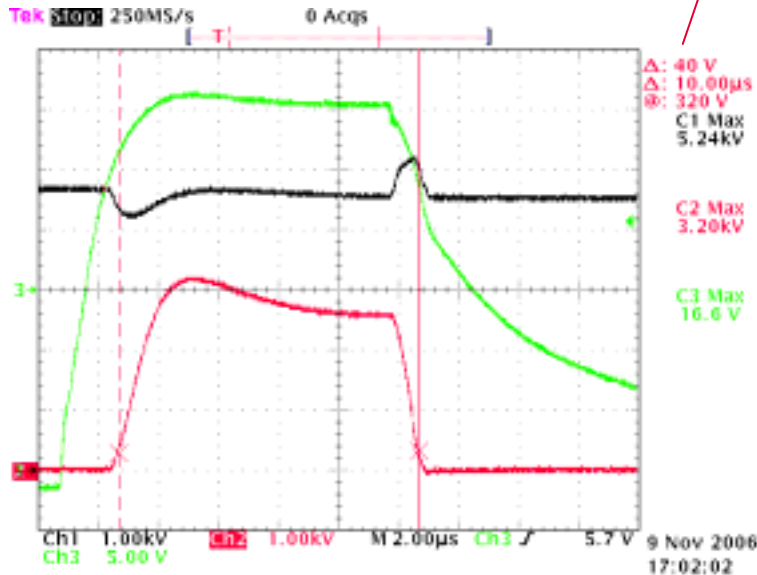


Soft turnoff SC1 – reduce voltage overshoot

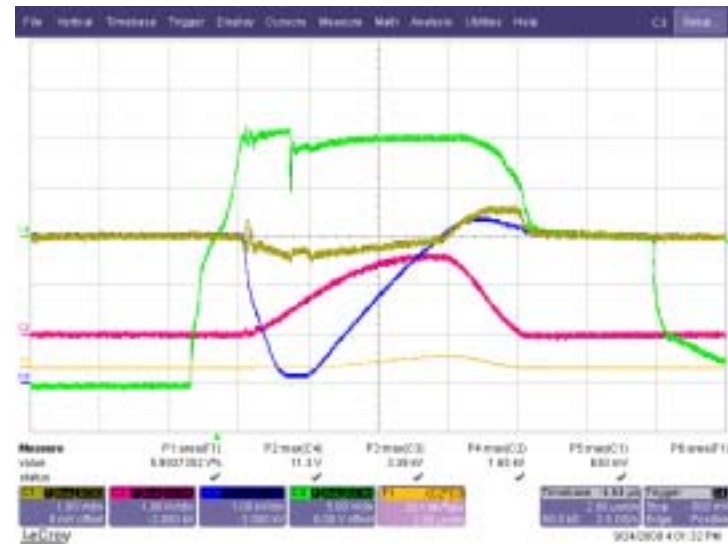
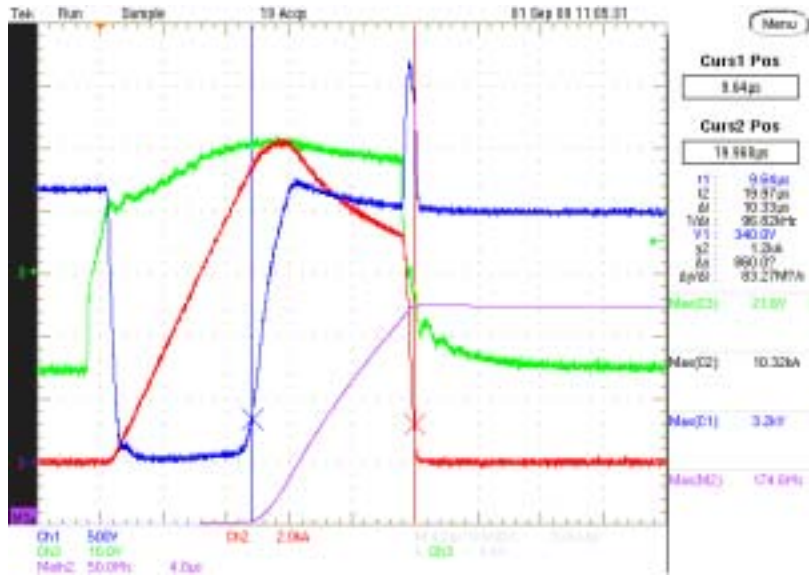
IGBT short circuit protection – gate clamping



