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|------------------------|--|
| Title | <i>Reference Design Report for a 6 W Constant Voltage (CV) Adapter Using LNK625PG</i> |
| Specification | 90 – 265 VAC Input; 5 V, 1.2 A Output |
| Application | Low-Cost Charger or Adapter |
| Author | Applications Engineering Department |
| Document Number | RDR-201 |
| Date | November 11, 2008 |
| Revision | 1.5 |

Summary and Features

- Eliminates optocoupler and all secondary side control circuitry
- EcoSmart[®] – Easily meets all existing and proposed international energy efficiency standards – China (CECP) / EISA / Energy STAR / European Commission
 - ON/OFF control provides constant efficiency event at very light loads
 - No-load consumption <100 mW at 265 VAC
 - Meets ENERGY STAR 2.0 active mode efficiency
 - 76 % vs 70 % requirement at board
 - 73 % vs 70 % requirement at end of 6 ft, 315 mΩ output cable
 - Ultra-low leakage current: <5 μA at 265 VAC input (no Y capacitor required)
- Over-temperature protection – tight tolerance (±5%) with hysteretic recovery for safe PCB temperatures under all conditions
- Auto-restart output short circuit and open-loop protection
- Easy compliance to EN550022 and CISPR-22 Class B EMI standards
- Green package: halogen free and RoHS compliant

PATENT INFORMATION

The products and applications illustrated herein (including transformer construction and circuits external to the products) may be covered by one or more U.S. and foreign patents, or potentially by pending U.S. and foreign patent applications assigned to Power Integrations. A complete list of Power Integrations' patents may be found at www.powerint.com. Power Integrations grants its customers a license under certain patent rights as set forth at <<http://www.powerint.com/ip.htm>>.

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Important Note:

Although this board is designed to satisfy safety isolation requirements, the engineering prototype has not been agency approved. Therefore, all testing should be performed using an isolation transformer to provide the AC input to the prototype board.



1 Introduction

This document is an engineering report describing a universal input, 5 V, 6 W output Flyback power supply. This reference design is based on the LinkSwitch-CV family of devices and utilizes the LNK625PG.

The document contains the power supply specification, schematic, bill of materials, transformer documentation, printed circuit layout, and performance data.



Figure 1 – Populated Circuit Board Photograph.

2 Power Supply Specification

| Description | Symbol | Min | Typ | Max | Units | Comment |
|--|------------------------|--|-------|------|----------|--|
| Input | | | | | | |
| Voltage | V_{IN} | 90 | | 265 | VAC | 2 Wire – no P.E. |
| Frequency | f_{LINE} | 47 | 50/60 | 64 | Hz | |
| No-load Input Power (230 VAC) | | | | 90 | mW | |
| Output | | | | | | |
| Output Voltage | V_{OUT} | 4.75 | 5.0 | 5.25 | V | ± 5% 20 MHz bandwidth |
| Output Ripple Voltage | V_{RIPPLE} | | 100 | | mV | |
| Output Current | I_{OUT} | 0 | | 1.2 | A | |
| Total Output Power | | | | | | |
| Continuous Output Power | P_{OUT} | 6 | | | W | |
| Efficiency | | | | | | |
| Full Load | η η_{CBL} | 76 | 71 | | % % | Measured at P_{OUT} 25 °C At board terminals At end of 6 ft, 315 mΩ output cable |
| Required average efficiency at 25, 50, 75 and 100 % of P_{OUT} | $\eta_{ES2.0}$ | 70 | | | % | Per ENERGY STAR V2.0 |
| Measured average efficiency | | 76 | | | % | At board |
| | | 73 | | | % | At end of 6 ft, 315 mΩ output cable |
| Environmental | | | | | | |
| Conducted EMI | | Meets CISPR22B / EN55022B | | | | 1.2/50 μs surge, IEC 1000-4-5, Series Impedance: Differential Mode: 2 Ω Common Mode: 12 Ω |
| Safety | | Designed to meet IEC950 / UL1950 Class II | | | | |
| Line Surge | | | | | | |
| Differential Mode (L1-L2) Common mode (L1/L2-PE) | | 1 2 | | | kV kV | |
| Ambient Temperature | T_{AMB} | 0 | 25 | 40 | °C | Free convection, sea level |



3 Schematic

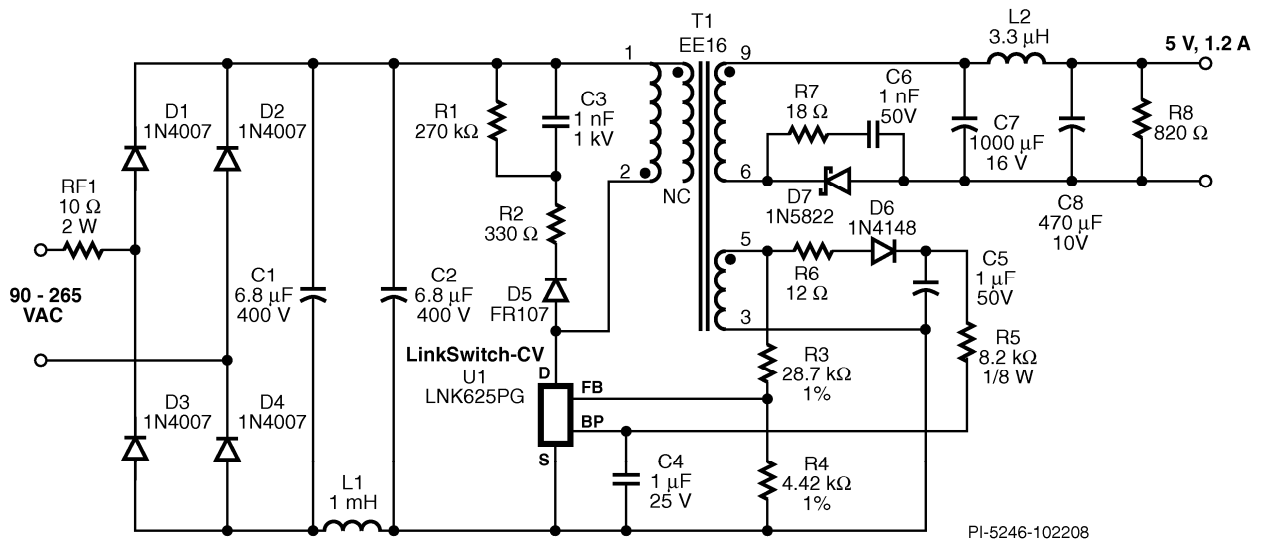


Figure 2 – Schematic



4 Circuit Description

The schematic in Figure 2 shows an adapter design using the LNK625PG that provides very tight constant voltage (CV) performance. The circuit is designed to operate from 90 VAC to 265 VAC input, with an output voltage of 5 V providing a maximum load current of 1.2 A. It consumes very little standby power and uses no Y-Capacitor to meet stringent EMI requirements. The adapter also meets and exceeds both CEC and Energy Star 2.0 regulations.

4.1 Input and EMI Filtering

Diodes D1 through D4 form a full wave rectifier. The rectified DC is then filtered by capacitors C1 and C2. Inductor L1 forms a pi (π) filter with capacitors C1 and C2 which helps to reduce differential EMI noise. This filtering, together with the integrated switching frequency jitter provided in U1 and transformer E-Shield techniques, provide a generous EMI margin without the need for a Y capacitor across the primary and secondary windings of transformer T1.

4.2 LinkSwitch-CV Device

The LinkSwitch-CV family of devices has been developed to cost effectively replace all existing solutions in low power adapter applications. It is optimized for constant voltage (CV) adapter applications while using minimal external parts including the complete elimination of the optocoupler and shunt regulator.

The LNK625PG monolithically integrates the 700 V power MOSFET switch and controller, which consists of an oscillator, feedback (sense and logic) circuit, 6 V regulator, BYPASS pin programming functions, over-temperature protection, frequency jittering, current limit circuit and leading-edge blanking.

The LNK625PG also provides a sophisticated range of protection features including auto-restart for control loop component open/short circuit faults and output short circuit conditions. The use of a low auto-restart on time reduces the power delivered by more than 95 % for output short circuits and control loop faults. Accurate hysteretic thermal shutdown ensures safe average PCB temperatures under all conditions. Extended creepage distance between high and low voltage pins prevent arcing and helps meet safety requirements. *LinkSwitch-CV* also can be used without a bias winding as the IC is completely self biased.

4.3 Primary Circuit

During U1's on time current flows through the primary winding of transformer T1 and stores energy in its magnetic field. During U1's off time, the energy stored in the transformer is transferred to the secondary side, delivering current to both the output capacitors and the load.



