

General Description

The D804 is a high performance AC/DC power supply controller for battery charger and adapter applications. The device uses Pulse Frequency Modulation (PFM) method to build discontinuous conduction mode (DCM) flyback power supplies.

The D804 provides constant voltage, constant current (CV/CC) regulation without requiring an optocoupler and secondary control circuitry. It also eliminates the need of loop compensation circuitry while maintaining stability.

The D804 achieves excellent regulation and high power efficiency, the no-load power consumption is less than 200mW at 265VAC input.

Features

- Primary Side Control for Rectangular Constant Current and Constant Voltage Output
- Eliminates Opto-Coupler and Secondary CV/CC Control Circuitry
- Eliminates Control Loop Compensation Circuitry
- Flyback Topology in DCM Operation
- Random Frequency Modulation to Reduce System EMI
- Valley Turn on of External Power NPN Transistor
- Built-in Soft Start
- Open Circuit Protection
- Over Voltage Protection
- Short Circuit Protection

Applications

- Adapters/Chargers for Cell/Cordless Phones, PDAs, MP3 and Other Portable Apparatus
- Standby and Auxiliary Power Supplies

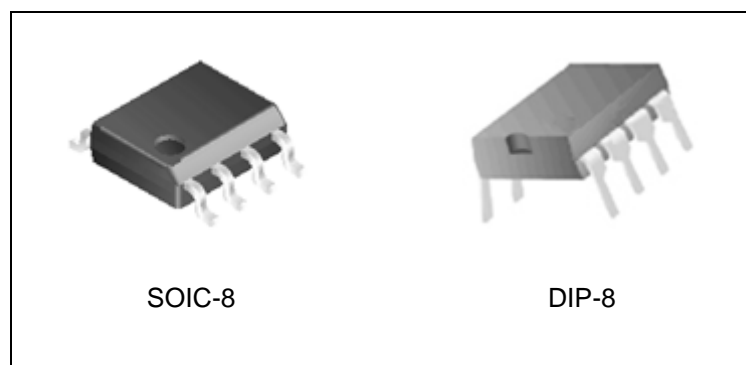


Figure 1. Package Types of D804

Pin Configuration

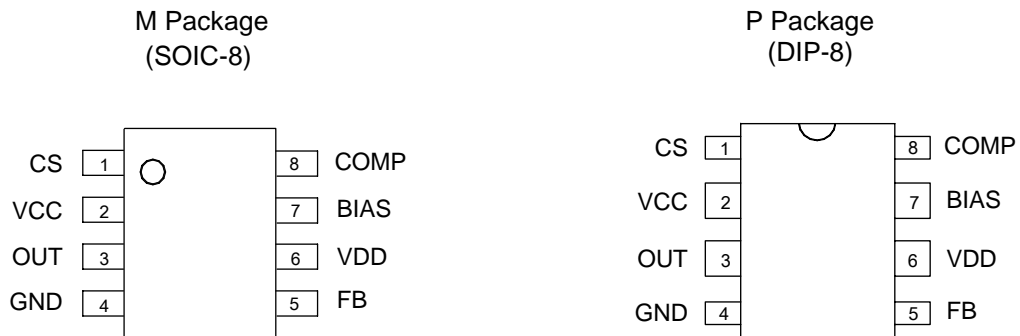


Figure 2. Pin Configurations of D804 (Top View)

Pin Description

Pin Number	Pin Name	Function
1	CS	The primary current sense
2	VCC	Supply voltage
3	OUT	This pin drives the base of external power NPN switch
4	GND	Ground
5	FB	The voltage feedback from the auxiliary winding
6	VDD	The 5.5V output of the internal voltage regulator
7	BIAS	This pin sets the bias current inside D804 with an external resistance to GND
8	COMP	This pin connects a bypass capacitor for CC function