- Without clamping
- With TVS clamping between GE
- With clamping to Vcc



IGBT short circuit protection – short circuit II

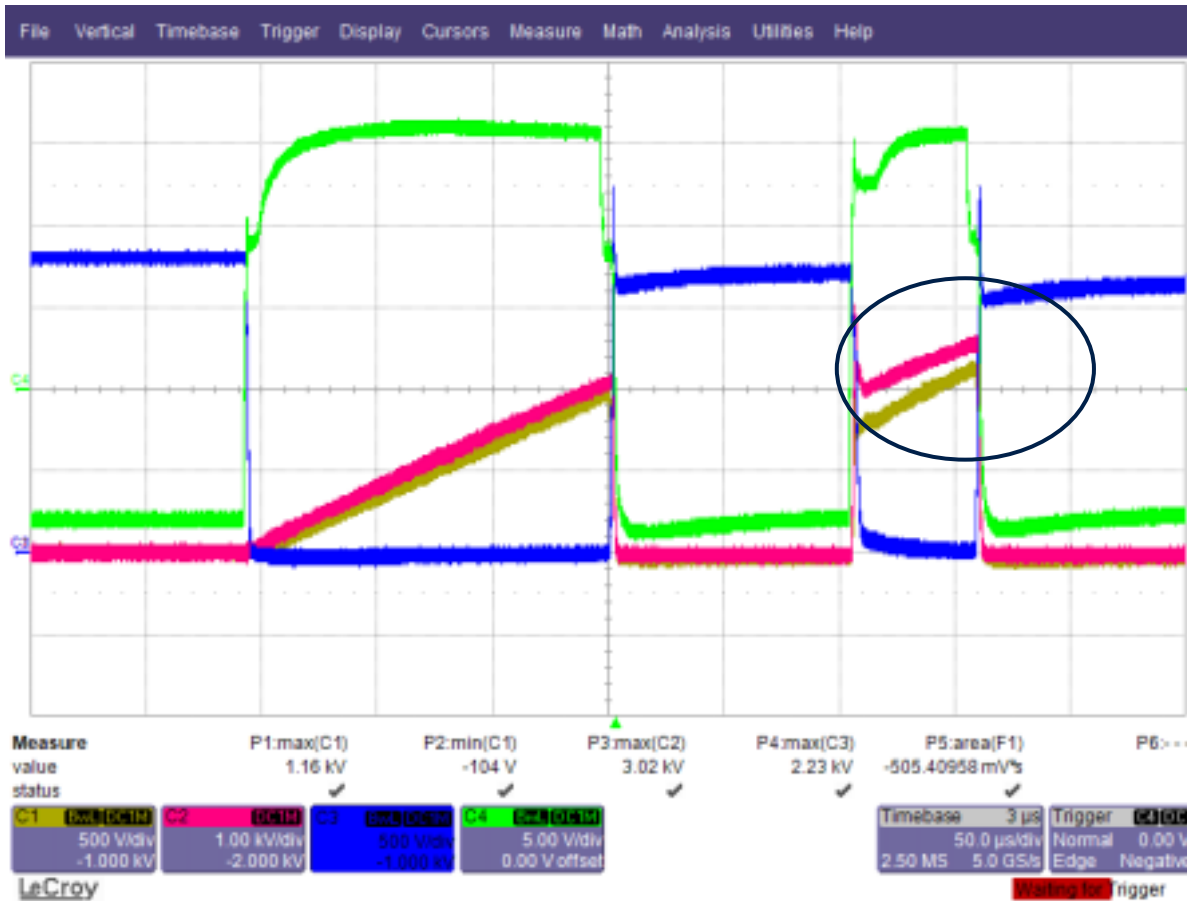


Turn off SC-II only when desaturated.