The clamp circuit formed by resistors R1 and R2 along with blocking diode D5 and capacitor C3 ensures that the drain node voltage is well below the 700 V rating of the internal MOSFET of U1. The clamp circuit is also carefully designed to reduce and dampen any oscillation present in the voltage spike caused by the transformer's leakage inductance.

4.4 Output Rectification

The secondary output is rectified by diode D7 which is placed in the return leg to help reduce EMI and simplify the transformer construction. An RC snubber circuit composed of resistor R7 and capacitor C6 is placed across the output diode to also reduce high frequency EMI. A stable output voltage is maintained by capacitor C7.

Inductor L2 and capacitor C8 form an LC post filter which helps to attenuate switching noise and reduces output ripple. Resistor R8 is a preload resistor whose value has been empirically chosen to provide the best possible regulation at light loads without significantly affecting no-load input power or efficiency.

4.5 Feedback Winding

LinkSwitch-CV eliminates the need for an optocoupler for tight output voltage regulation, as good as $\pm 5\%$, through the use of a feedback winding. The feedback (FB) pin voltage, which is derived from the voltage divider formed by resistors R3 and R4, is sampled approximately 2.5 μs after U1's internal MOSFET turns off. Based upon this information the device regulates the output voltage.

The feedback winding was also designed with more turns than necessary so that it may act as a bias winding. The winding provides bias current to U1 through the bypass pin (BP) and reduces the input power consumption during light loads and no-load conditions. Resistor R6 helps to dampen out any ringing present on the feedback winding and ensures that the waveform at the FB pin at 2.5 μs is free from any ringing. Capacitor C5 provides a stable bias voltage while resistor R5 is chosen to supply the necessary BP pin current. Capacitor C4 is the BP pin capacitor and should be placed as close as possible to the BP pin and source pins of the device.



5 PCB Layout

Notable layout design points are:

- 1 A spark gap and associated slot in the PCB between the primary and secondary is placed to protect the power supply from electro-static discharge (ESD).
 - The preferential arcing point routes the energy from ESD discharges back to the AC input, away from the transformer and primary circuitry.
 - The trace connected to the AC input side of the spark gap is spaced away from the rest of the board and its components to prevent arc discharges to other sections of the circuit.
- 2 The drain trace length (pin 4 of U1) has been minimized to reduce EMI.
- 3 Clamp and output diode loop areas are minimized to reduce EMI.
- 4 A large copper area around the (electrically quiet) Source pins is used to provide heatsinking. Provide sufficient copper area to keep the source pin temperature below 90 °C.
- 5 The AC input is located away from switching nodes to minimize noise coupling that may bypass input filtering.
- 6 Place C4 (the bypass capacitor) as close as possible to the BYPASS pin on U1.

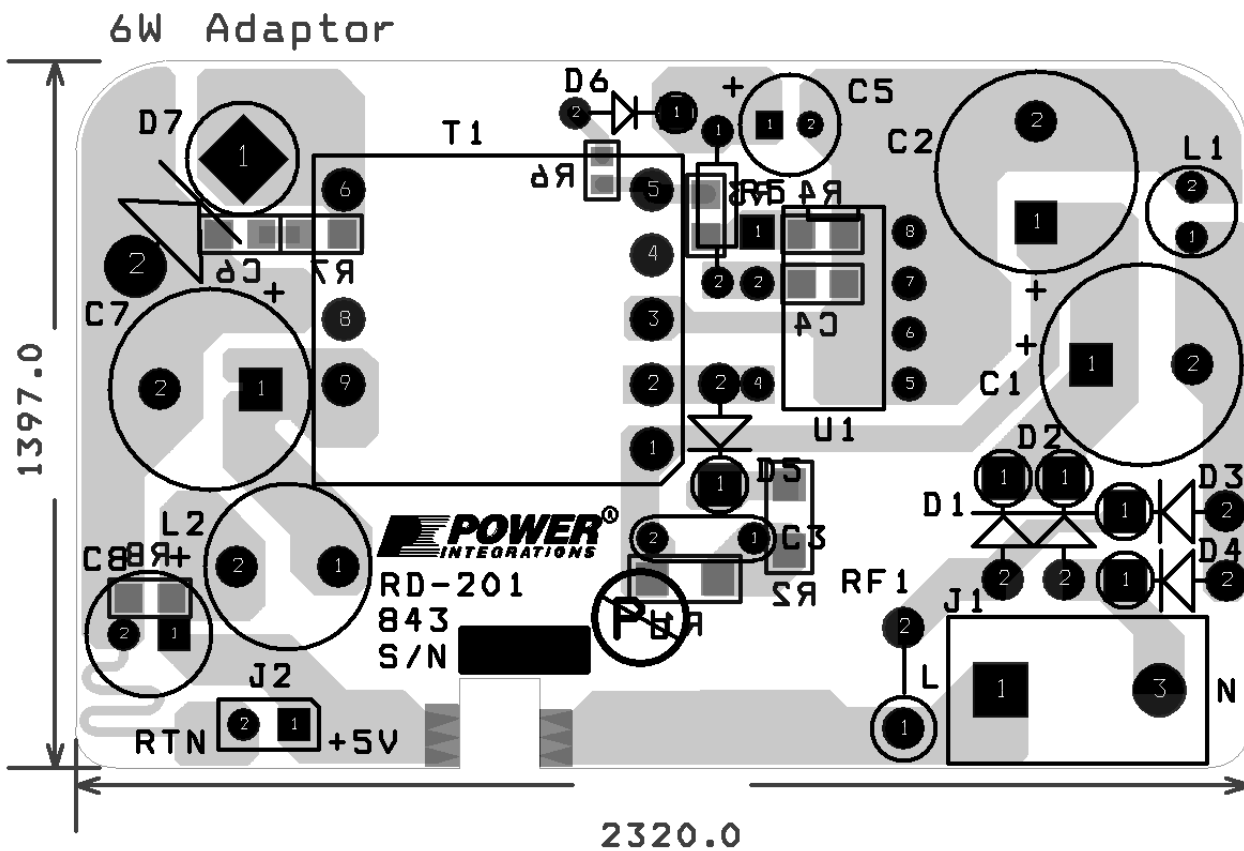


Figure 3 – Printed Circuit Layout, dimensions in mils (one thousandth of an inch).

6 Bill of Materials

| Item | Qty | Ref Des | Description | Manufacturer Part Number | Manufacturer |
|------|-----|----------------|---|---|---|
| 1 | 2 | C1 C2 | 6.8 μ F, 400 V, Electrolytic, (10 x 16) | EKXG401ELL6R8MJ16S | Nippon Chemi-Con |
| 2 | 1 | C3 | 1 nF, 1 kV, Disc Ceramic | ECK-D3A102KBP | Panasonic - ECG |
| 3 | 1 | C4 | 1 μ F, 25 V, Ceramic, X7R, 0805 | ECJ-2FB1E105K | Panasonic |
| 4 | 1 | C5 | 1 μ F, 50 V, Electrolytic, Gen. Purpose, (5 x 11) | EKMG500ELL1R0ME11D | Nippon Chemi-Con |
| 5 | 1 | C6 | 1 nF, 50 V, Ceramic, X7R, 0805 | ECJ-2VB1H102K | Panasonic |
| 6 | 1 | C7 | 1000 μ F, 16 V, Electrolytic, Very Low ESR, 23 m Ω , (10 x 20) | EKZE160ELL102MJ20S | Nippon Chemi-Con |
| 7 | 1 | C8 | 470 μ F, 10 V, Electrolytic, Very Low ESR, 72 m Ω , (8 x 11.5) | EKZE100ELL471MHB5D | Nippon Chemi-Con |
| 8 | 4 | D1 D2 D3 D4 | 1000 V, 1 A, Rectifier, DO-41 | 1N4007-E3/54 | Vishay |
| 9 | 1 | D5 | 1000 V, 1 A, Fast Recovery Diode, DO-41 | FR107-T-F | Diodes Inc. |
| 10 | 1 | D6 | 75 V, 300 mA, Fast Switching, DO-35 | 1N4148 | Vishay |
| 11 | 1 | D7 | 40 V, 3 A, Schottky, DO-41 | 1N5822 | Vishay |
| 12 | 1 | J1 | 3 Position (1 x 3) header, 0.156 pitch, Vertical | 26-48-1031 | Molex |
| 13 | 1 | J2 | 2 Position (1 x 2) header, 0.1 pitch, Vertical | 22-03-2021 | Molex |
| 14 | 1 | L1 | 1 mH, 0.15 A, Ferrite Core | SBCP-47HY102B | Tokin |
| 15 | 1 | L2 | 3.3 μ H, 5.5 A | RL622-3R3K-RC | JW Miller |
| 16 | 1 | R1 | 270 k, 5%, 1/4 W, Metal Film, 1206 | ERJ-8GEYJ274V | Panasonic |
| 17 | 1 | R2 | 330 Ω , 5%, 1/4 W, Metal Film, 1206 | ERJ-8GEYJ331V | Panasonic |
| 18 | 1 | R3 | 28.7 k Ω , 1%, 1/8 W, Metal Film, 0805 | ERJ-6ENF2872V | Panasonic |
| 19 | 1 | R4 | 4.42 k Ω , 1%, 1/8 W, Metal Film, 0805 | ERJ-6ENF4421V | Panasonic |
| 20 | 1 | R5 | 8.2 k Ω , 5%, 1/8 W, Carbon Film | CFR-12JB-8K2 | Yageo |
| 21 | 1 | R6 | 12 Ω , 5%, 1/10 W, Metal Film, 0603 | ERJ-3GEYJ120V | Panasonic |
| 22 | 1 | R7 | 18 Ω , 5%, 1/8 W, Metal Film, 0805 | ERJ-6GEYJ180V | Panasonic |
| 23 | 1 | R8 | 820 Ω , 5%, 1/8 W, Metal Film, 0805 | ERJ-6GEYJ821V | Panasonic |
| 24 | 1 | RF1 | 10 Ω , 2 W, Fusible/Flame Proof Wire Wound | CRF253-4 10R | Vitrohm |
| 25 | 1 | T1 | Bobbin, EE16, Horizontal, 10 pins | PM-9820 Santronics Würth Electronics Ice Components Precision Electronics | Ho Jinn Plastic Elect. Co. SNX R1494 750811016 TP08152 019-6361-00R |
| 26 | 1 | U1 | LinkSwitch-II, LNK625PG, CV, DIP-8C | LNK625PG | Power Integrations |