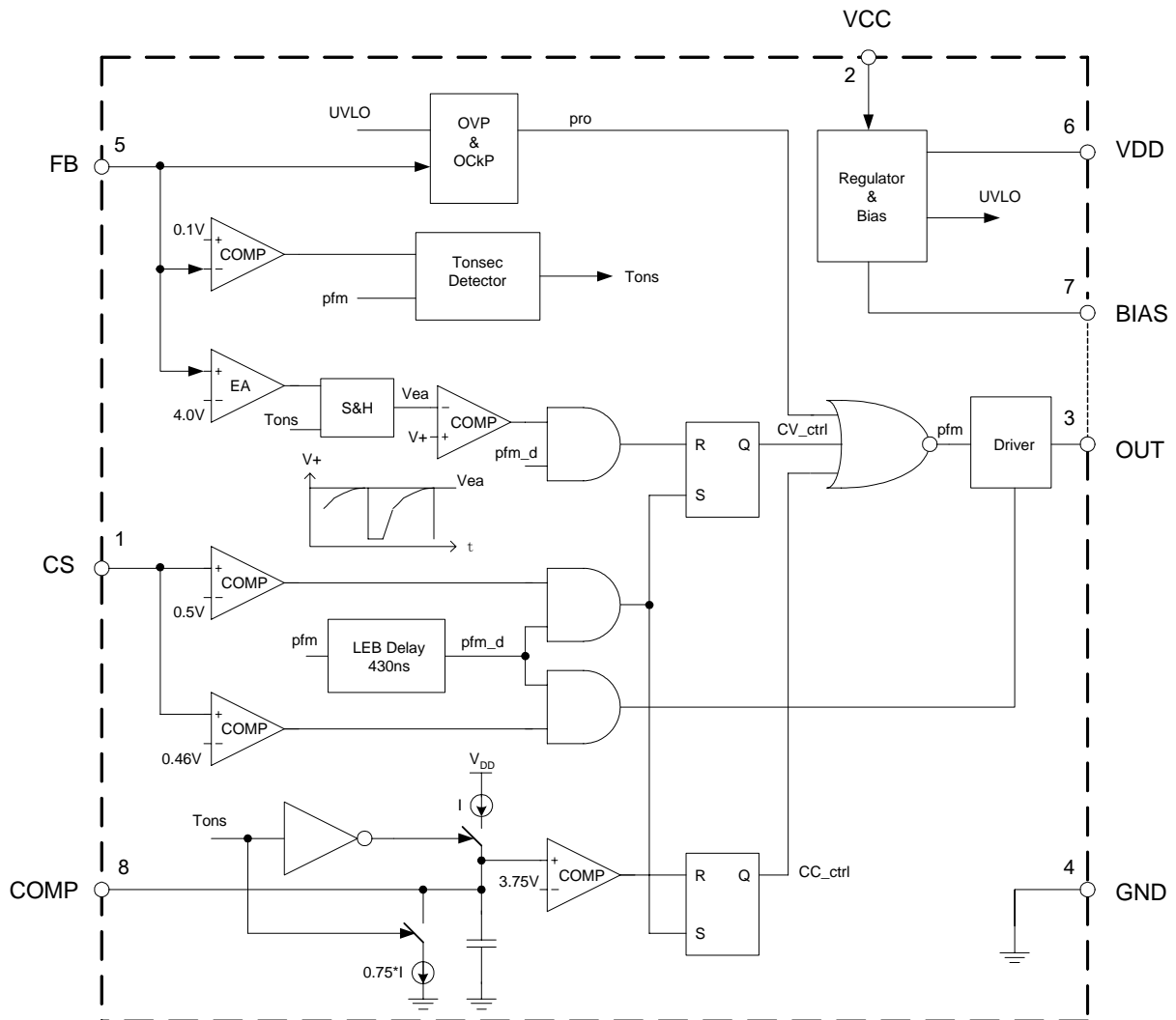


Figure 3. Functional Block Diagram of D804



D804

Electrical Characteristics

($V_{CC}=15V$, $T_A=25^{\circ}C$, unless otherwise specified.)