IGBT Double Pulse Test

- Basic principle of double pulse test
- Safe operation of IGBT
- **What can be done with double pulse test**
 - Check switching waveforms – oscillation?
 - Measurement of loss, switching time, stray inductance
 - Verify IGBT RBSOA, diode SOA
 - Verify short-circuit protection
 - **IGBT Paralleling**
 - Optimize Driver design
- Impact of R_g , C_{ge} on IGBT switching

IGBT parallel



C1: the left IGBT

C2: I_{DC} totally

C3: V_{CE}

C4: V_{GE}

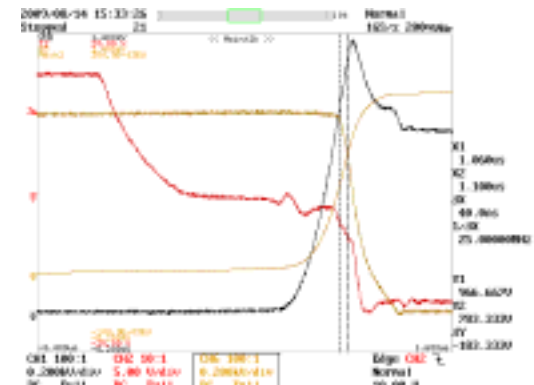
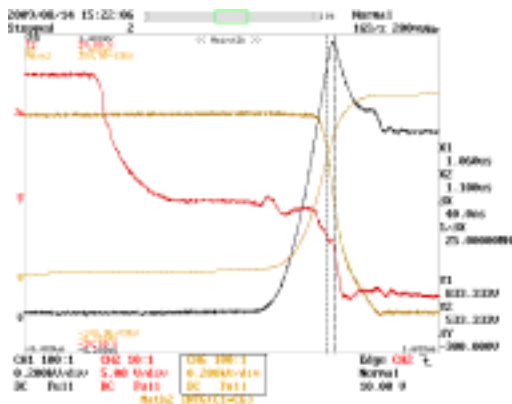
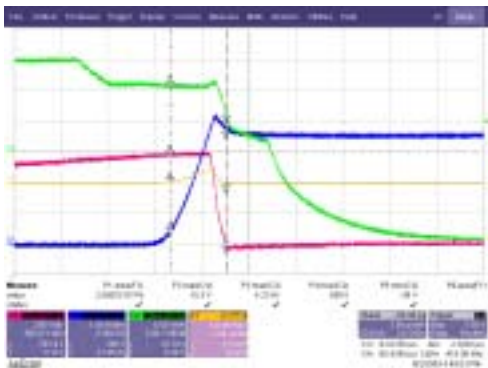
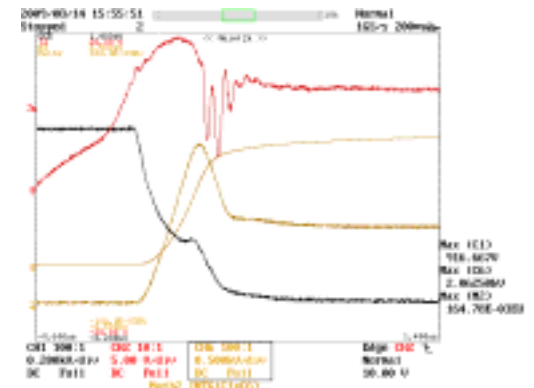
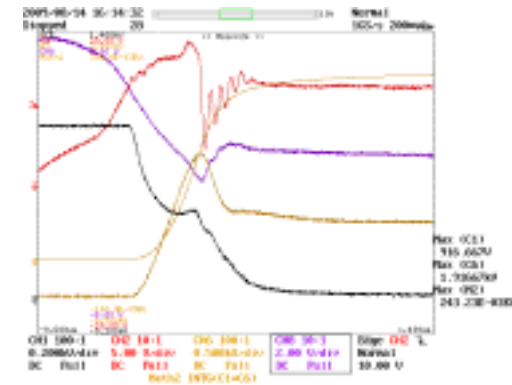
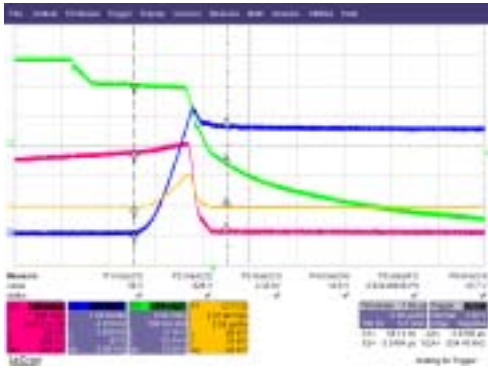