7 Transformer Specification

7.1 Electrical Diagram

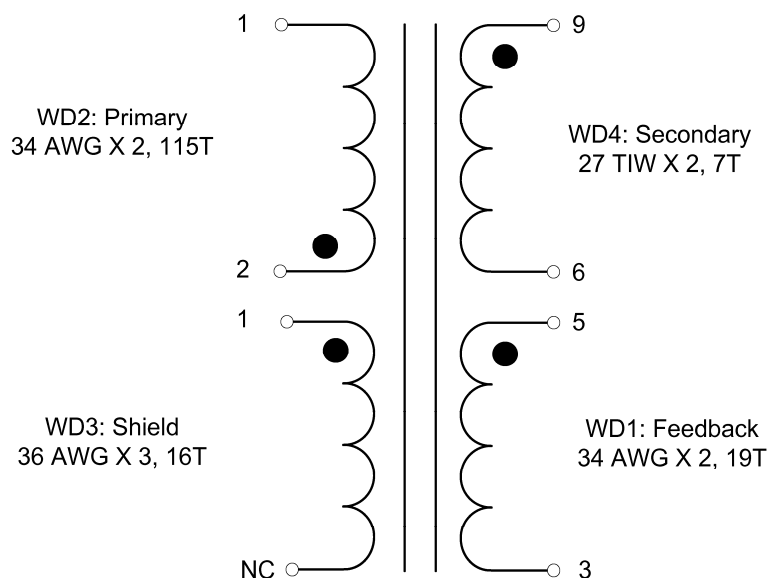


Figure 4 – Transformer Electrical Diagram.

7.2 Electrical Specifications

| | | |
|-----------------------------------|--|--------------------|
| Electrical Strength | 1 second, 60 Hz, from Pins 1-2 to Pins 6-9 | 3000 VAC |
| Primary Inductance | Pins 1-2, all other windings open | 1.67 mH $\pm 10\%$ |
| Resonant Frequency | Pins 1-2, all other windings open | 500 kHz (Min.) |
| Primary Leakage Inductance | Pins 1-2, with Pins 6,9,3 and 5 shorted. | 100 μ H (Max.) |

7.3 Materials

| Item | Description |
|------|---|
| [1] | Core: EE16, NC-2H or equivalent, gapped for 126 nH/T ² |
| [2] | Bobbin: EE16, Horizontal, 10 pins, (5/5) |
| [3] | Magnet wire: #34 AWG, double coated. |
| [4] | Magnet wire: #36 AWG, double coated. |
| [5] | Triple Insulated wire: #27 TIW |
| [6] | Tape: 3M 1298 Polyester Film, 2 mils thick, 8.6 mm wide. |
| [7] | Varnish |



7.4 Transformer Build Diagram

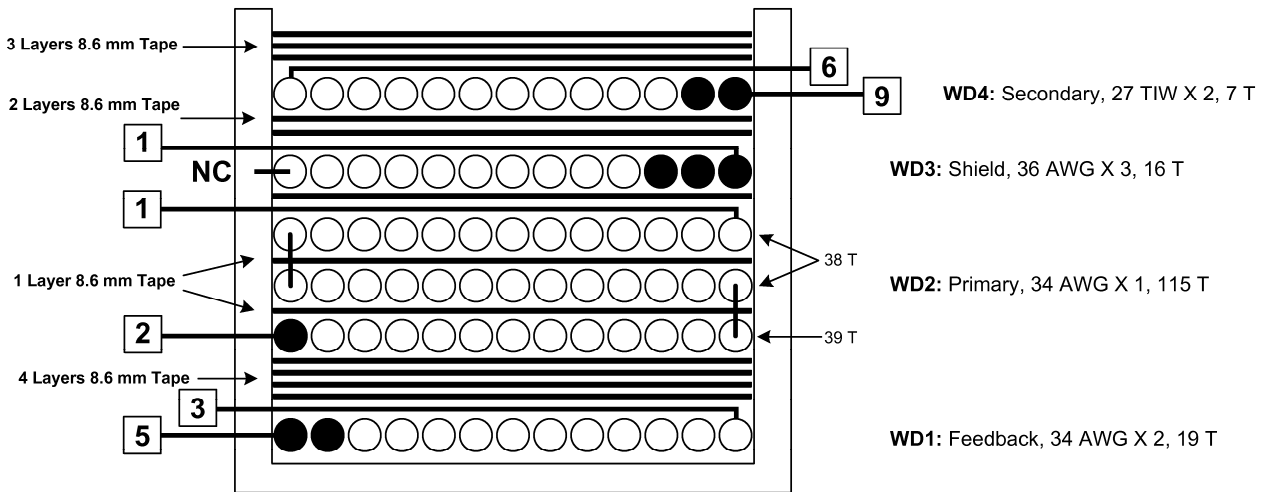


Figure 5 – Transformer Build Diagram.

7.5 Transformer Construction

| | |
|-----------------------------------|---|
| Bobbin Preparation | Pull pins 7 and 10. Orient the bobbin with primary pins on the left hand side. Winding direction is clockwise. |
| Winding 1: Feedback | Start on pin 5 and wind 19 turns of bifilar item [3] from left to right terminating on pin 3. |
| Insulation | Apply 4 layers of tape item [6]. |
| Winding 2: Primary Winding | Start on pin 2 and wind 39 turns of item [3], with tight tension, from left to right filling one layer. Apply one layer of tape item [6]. Continue winding 38 turns of item [3] from right to left and apply one layer of tape item [6]. Wind the remaining 38 turns of item [3] from left to right and terminate the winding on pin 1. |
| Insulation | Apply one layer of tape item [6]. |
| Winding 3: Shield | Start on any temporary secondary pin, leaving extra wire, and wind 16 turns of trifilar item [4] from right to left filling one layer. Secure the winding temporarily with tape and cut the finish end of the winding. Move the starting end of the winding to pin 1. |
| Insulation | Wrap two complete layers of tape over any leads of the previous winding. |
| Winding 4: Secondary | Start on pin 9 and wind 7 turns of item [6] from right to left terminating on pin 6. |
| Insulation | Apply 2 layers of tape item [6] |
| Assembly | Assemble core halves and tape together after properly gapping center leg of core. |
| Finish | Dip Varnish |

