

# NC-Cap/PSR<sup>™</sup> (Primary Side Regulation) CV/CC Power Switch

#### **FEATURES**

- **♦** Built-in 600V Power MOSFET
- ◆ Proprietary NC-Cap/PSR<sup>TM</sup> (Primary Side Regulation) Control without External Compensation/Filtering Capacitor Needed
- ♦ Low Standby Power Under 50mW, Easily to Pass Energy Star EPS2.0
- ♦ ±5% Constant Current (CC) and Constant Voltage (CV) Regulation at Universal AC Input
- ◆ Proprietary Cable Voltage Drop Compensation in CV Mode
- ◆ Compensate for Line Voltage Variation
- ♦ Compensate for Transformer Inductance Tolerances
- Built-in Control Loop Compensation in CV Mode
- **♦** Pins Floating Protection
- ♦ PFM Control Eases EMI Design
- ◆ Cycle-by-Cycle Current Limiting
- Built-in Leading Edge Blanking (LEB)
- **♦** Built-in Soft Start
- Output Over Voltage Protection
- ♦ VDD UVLO, OVP & Clamp

#### **APPLICATIONS**

- Battery chargers for cellular phones, cordless phones, PDA, digital cameras, etc
- ◆ Replaces linear transformer and RCC SMPS
- ♦ Small power adapter
- ◆ AC/DC LED lighting

#### **GENERAL DESCRIPTION**

SF5928 is a high performance, highly integrated DCM (Discontinuous Conduction Mode) Primary Side Regulation (PSR) power switch for offline small power converter applications.

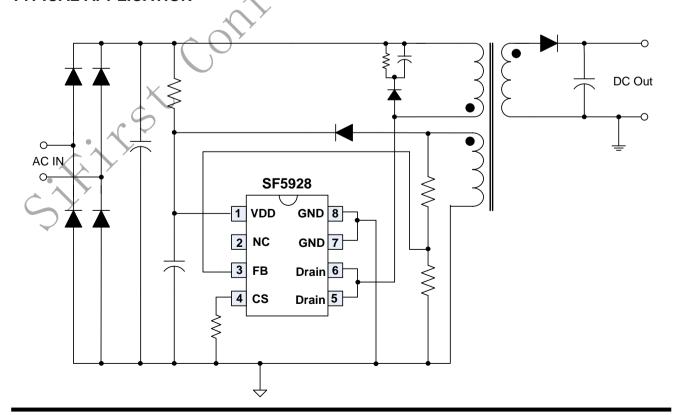
SF5928 uses Pulse Frequency Modulation (PFM) control to improve efficiency and eases system EMI design. The IC dramatically lowers system cost by eliminating the opto-coupler and secondary control circuits. It also can provide very tight output voltage regulation (CV), in addition to output current control (CC) ideal for charging applications.

SF5928 has built-in proprietary **NC-Cap/PSR<sup>TM</sup>** control for CV control, which eliminates external compensation or filtering capacitor. It has built-in cable drop compensation function, which can provide excellent CV performance. The IC also has built-in soft start function to soften the stress on the MOSFET during power on period. Under light load conditions, the IC decreases switching frequency to achieve excellent regulation and high efficiency, yet meets the requirement for no-load consumption less than 50mW.

SF5928 integrates functions and protections of Under Voltage Lockout (UVLO), VDD Over Voltage Protection (VDD OVP), Output Over Voltage Protection (Output OVP), Soft Start, Cycle-by-cycle Current Limiting (OCP), All Pins Floating Protection, VDD Clamping.

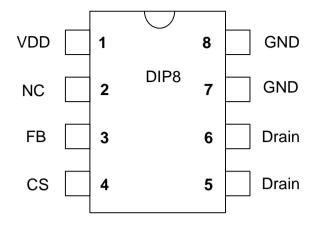
SF5928 is available in DIP8 package.