The difference between paralleled IGBT becomes smaller and smaller

IGBT Double Pulse Test

- Basic principle of double pulse test
- Safe operation of IGBT
- **What can be done with double pulse test**
 - Check switching waveforms – oscillation?
 - Measurement of loss, switching time, stray inductance
 - Verify IGBT RBSOA, diode SOA
 - Verify short-circuit protection
 - IGBT Paralleling
 - **Optimize Driver design**
- Impact of R_g , C_{ge} on IGBT switching

Optimize Driver Design



Driver comparison

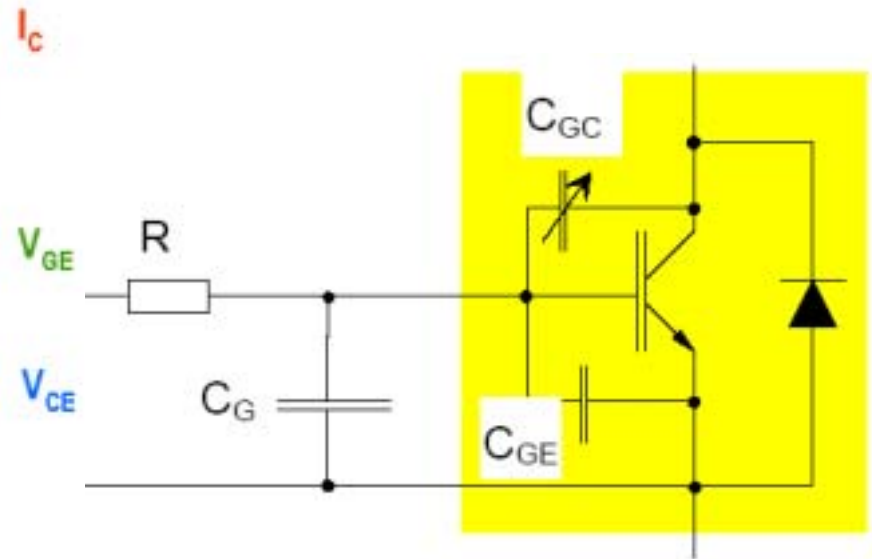
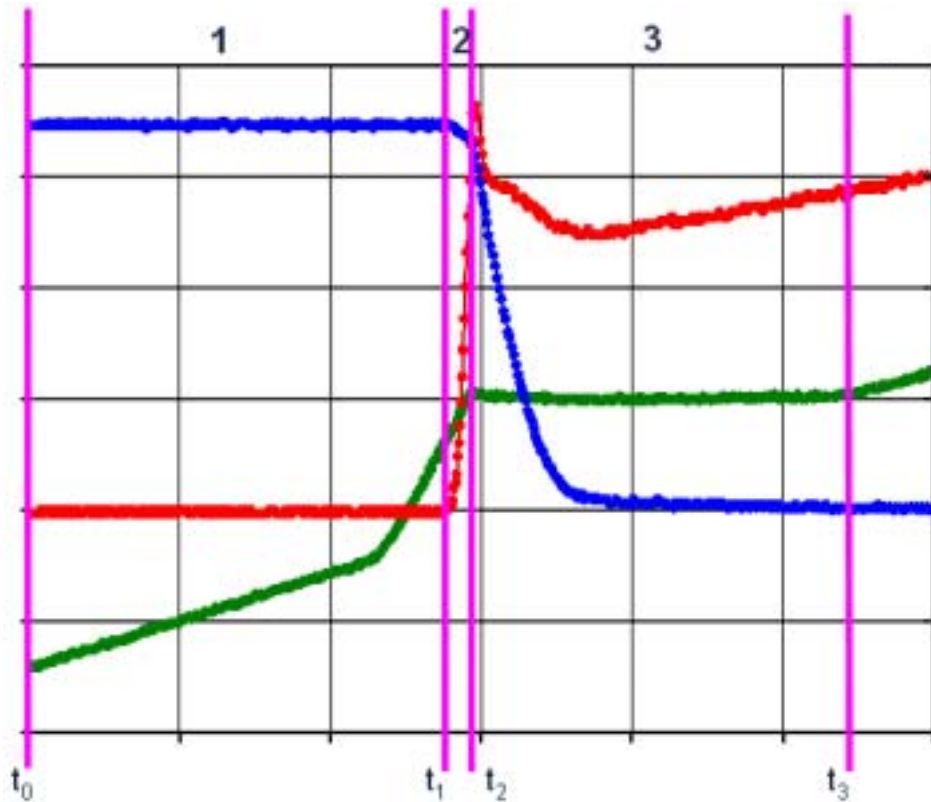
6cm Gate cable