8 Transformer Spreadsheets

| ACDC_LNK-CV_102208; Rev.1.1; Copyright Power Integrations 2008 | INPUT | INFO | OUTPUT | UNIT | ACDC_LNK-CV_102208_Rev1-1.xls; LinkSwitch-CV Continuous/Discontinuous Flyback Transformer Design Spreadsheet |
|---|-------|------|--------|----------|--|
| ENTER APPLICATION VARIABLES | | | | | RDk-201: 5 V, 6 W Adapter Design |
| VACMIN | 90 | | | Volts | Minimum AC Input Voltage |
| VACMAX | 265 | | | Volts | Maximum AC Input Voltage |
| fL | 50 | | | Hertz | AC Mains Frequency |
| VO | 5.00 | | | Volts | Output Voltage |
| PO | 6.00 | | | Watts | Output Power |
| n | 0.72 | | | | Efficiency Estimate |
| Z | | | 0.5 | | Loss Allocation Factor |
| tC | | | 3 | mSeconds | Bridge Rectifier Conduction Time Estimate |
| CIN | 13.60 | | | uFarads | Input Filter Capacitor |

| ENTER LinkSwitch-CV VARIABLES | | | | | |
|--------------------------------------|----------------|--|----------------|-------------------|--|
| LinkSwitch-CV | LNK625P | | LNK625P | | Chosen LinkSwitch-CV device |
| ILIMITMIN | | | 0.307 | Amps | LinkSwitch-CV Minimum Current Limit |
| ILIMITMAX | | | 0.353 | Amps | LinkSwitch-CV Maximum Current Limit |
| fS | | | 100000 | Hertz | LinkSwitch-CV Switching Frequency |
| I2FMIN | | | 9801 | A ² Hz | LinkSwitch-CV Min I2F (power Coefficient) |
| I2FMAX | | | 12741 | A ² Hz | LinkSwitch-CV Max I2F (power Coefficient) |
| VOR | | | 90 | Volts | Reflected Output Voltage |
| VDS | | | 10 | Volts | LinkSwitch-CV on-state Drain to Source Voltage |
| VD | | | 0.5 | Volts | Output Winding Diode Forward Voltage Drop |
| DCON | | | 5.11 | us | Output Diode conduction time |
| KP_TRANSIENT | | | 0.82 | | Worst case ripple to peak current ratio. Maintain KP_TRANSIENT below 0.25 |

| ENTER TRANSFORMER CORE/CONSTRUCTION VARIABLES | | | | | |
|--|-------------|-------------|-------------|-------------------|---|
| Core Type | Auto | | EE16 | | Transformer Core size |
| Core | | EE16 | | P/N: | PC40EE16-Z |
| Bobbin | | EE16_BOBBIN | | P/N: | BE-16-118CPH |
| AE | | | 0.192 | cm ² | Core Effective Cross Sectional Area |
| LE | | | 3.5 | cm | Core Effective Path Length |
| AL | | | 1140 | nH/T ² | Ungapped Core Effective Inductance |
| BW | | | 8.5 | mm | Bobbin Physical Winding Width |
| M | | | 0.00 | mm | Safety Margin Width (Half the Primary to Secondary Creepage Distance) |
| L | | | 3 | | Number of Primary Layers |



| | | | | | |
|------------------------------------|-------|--|-------|--------|---|
| NS | | | 7 | | Number of Secondary Turns |
| DC INPUT VOLTAGE PARAMETERS | | | | | |
| VMIN | | | 87 | Volts | Minimum DC Input Voltage |
| VMAX | | | 375 | Volts | Maximum DC Input Voltage |
| FEEDBACK VARIABLES | | | | | |
| NFB | 19.00 | | 19.00 | | Feedback winding number of turns |
| VFLY | | | 14.93 | Volts | Voltage on the Feedback winding when LinkSwitch-CV turns off |
| RUPPER | | | 28.70 | k-ohms | Upper resistor of feedback network |
| RLOWER | | | 4.53 | k-ohms | Lower resistor of feedback network |
| Fine Tuning Section | | | | | |
| Measured Output Voltage | 5.10 | | 5.10 | k-ohms | Actual (Measured) Voltage at the output of power supply |
| RLOWER_FINE | | | 4.42 | k-ohms | Adjusted (Fine tuned) value of lower resistor (RLOWER). Do not change value of RUPPER |

| | | | | | |
|--------------------------------|-----|--|-----|-------|---|
| Bias Winding Parameters | | | | | |
| Add Bias winding | YES | | NO | | Bias winding is not necessary. The feedback winding itself can be used as a bias winding |
| VB | | | N/A | Volts | Bias Winding Voltage |
| NB | | | N/A | | Number of Bias winding turns. Bias winding is assumed to be AC stacked on top of the Feedback winding |

| | | | | | |
|--|--|--|------|------|------------------------------|
| CURRENT WAVEFORM SHAPE PARAMETERS | | | | | |
| DMAX | | | 0.54 | | Maximum Duty Cycle |
| I AVG | | | 0.10 | Amps | Average Primary Current |
| IP | | | 0.31 | Amps | Minimum Peak Primary Current |
| IR | | | 0.26 | Amps | Primary Ripple Current |
| IRMS | | | 0.14 | Amps | Primary RMS Current |

| | | | | | |
|--|--|--|------|----------|--|
| TRANSFORMER PRIMARY DESIGN PARAMETERS | | | | | |
| LPMIN | | | 1499 | uHenries | Minimum Primary Inductance |
| LP_TYP | | | 1649 | uHenries | Typical (Nominal) Primary Inductance |
| LP_TOL | | | 10 | | Tolerance of Primary inductance |
| NP | | | 115 | | Primary Winding Number of Turns |
| ALG | | | 126 | nH/T^2 | Gapped Core Effective Inductance |
| BM | | | 2474 | Gauss | Maximum Flux Density, (BM<2500) Calculated at typical current limit and typical primary inductance |



| | | | | | |
|-----|--|--|------|-----------|---|
| BP | | | 2887 | Gauss | Peak Flux Density, (BP<3100) Calculated at maximum current limit and maximum primary inductance |
| BAC | | | 908 | Gauss | AC Flux Density for Core Loss Curves (0.5 X Peak to Peak) |
| ur | | | 1654 | | Relative Permeability of Ungapped Core |
| LG | | | 0.19 | mm | Gap Length (Lg > 0.1 mm) |
| BWE | | | 25.5 | mm | Effective Bobbin Width |
| OD | | | 0.22 | mm | Maximum Primary Wire Diameter including insulation |
| INS | | | 0.04 | mm | Estimated Total Insulation Thickness (= 2 * film thickness) |
| DIA | | | 0.18 | mm | Bare conductor diameter |
| AWG | | | 34 | AWG | Primary Wire Gauge (Rounded to next smaller standard AWG value) |
| CM | | | 40 | Cmils | Bare conductor effective area in circular mils |
| CMA | | | 285 | Cmils/Amp | Primary Winding Current Capacity (200 < CMA < 500) |

TRANSFORMER SECONDARY DESIGN PARAMETERS**Lumped parameters**

| | | | | | |
|---------|--|--|------|-------|---|
| ISP | | | 5.02 | Amps | Peak Secondary Current |
| ISRMS | | | 2.14 | Amps | Secondary RMS Current |
| IO | | | 1.20 | Amps | Power Supply Output Current |
| IRIPPLE | | | 1.77 | Amps | Output Capacitor RMS Ripple Current |
| CMS | | | 428 | Cmils | Secondary Bare Conductor minimum circular mils |
| AWGS | | | 23 | AWG | Secondary Wire Gauge (Rounded up to next larger standard AWG value) |
| DIAS | | | 0.58 | mm | Secondary Minimum Bare Conductor Diameter |
| ODS | | | 1.21 | mm | Secondary Maximum Outside Diameter for Triple Insulated Wire |
| INSS | | | 0.32 | mm | Maximum Secondary Insulation Wall Thickness |

VOLTAGE STRESS PARAMETERS

| | | | | | |
|--------|--|--|-----|-------|--|
| VDRAIN | | | 584 | Volts | Maximum Drain Voltage Estimate (Includes Effect of Leakage Inductance) |
| PIVB | | | N/A | Volts | Bias Diode Maximum Peak Inverse Voltage |
| PIVS | | | 28 | Volts | Output Rectifier Maximum Peak Inverse Voltage |



9 Performance Data

All measurements performed at room temperature, 60 Hz input frequency unless otherwise stated.