Parameter	Symbol	Conditions	Min	Typ	Max	Unit
UVLO SECTION						
Start-up Threshold	$V_{TH(ST)}$		17.5	19	20.5	V
Minimal Operating Voltage	$V_{OPR(min)}$	After turn on	7	8.7	9.8	V
REFERENCE VOLTAGE SECTION						
BIAS Pin Voltage	V_{BIAS}	$R_{BIAS}=200k\Omega$ Before turn on	1.150	1.19	1.23V	
VDD Pin Voltage	VDD		5.2	5.5	5.8	V
STANDBY CURRENT SECTION						
Start-up Current	I_{ST}	$V_{CC} = V_{TH(ST)} - 0.5V$, $R_{BIAS}=200k\Omega$, Before turn on		70	80	μA
Operating Current	$I_{CC(OPR)}$	$R_{BIAS}=200k\Omega$		680	900	μA
DRIVE OUTPUT SECTION						
OUT Maximum Current	Sink	I_{OUT}	$R_{BIAS}=200k\Omega$	50		mA
	Source			25	30	
CURRENT SENSE SECTION						
Current Sense Threshold	V_{CS}		485	505	525	mV
Pre-Current Sense	$V_{CS(PRE)}$		385	410	435	mV
Leading Edge Blanking				430		ns
FEEDBACK INPUT SECTION						
Feedback Pin Input Leakage Current	I_{FB}	$V_{FB}=4V$	7.0	9.8	12.6	μA
Feedback Threshold	V_{FB}		3.90	4.00	4.10	V
Enable Turn-on Voltage	$V_{FB(EN)}$		-0.9	-0.7	-0.5	V
COMP THRESHOLD VOLTAGE SECTION						
Turn-on Threshold Voltage	V_{COMP}		3.42	3.60	3.78	V
PROTECTION SECTION						
Over Voltage Protection	$V_{FB(OVP)}$		7	8	9	V

Operation Description

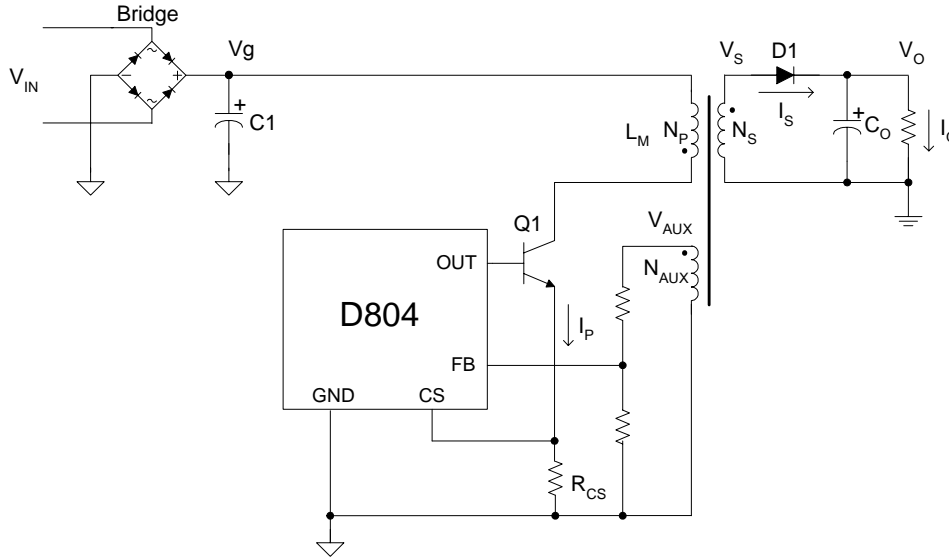


Figure 4. Simplified Flyback Converter Controlled by D804

Figure 4 illustrates a simplified flyback converter controlled by D804.

The energy stored in the magnetizing inductance L_M each cycle is therefore:

Constant Primary Peak Current

The primary current $i_p(t)$ is sensed by a current sense resistor R_{CS} as shown in Figure 4.

$$E_g = \frac{1}{2} \times L_M \cdot I_{pk}^2 \quad \dots\dots\dots(3)$$

The current rises up linearly at a rate of:

So the power transferring from the input to the output is given by:

$$\frac{di_p(t)}{dt} = \frac{v_g(t)}{L_M} \quad \dots\dots\dots(1)$$

$$P = \frac{1}{2} \times L_M \times I_{pk}^2 \times f_{sw} \quad \dots\dots\dots(4)$$

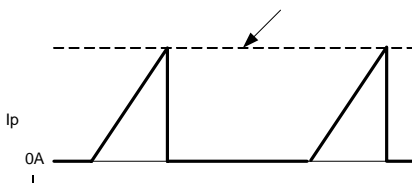


Figure 5. Primary Current Waveform

Where the f_{sw} is the switching frequency. When the peak current I_{pk} is constant, the output power depends on the switching frequency f_{sw} .