#### TYPICAL APPLICATION





### **Pin Configuration**



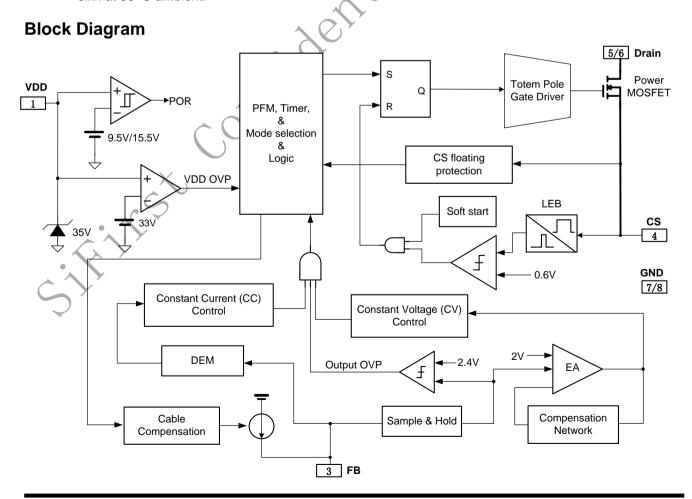
**Ordering Information** 

Part Number	Top Mark	Packa	age	Tape & Reel
SF5928DP	SF5928DP	DIP8	RoHs	

Output Power Table<sup>(1)</sup>

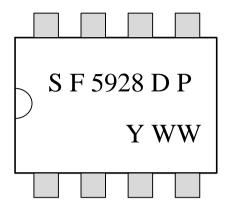
Part Number	230VAC ± 15% <sup>(2)</sup>	85-265VAC		
	Adapter <sup>(3)</sup>	Adapter <sup>(3)</sup>		
SF5928	18W ^	13W		

- Note 1. The Max. output power is limited by junction temperature
- Note 2. 230VAC or 100/115VAC with doublers
- **Note 3.** Typical continuous power in a non-ventilated enclosed adapter with sufficient drain pattern as a heat sink at 50 °C ambient.





### **Marking Information**



YWW: Year&Week code

# **Pin Description**

Pin Num	Pin Name	I/O	Description
1	VDD		IC power supply pin.
2	NC	-	No connection.
3	FB	I	System feedback pin. This control input regulates both the output voltage in CV mode and output current in CC mode based on the flyback voltage of the auxiliary winding.
4	CS	I	Current sense pin.
5-6	Drain	Р	High voltage power MOSFET drain connection.
7-8	GND	Р	Ground

Absolute Maximum Ratings (Note 4)

Parameter	Value	Unit
VDD DC Supply Voltage	35	V
VDD DC Clamp Current	10	mA
Drain pin	-0.3 to 600	V
FB, CS voltage range	-0.3 to 7	V
Package Thermal Resistance (DIP-8)	84	°C/W
Maximum Junction Temperature	150	°C
Operating Temperature Range	-40 to 85	°C
Storage Temperature Range	-55 to 150	°C
Lead Temperature (Soldering, 10sec.)	260	°C
ESD Capability, HBM (Human Body Model)	3	kV
ESD Capability, MM (Machine Model)	250	V

**Recommended Operation Conditions** (Note 5)

Parameter	Value	Unit
Supply Voltage, VDD	11 to 30	V
Operating Ambient Temperature	-40 to 85	°C



#### **ELECTRICAL CHARACTERISTICS**

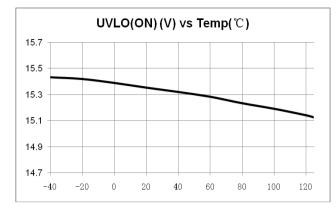
 $(T_A = 25^{\circ}C, VDD=16V, if not otherwise noted)$ 