9.1 Active Mode Efficiency

Efficiency was measured at the end of a AWG24 cable 6 ft in length (cable resistance = 315 mΩ)

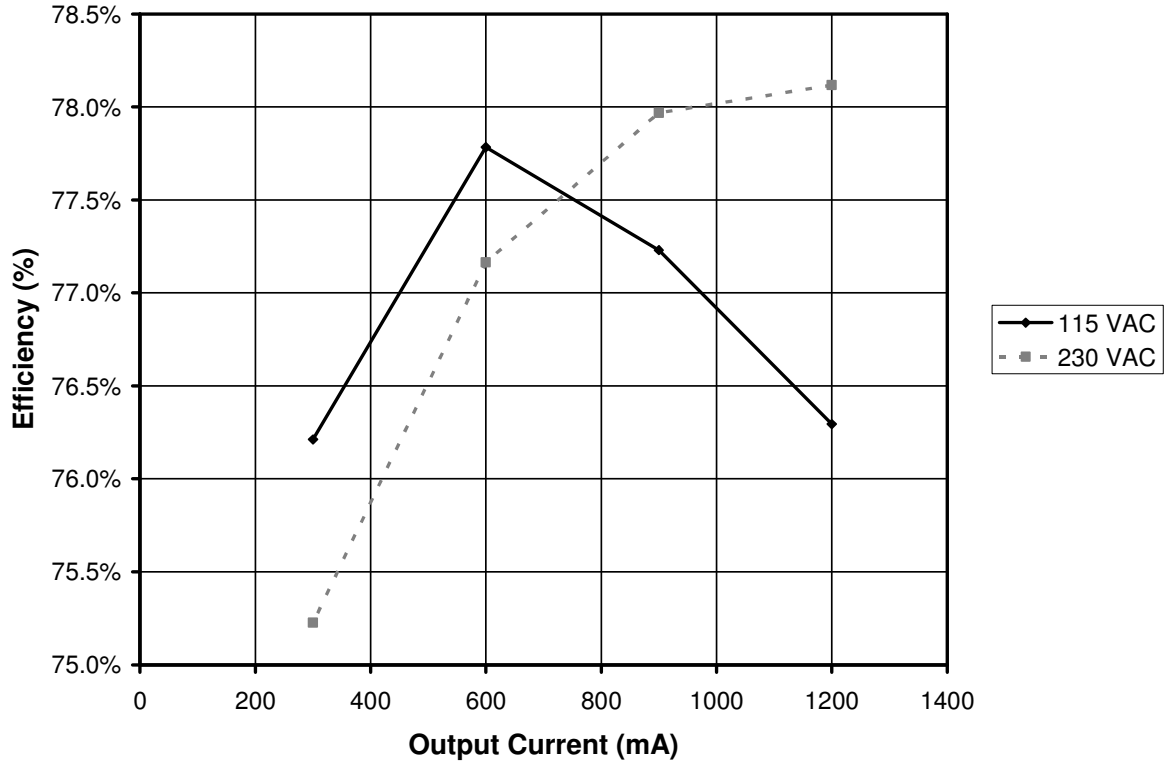


Figure 6 – Efficiency vs. Load, Room Temperature, 60 Hz.

| Percent of Full Load | Efficiency (%) | |
|-----------------------------|----------------|---------|
| | 115 VAC | 230 VAC |
| 25 | 74.6 | 73.6 |
| 50 | 74.6 | 74.3 |
| 75 | 72.9 | 73.6 |
| 100 | 70.8 | 72.5 |
| Average | 73.2 | 73.5 |
| US EISA (2007) requirement | 66.1 | |
| ENERGY STAR 2.0 requirement | 69.5 | |

Table 1 – Efficiency Table.



9.2 Energy Efficiency Requirements

The external power supply requirements below all require meeting active mode efficiency and no-load input power limits. Minimum active mode efficiency is defined as the average efficiency of 25, 50, 75 and 100% of output current (based on the nameplate output current rating).

For adapters that are single input voltage only then the measurement is made at the rated single nominal input voltage (115 VAC or 230 VAC), for universal input adapters the measurement is made at both nominal input voltages (115 VAC and 230 VAC).

To meet the standard the measured average efficiency (or efficiencies for universal input supplies) must be greater than or equal to the efficiency specified by the standard.

The test method can be found here:

http://www.energystar.gov/ia/partners/prod_development/downloads/power_supplies/EP_SupplyEffic_TestMethod_0804.pdf

For the latest up to date information please visit the PI Green Room:

<http://www.powerint.com/greenroom/regulations.htm>



9.2.1 USA Energy Independence and Security Act 2007

This legislation mandates all single output single output adapters, including those provided with products, manufactured on or after July 1st, 2008 must meet minimum active mode efficiency and no load input power limits.

Active Mode Efficiency Standard Models

| Nameplate Output (P_o) | Minimum Efficiency in Active Mode of Operation |
|----------------------------|--|
| < 1 W | $0.5 \times P_o$ |
| ≥ 1 W to ≤ 51 W | $0.09 \times \ln(P_o) + 0.5$ |
| > 51 W | 0.85 |

\ln = natural logarithm

No-load Energy Consumption

| Nameplate Output (P_o) | Maximum Power for No-load AC-DC EPS |
|----------------------------|-------------------------------------|
| All | ≤ 0.5 W |

This requirement supersedes the legislation from individual US States (for example CEC in California).



9.2.2 ENERGY STAR EPS Version 2.0

This specification takes effect on November 1st, 2008.

Active Mode Efficiency Standard Models

| Nameplate Output (P_O) | Minimum Efficiency in Active Mode of Operation |
|----------------------------|--|
| ≤ 1 W | $0.48 \times P_O + 0.14$ |
| > 1 W to ≤ 49 W | $0.0626 \times \ln(P_O) + 0.622$ |
| > 49 W | 0.87 |

\ln = natural logarithm

Active Mode Efficiency Low Voltage Models ($V_O < 6$ V and $I_O \geq 550$ mA)

| Nameplate Output (P_O) | Minimum Efficiency in Active Mode of Operation |
|----------------------------|--|
| ≤ 1 W | $0.497 \times P_O + 0.067$ |
| > 1 W to ≤ 49 W | $0.075 \times \ln(P_O) + 0.561$ |
| > 49 W | 0.86 |

\ln = natural logarithm

No-load Energy Consumption (both models)

| Nameplate Output (P_O) | Maximum Power for No-load AC-DC EPS |
|-----------------------------|-------------------------------------|
| 0 to < 50 W | ≤ 0.3 W |
| ≥ 50 W to ≤ 250 W | ≤ 0.5 W |



9.3 No-load Input Power

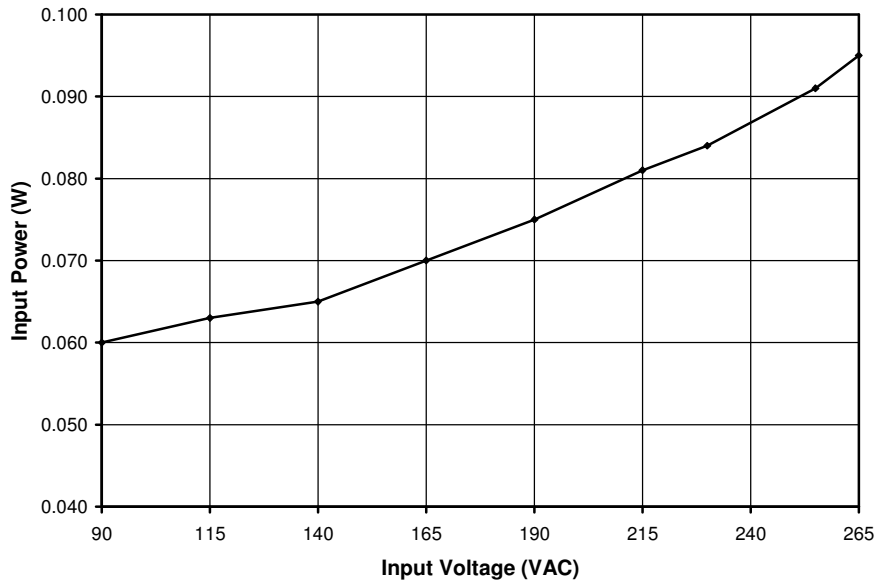


Figure 7 – Zero Load Input Power vs. Input Line Voltage, Room Temperature, 60 Hz.