50cm Gate cable

IGBT Double Pulse Test

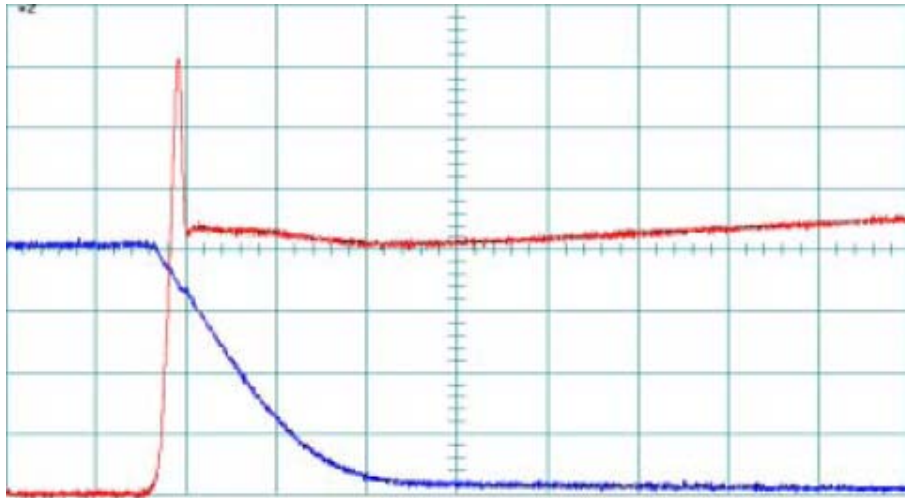
- Basic principle of double pulse test
- Safe operation of IGBT
- What can be done with double pulse test
- **Impact of R_g , C_{ge} on IGBT switching**