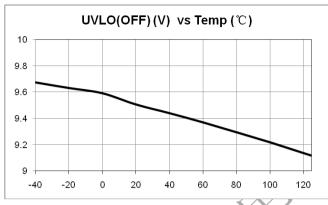
	=16V, if not otherwise noted		N/1:10	T	Max	11:::4
Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit
<u> </u>	e (VDD) Section		1	1	1	T
I_Startup	VDD Start up Current	VDD = UVLO(ON)-1V,		5	20	uA
		Measure current into VDD				
I_VDD_Op	Operation Current	V <sub>FB</sub> =3V, VDD=20V		1	2	mA
UVLO(ON)	VDD Under Voltage		14	15.5	16.5	V
	Lockout Exit (Startup)					
UVLO(OFF)	VDD Under Voltage		8.5	9.5	10.5	V
	Lockout Enter					/   7
VDD_OVP	VDD Over Voltage		31	33	35	V
	Protection trigger				4.4	/ )
V <sub>DD</sub> _Clamp	VDD Zener Clamp	$I(V_{DD}) = 10 \text{ mA}$	33	35	37	V
	Voltage					
T_Softstart	Soft Start Time <sup>(6)</sup>			2		mSec
Feedback Inpu	ut Section(FB Pin)			_ < 1	7	
V <sub>FB</sub> _EA_Ref	Internal Error		1.97	2.0	2.03	V
	Amplifier(EA)		K	X1///	,	
	reference input					
V <sub>FB</sub> _OVP	Output over voltage			2.4		V
	protection threshold					
V <sub>FB</sub> _DEM	Demagnetization	<u> </u>		0.1		V
	comparator threshold		$\sim$			
T <sub>min</sub> _OFF	Minimum OFF time	Note 6		2		uSec
T <sub>max</sub> OFF	Maximum OFF time			12		mSec
max_		•				
$T_{CC}/T_{DEM}$	Ratio between	Note 6		2		
	switching period in					
	CC mode and					
	demagnetization time	A 0 3				
I <sub>Cable</sub> _max	Max Cable	X		40		uA
	compensation current	O*				
	Input Section (CS Pin)	<u></u>				
T_blanking	CS Input Leading			500		nSec
	Edge Blanking Time					
Vth_OC	Current limiting		588	600	612	mV
	threshold					
T <sub>D</sub> OC	Over Current			100		nSec
	Detection and Control					
	Delay					
Power MOSFE	T Section <sup>(7)</sup>		ı	1	1	ı
BVdss	Power MOSFET		600			V
	Drain Source					
	Breakdown Voltage					
Rdson	Static Drain-Source On	I(Drain)=1A		3.8	4.7	Ω
~ ~ ·	Resistance					
Idss	Zero Gate Voltage				1	uA
$\overline{}$	Drain Current					
Td <sub>(on)</sub>	Turn-on delay time			9		ns
Td <sub>(off)</sub>	Turn-off delay time			24		ns

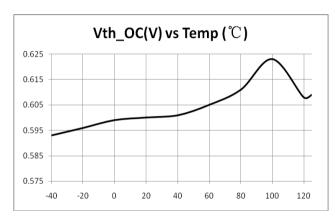
- Note 4. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.
- **Note 5.** The device is not guaranteed to function outside its operating conditions.
- Note 6. Guaranteed by design.
- Note 7. These parameters, although guaranteed, are not 100% tested in production

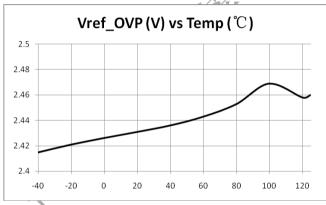


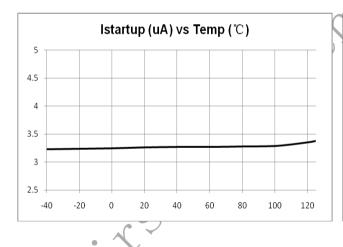
### **CHARACTERIZATION PLOTS**

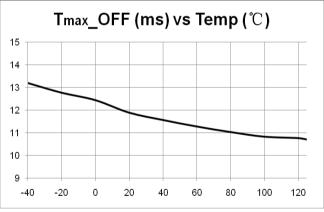


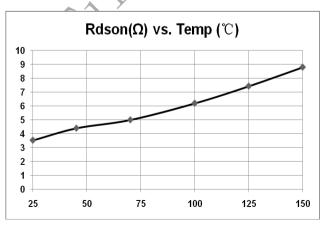














### **OPERATION DESCRIPTION**

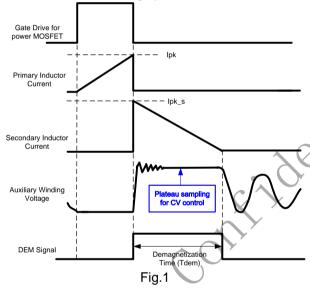
SF5928 is a high performance, highly integrated DCM (Discontinuous Conduction Mode) Primary Side Regulation (PSR) power switch working in PFM (Pulse Frequency Modulation) mode. The built-in power MOSFET and high precision CV/CC control with high level protection features make it very suitable for offline small power converter applications.