9.4 Regulation

9.4.1 Load

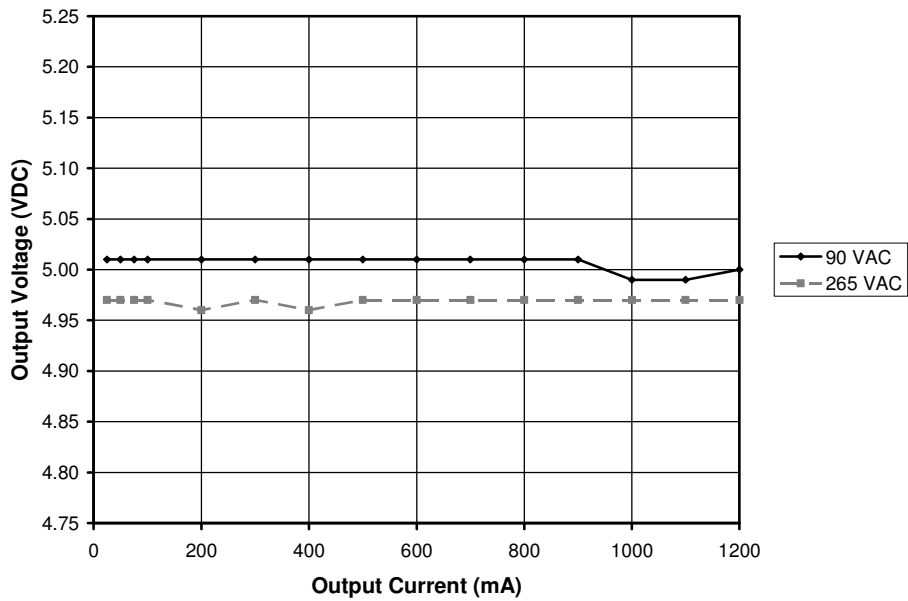


Figure 8 – Load Regulation, Room Temperature.



9.4.2 Line

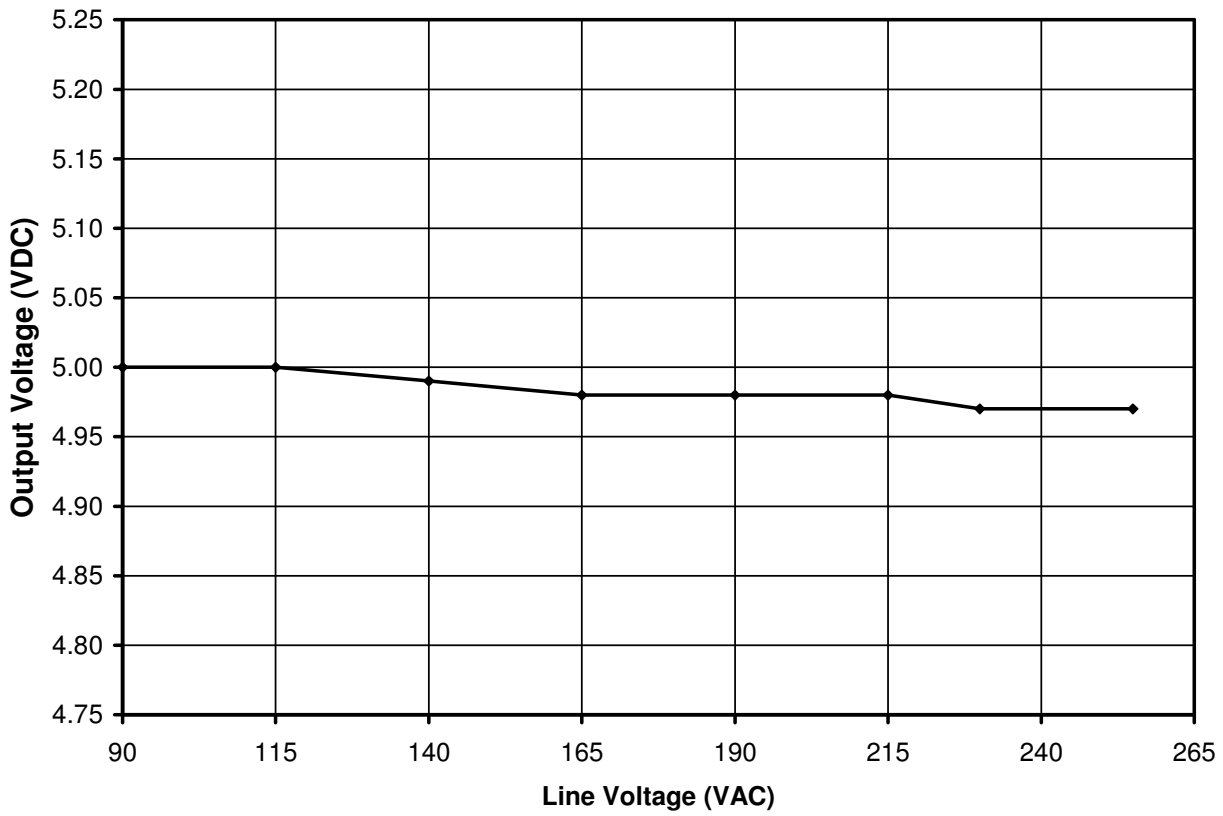


Figure 9 – Line Regulation, Room Temperature, Full Load.



10 Thermal Performance

The temperature of key components was measured to evaluate the thermal performance of the power supply. The power supply's thermal performance was measured at both low line (90 VAC) and high line (265 VAC) with line frequency of 50 Hz. The power supply was loaded to full load (1200 mA).

The power supply was placed inside a cardboard box, to minimize convective cooling, which was then placed in a thermal chamber. The ambient temperature within the box was monitored by a T-type thermocouple placed above the power supply. The temperature of the Power Integrations device, U1, was monitored by soldering a T-type thermocouple to one of the middle source pins, close to the case. The temperature of the transformer windings (T1) was monitored by gluing a T-type thermocouple, using thermally conductive epoxy, to the outermost layer of tape on the bottom of the transformer (closest to the PCB). Lastly the output diode's (D7) temperature was monitored by soldering a T-type thermocouple to the anode, near the case.

| Item | Temperature (°C) | |
|------------------------------|------------------|-----------------|
| | 90 VAC / 50 Hz | 265 VAC / 50 Hz |
| Ambient | 40 | 40 |
| Source Pin of PI Device (U1) | 87 | 69 |
| Transformer Windings (T1) | 70 | 67 |
| Anode of Output Diode (D7) | 76 | 76 |

Table 2 – Thermal Performance of the Power Supply.



11 Waveforms

11.1 Drain Voltage and Current, Normal Operation

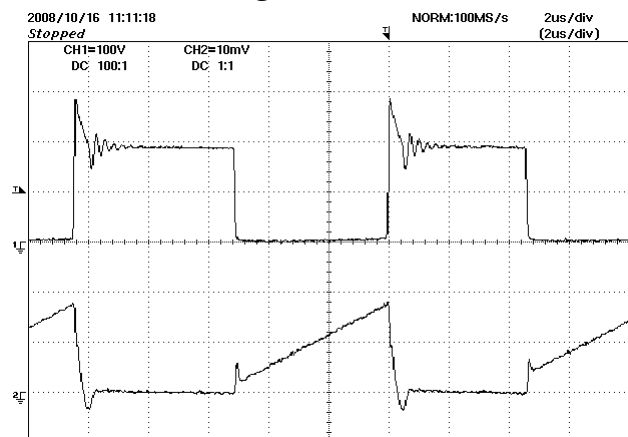


Figure 10 – 90 VAC, Full Load.
Upper: V_{DRAIN} , 100 V / div, 2 μ s / div.
Lower: I_{DRAIN} , 0.2 A / div.

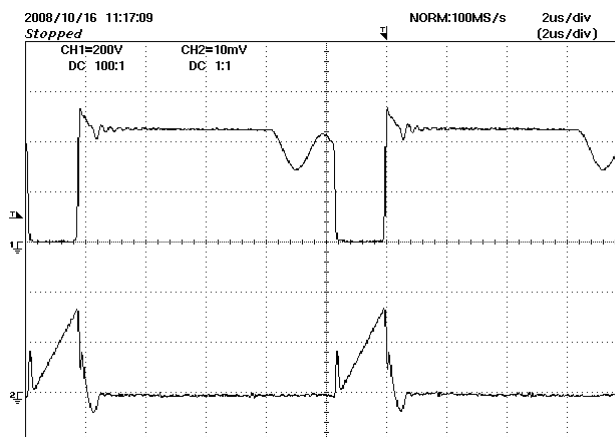


Figure 11 – 265 VAC, Full Load.
Upper: V_{DRAIN} , 200 V, 2 μ s / div.
Lower: I_{DRAIN} , 0.2 A / div.

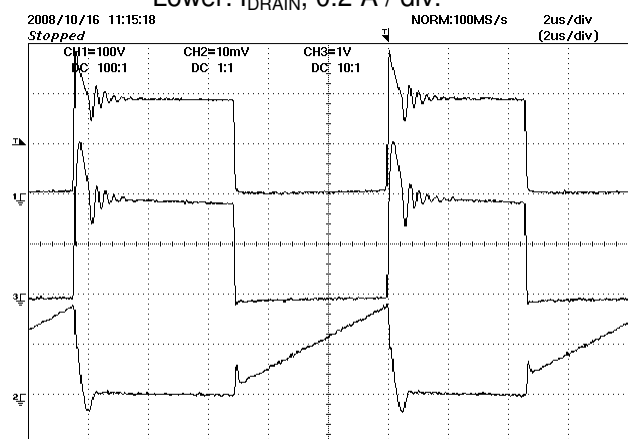


Figure 12 – 90 VAC, Full Load.
Upper: V_{DRAIN} , 100 V, 2 μ s / div.
Middle: V_{FB} , 1 V / div.
Lower: I_{DRAIN} , 0.2 A / div.