Control turn on di/dt (diode reverse recover) and dv/dt (Eon) independently

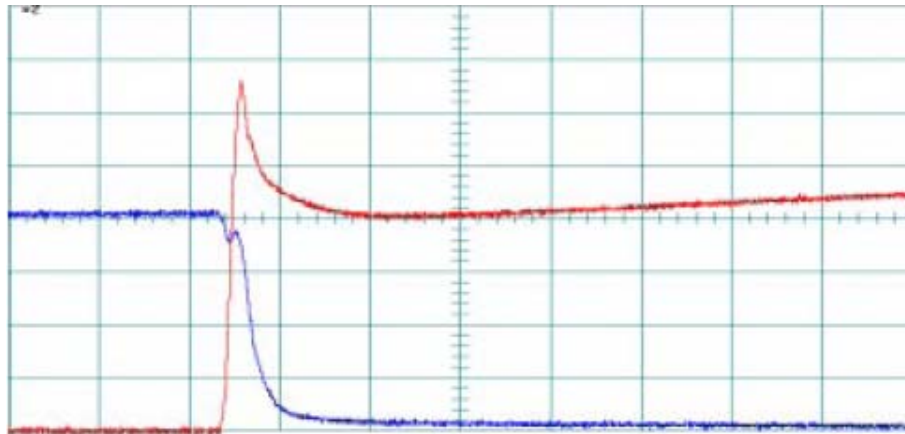


| Range | Determined by | Condition | Influenced by | Influence on |
|-------|-------------------------------|--------------------------|---------------------------------------|--------------|
| 1 | $V_{GE} < V_{GEth}$ | $C_{iss} = \text{const}$ | $R_G, C_{GE} \parallel C_G$ | t_{don} |
| 2 | $V_{GEth} < V_{GE} < V_{GEM}$ | $C_{iss} = \text{const}$ | $R_G, C_{GE} \parallel C_G$ | di/dt |
| 3 | $V_{GE} = V_{GEM}$ | $V_{GE} = \text{const}$ | R_G, C_{GC} ($C_{GC} \gg C_G$) | dv/dt |

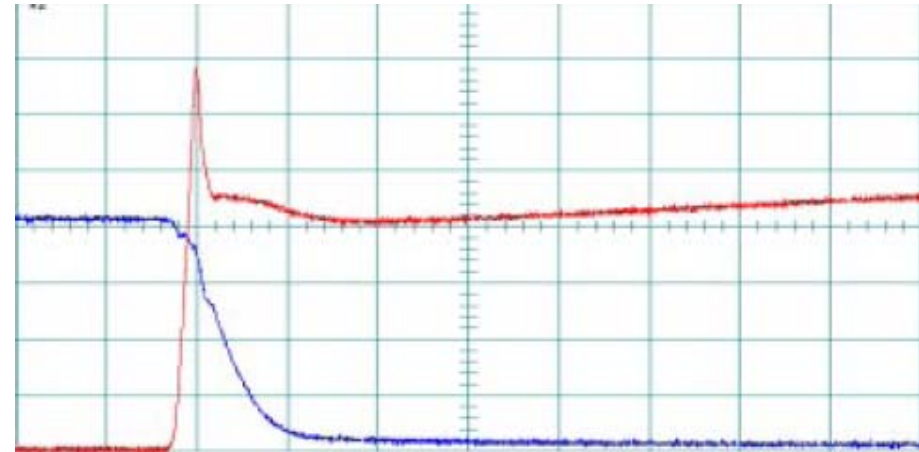
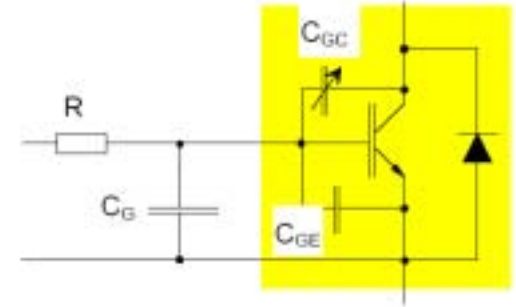
Control turn on di/dt (diode reverse recover) and dv/dt (Eon) independently



$R_G=8,2\Omega$, $C_G=0$, $I_C/dt= 5\text{kA}/\mu\text{s}$, $dV_{CE}/dt=0,6\text{kV}/\mu\text{s}$, $E_{on}=6,4\text{J}$



$R_G=1,0\Omega$, $C_G=330\text{nF}$, $I_C/dt= 5,1\text{kA}/\mu\text{s}$, $dV_{CE}/dt=2,8\text{kV}/\mu\text{s}$, $E_{on}=2,8\text{J}$



$R_G=3,3\Omega$, $C_G=100\text{nF}$, $I_C/dt= 4,5\text{kA}/\mu\text{s}$, $dV_{CE}/dt=1\text{kV}/\mu\text{s}$, $E_{on}=4,1\text{J}$

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We commit.
We innovate.
We partner.
We create value.



Never stop thinking