### **♦ PSR Technology Introduction**

Assuming the system works in DCM mode, the power transfer function is given by

$$P = \frac{\eta}{2} \times L_m \times I_{pk}^2 \times f_S = V_o \times I_o \quad \text{ (Eq.1)}$$

In the equation above, P is output power, Vo and Io are system output voltage and current respectively, n is system power transfer efficiency, Lm is transformer primary inductance, fs is system switching frequency, lpk is primary peak current in a switching cycle. The following figure illustrates the waveform in a switching cycle.



In the figure shown above, the IC generates a demagnetization signal (DEM) in each switching through auxiliary winding. demagnetization time for CV/CC control. In DCM mode, Tdem can be expressed as;

$$\frac{V_o}{L_m} \times T_{dem} = \frac{N_S}{N_P} \times I_{pk}$$
 (Eq.2)

In Eq.2, Np and Ns are primary and secondary winding turns respectively.

Combined with Eq.1 and Eq. 2, the average output current can be expressed as:

$$I_o = \frac{\eta}{2} \times I_{pk} \times \frac{N_P}{N_S} \times f_S \times T_{dem}$$
 (Eq.3)

#### **CC (Constant Current) Control Scheme**

From Eq.3, it can be easily seen that there are two ways to implement CC control: one is PFM (Pulse Frequency Modulation), the control scheme is to keep lpk to be constant, let the product of Ts and Tdem (fs\*Tdem) to be a constant. In this way, lo will be a value independent to the variation of Vo, Lm, and line input voltage. Another realization method is PWM duty control, the control scheme is to keep fs to be constant, let the product of Tdem and lpk (Tdem\*lpk) to be a constant, in another words, by modulating system duty cycle to realize a constant lo independent to the variation of No. Lm and line voltages.

SF5928 adopts PFM for CC control, the product of Ts and Tdem is given by

$$f_S \times T_{dem} = 0.5$$
 Eq.4)

## CV (Constant Voltage) Control Scheme

CV control should sample the plateau of auxiliary winding voltage in flyback phase, as shown in Fig.1 The CV control has many implementations, for example, PWM, or PFM, or a combination of both one. In SF5928, the PFM control is adopted for CV control.

#### NC-Cap/PSR™ **Eliminates External Compensation/Filtering Capacitor**

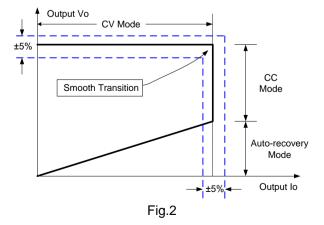
SF5928 uses a proprietary control to eliminate external compensation capacitor, simplify system design and lower system cost.

#### PFM Control Eases System EMI Design

As mentioned above, the CC/CV control in SF5928 uses PFM control, which will eases system EMI design greatly. Since PFM control is a frequency variation system with inherent frequency shuffling function, it will have superior EMI performance than that of PWM control.

#### Precision CV/CC Performance with Smooth Transition between CV and CC

In SF5928, the parameters is trimmed to tight range, which makes the system CC/CV to have less than 5% variation.



In SF5928, the IC has specially designed to be able to smoothly transit between CC and CV mode.



When the output voltage is too low, the IC will enters into auto-recovery mode, as shown in Fig. 2

#### ◆ Startup Current and Startup Control

Startup current of SF5928 is designed to be very low (typically 5uA) so that VDD could be charged up above UVLO(ON) threshold level and device starts up quickly. A large value startup resistor can therefore be used to minimize the power loss yet reliable startup in application.