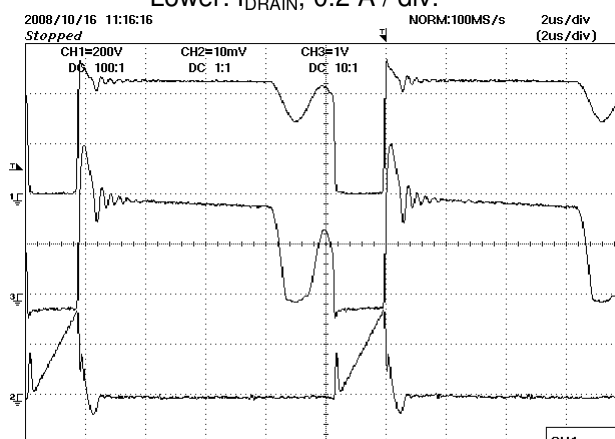


Figure 13 – 265 VAC, Full Load.
Upper: V_{DRAIN} , 200 V, 2 μ s / div.
Middle: V_{FB} , 1 V / div.
Lower: I_{DRAIN} , 0.2 A / div.

CH1
V/DIV
200 V



11.2 Output Voltage Start-up Profile

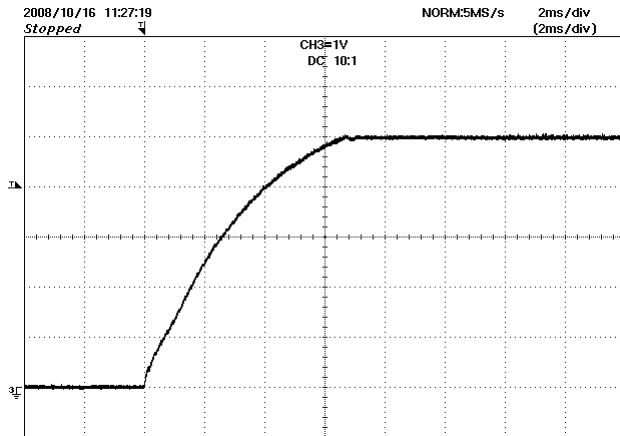


Figure 14 – Start-up Profile, 90 VAC, Full Load
1 V, 2 ms / div.

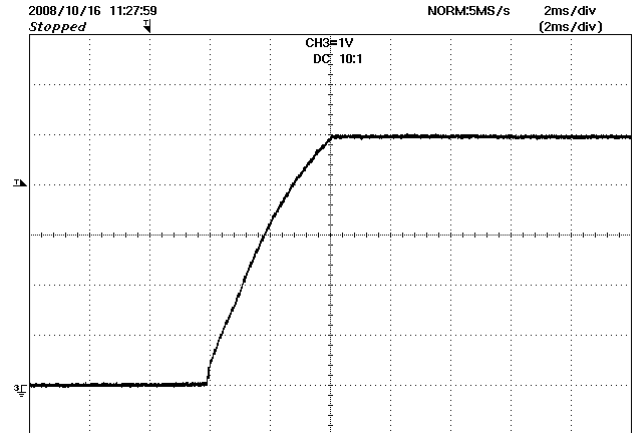


Figure 15 – Start-up Profile, 265 VAC, Full Load
1 V, 2 ms / div.

11.3 Drain Voltage and Current Start-up Profile

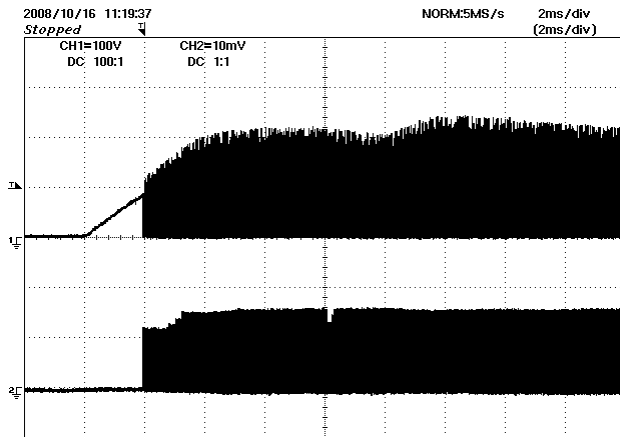


Figure 16 – 90 VAC Input and Full Load.
Upper: V_{DRAIN} , 100 V, 2 ms / div.
Lower: I_{DRAIN} , 0.2 A / div.

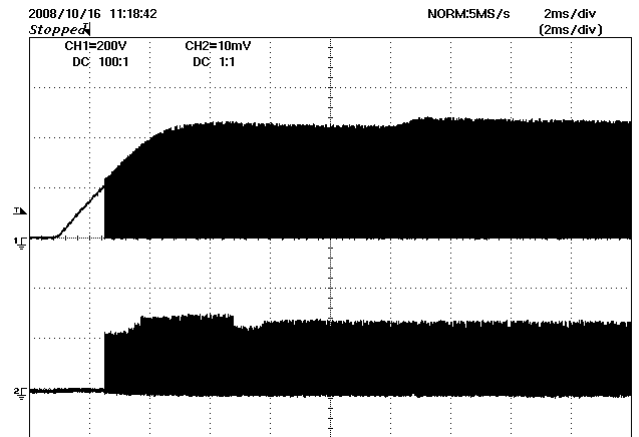


Figure 17 – 265 VAC Input and Full Load.
Upper: V_{DRAIN} , 200 V, 2 ms / div.
Lower: I_{DRAIN} , 0.2 A / div.

11.4 Load Transient Response (75% to 100% Load Step)

In the figures shown below, signal averaging was used to better enable viewing the load transient response. The oscilloscope was triggered using the load current step as a trigger source. Since the output switching and line frequency occur essentially at random with respect to the load transient, contributions to the output ripple from these sources will average out, leaving the contribution only from the load step response.



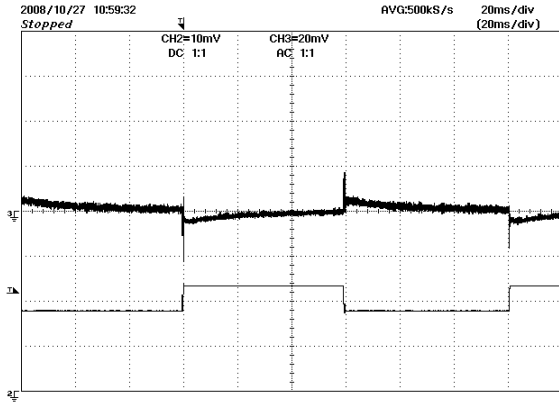


Figure 18 – Transient Response, 90 VAC
75-100-75% Load Step.
Upper: Output Voltage
10 mV, 10 ms / div.
Bottom: Load Current, 0.5 A / div.

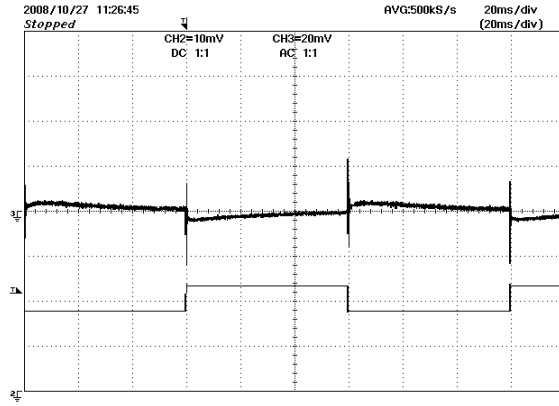


Figure 19 – Transient Response, 265 VAC
75-100-75% Load Step.
Upper: Load Current, 1 A / div.
Bottom: Output Voltage
50 mV, 2 ms / div.



11.5 Output Ripple Measurements

11.5.1 Ripple Measurement Technique

For DC output ripple measurements, use a modified oscilloscope test probe to reduce spurious signals. Details of the probe modification are provided in figures below.

Tie two capacitors in parallel across the probe tip of the 4987BA probe adapter. Use a 0.1 μF /50 V ceramic capacitor and a 1.0 μF /50 V aluminum-electrolytic capacitor. The aluminum-electrolytic capacitor is polarized, so always maintain proper polarity across DC outputs.