As illustrated in Figure 5, when the current $i_p(t)$ rises up to I_{pk} , the switch Q1 turns off. The constant peak current is given by:

Constant Voltage Operation

The D804 captures the auxiliary winding feedback voltage at FB pin and operates in constant-voltage (CV) mode to regulate the output voltage. Assuming the secondary winding is master, the auxiliary winding is slave during the D1 on-time. The auxiliary voltage is given by:

$$I_{pk} = \frac{V_{CS}}{R_{CS}} \quad \dots\dots\dots(2)$$

$$V_{AUX} = \frac{N_{AUX}}{N_S} \times (V_o + V_d) \quad \dots\dots\dots(5)$$

Operation Description (Continued)

Where the V_d is the diode forward drop voltage.

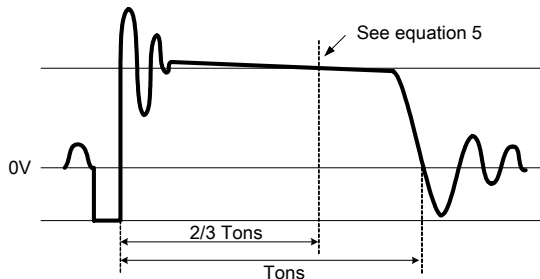


Figure 6. Auxiliary Voltage Waveform

The output voltage is different from the secondary voltage in a diode forward drop voltage. The diode drop voltage depends on the current. If the secondary voltage is always detected at a constant secondary current, the difference between the output voltage and the secondary voltage will be a fixed V_d . The voltage detection point is at two-thirds of the D1 on-time. The CV loop control function of D804 then generates a D1 off-time to regulate the output voltage.

Constant Current Operation

The D804 is designed to work in constant-current (CC) mode. Figure 7 shows the secondary current waveforms.

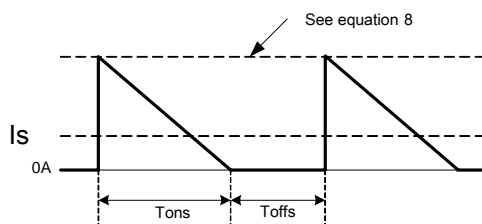


Figure 7 . Secondary Current Waveform

In CC operation, the CC loop control function of D804 will keep a fixed proportion between D1 on-time T_{ons} and D1 off-time T_{offs} by discharging or charging the capacitance connected in COMP pin. The fixed proportion is

$$\frac{T_{ons}}{T_{offs}} = \frac{4}{3} \quad \dots\dots\dots(6)$$

The relationship between the output constant-current and secondary peak current I_{pks} is given by:

$$I_{out} = \frac{1}{2} \times I_{pks} \times \frac{T_{ons}}{T_{ons} + T_{offs}} \quad \dots\dots\dots(7)$$

At the instant of D1 turn-on, the primary current transfers to the secondary at an amplitude of:

$$I_{pks} = \frac{N_p}{N_s} \times I_{pk} \quad \dots\dots\dots(8)$$

Thus the output constant-current is given by:

$$I_{out} = \frac{1}{2} \times \frac{N_p}{N_s} \times I_{pk} \times \frac{T_{ons}}{T_{ons} + T_{offs}} = \frac{2}{7} \times \frac{N_p}{N_s} \times I_{pk} \quad \dots\dots\dots(9)$$

Leading Edge Blanking

When the power switch is turned on, a turn-on spike will occur on the sense-resistor. To avoid false-termination of the switching pulse, a 430ns leading-edge blanking is built in. During this blanking period, the current sense comparator is disabled and the gate driver can not be switched off.

CCM Protection

The D804 is designed to operate in discontinuous conduction mode (DCM) in both CV and CC modes. To avoid operating in continuous conduction mode (CC), the D804 detects the falling edge of the FB input voltage on each cycle. If a 0.1V falling edge of FB is not detected, the D804 will stop switching.