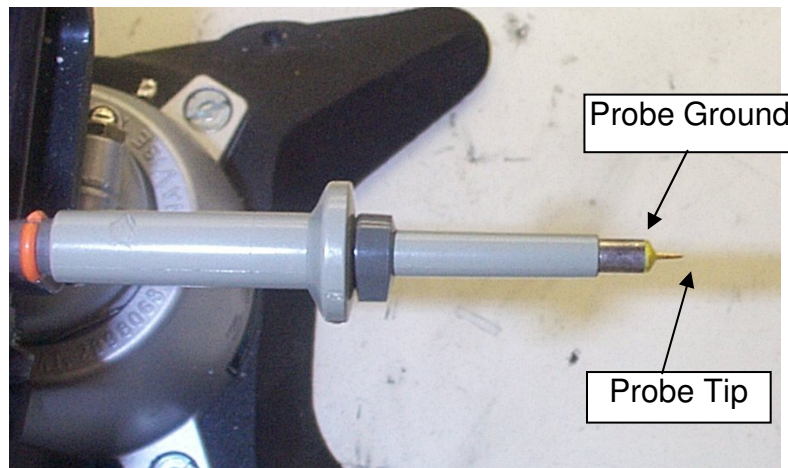


Figure 20 – Oscilloscope Probe Prepared for Ripple Measurement. (End Cap and Ground Lead Removed)

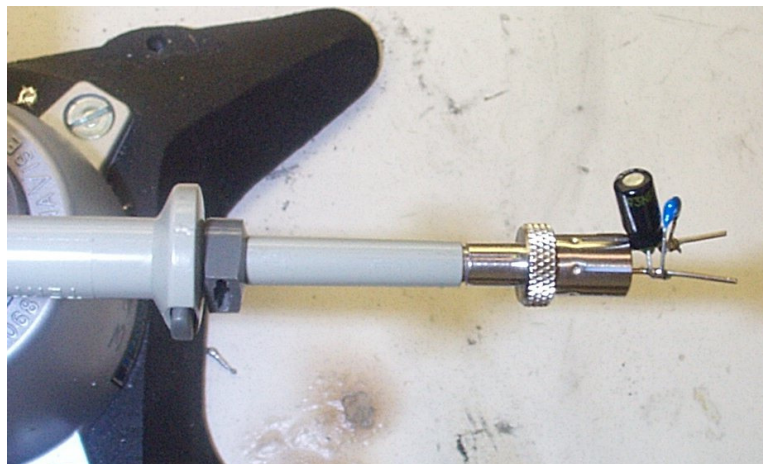


Figure 21 – Oscilloscope Probe with Probe Master (www.probemaster.com) 4987A BNC Adapter. (Modified with wires for ripple measurement, and two parallel decoupling capacitors added)

11.5.2 Measurement Results

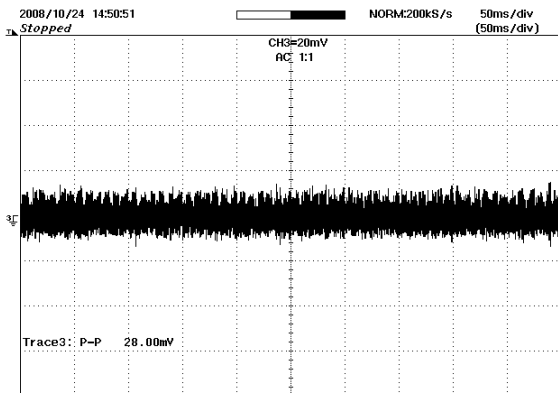


Figure 22 – Ripple, 90 VAC, Full Load.
50 ms, 20 mV / div.

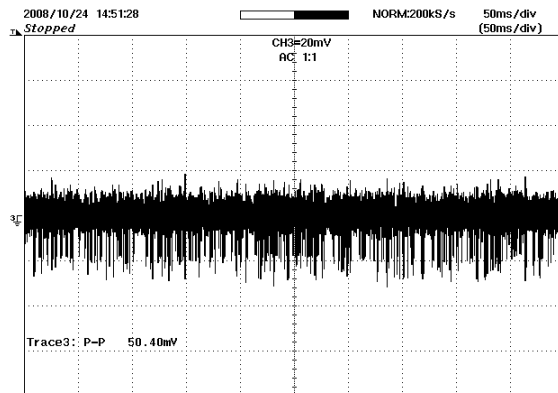


Figure 23 – 265 VAC, Full Load.
50 ms, 20 mV / div.



12 Line Surge

Differential input line 1.2/50 μ s surge testing was completed on a single test unit to IEC61000-4-5. Input voltage was set at 230 VAC / 60 Hz. Output was loaded at full load and operation was verified following each surge event.

| Surge Level (V) | Input Voltage (VAC) | Injection Location | Injection Phase (°) | Test Result (10 Strikes Pass/Fail) |
|-----------------|---------------------|------------------------|---------------------|------------------------------------|
| +500 | 230 | L to N (2 Ω) | 90 | Pass |
| +1000 | 230 | L to PE (12 Ω) | 90 | Pass |
| +1000 | 230 | L to N (2 Ω) | 90 | Pass |
| +2000 | 230 | L to PE (12 Ω) | 90 | Pass |

Unit passes under all test conditions.



13 Conducted EMI

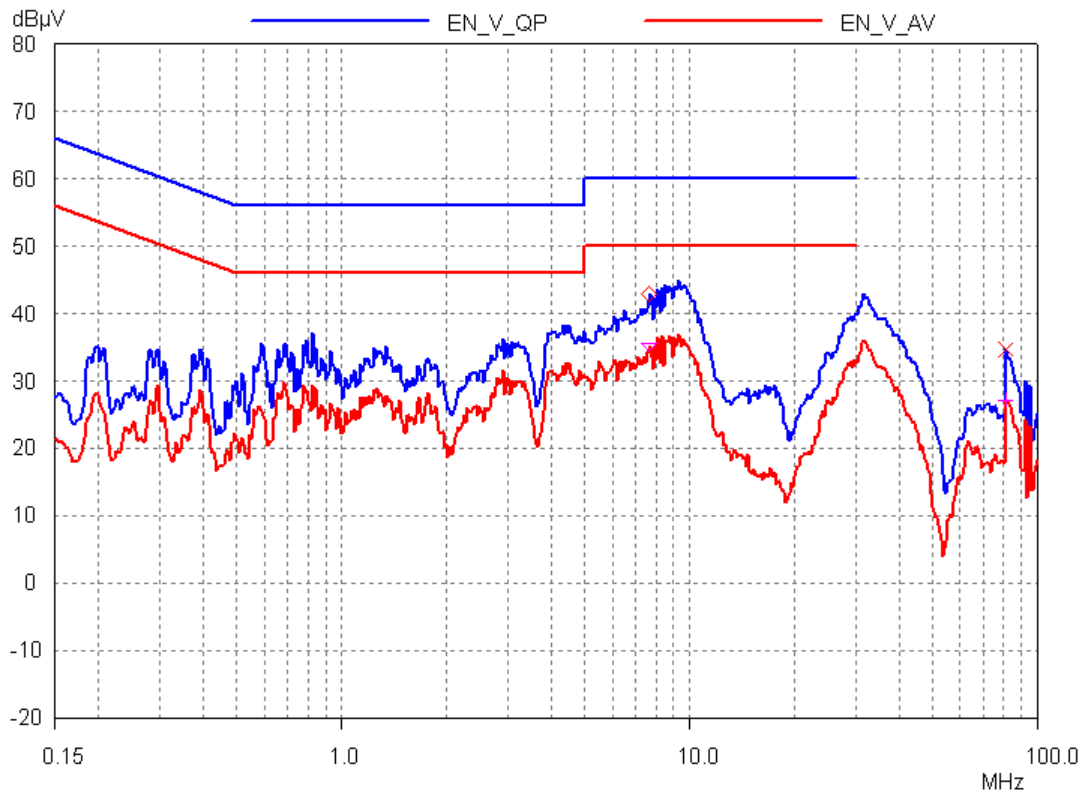


Figure 24 – Worst Case Conducted EMI, Maximum Steady State Load, 230 VAC, 60 Hz, and EN55022 B Limits, Artificial Hand.

14 Revision History

| Date | Author | Revision | Description & changes | Reviewed |
|-------------|---------------|-----------------|----------------------------------|-----------------|
| 11-Nov-08 | TS | 1.5 | Initial Release | SGK, PV |
| | | | | |
| | | | | |
| | | | | |



Notes



Notes



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