OVP & OCKP

The D804 includes output over-voltage protection (OVP) and open circuit protection (OCKP) circuitry as shown in Figure 8. If the voltage at FB pin exceeds 8V, 100% above the normal detection voltage, or the -0.7V falling edge of the FB input can not be monitored, the D804 will immediately shut off and enters hiccup mode. The D804 sends out a fault detection pulse every 8ms in hiccup mode until the fault has been removed.

Operation Description (Continued)

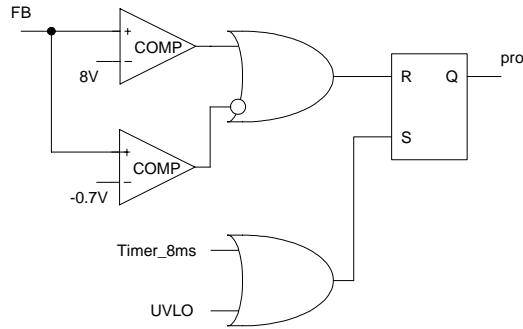


Figure 8. OVP and OCP Function Block

Typical Application

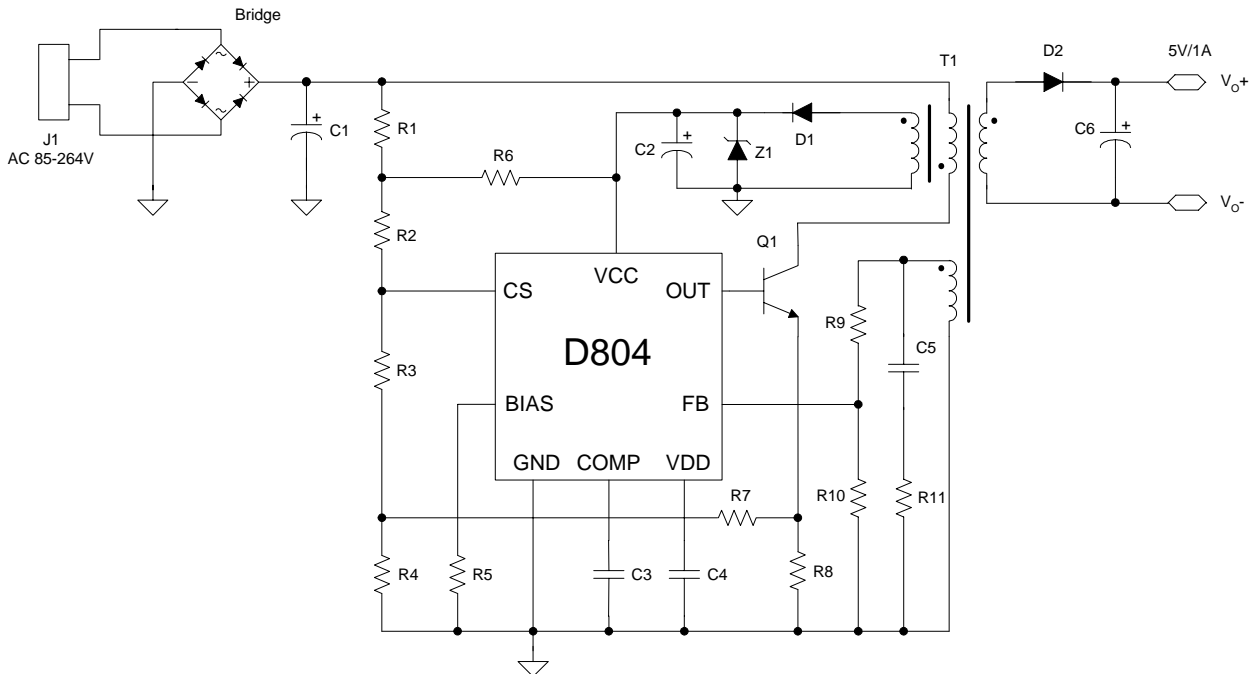


Figure 9. 5V/1A Output for Battery Charger of Mobile Phone