#### Operating Current

The operating current in SF5928 is as small as 1mA (typical). The small operating current results in higher efficiency and reduces the VDD hold-up capacitance requirement. Once SF5928 enters very low frequency PFM mode, the operating frequency is reduced to less than 0.3mA, assisting the power supply in meeting power conservation requirements.

#### ♦ Soft Start

SF5928 features an internal 2ms (typical) soft start that slowly increases the threshold of cycle-bycycle current limiting comparator during startup sequence. It reduces the stress on the secondary diode during startup. Every startup process is followed by a soft start activation.

#### **Proprietary** Voltage Cable Drop **Compensation in CV Mode**

When it comes to cellular phone charger applications, the battery is located at the end of cable, which causes typically several percentage of voltage drop on the actual battery voltage. SF5928 has a proprietary built-in cable voltage drop compensation block which can provide a constant output voltage at the end of the cable over the entire load range in CV mode.

### Leading Edge Blanking (LEB)

In SF5928, an internal leading edge blanking circuit is built in. During this blanking period (500ns, cycle-by-cycle current typical), the limiting comparator is disabled and cannot switch off the gate driver.

#### Minimum and Maximum OFF Time

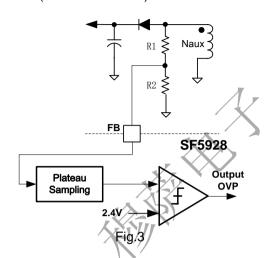
In SF5928, a minimum OFF time (typically 2us) is implemented to suppress ringing when GATE is off. The maximum OFF time in SF5928 is typically 10ms, which provides a large range for frequency reduction. In this way, a low standby power of 50mW can be achieved.

## ◆ Pins Floating Protection

In SF5928, if pin floating situation occurs, the IC is designed to have no damage to system.

#### **Output OVP(Over Voltage Protection)**

In SF5928, the output OVP is integrated by plateau sampling the auxiliary winding in flyback phase. The threshold voltage for output OVP is 2.4V, as shown in Fig.3. Output OVP is auto-recovery mode protection (mentioned below).



#### VDD OVP(Over Voltage Protection)

OVP (Over Voltage Protection) implemented in SF5928 and it is a protection of auto-recovery mode.

### Auto Recovery Mode Protection

As shown in Fig.4, once a fault condition is detected, switching will stop. This will cause VDD to fall because no power is delivered form the auxiliary winding. When VDD falls to UVLO(off) (typical 9V), the protection is reset and the operating current reduces to the startup current, which causes VDD to rise, as shown in Fig.4. However, if the fault still exists, the system will experience the above mentioned process. If the fault has gone, the system resumes normal operation. In this manner, the auto restart can alternatively enable and disable the switching until the fault condition is disappeared.

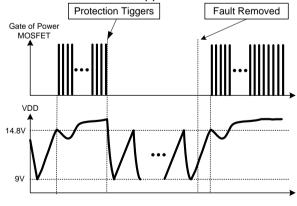


Fig.4

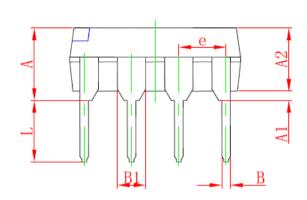
#### Soft Gate Driver for Power MOSFET

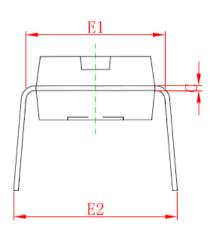
The driving stage of SF5928 is a soft totem-pole gate driver to minimize EMI. Cross conduction has been avoided to minimize heat dissipation, increase efficiency, and enhance reliability.

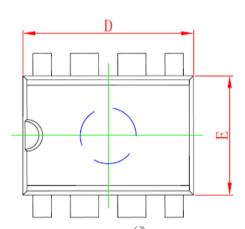


## **PACKAGE MECHANICAL DATA**

## **DIP8 PACKAGE OUTLINE DIMENSIONS**







Symbol	Dimensions In Millimeters		Dimensions In Inches		
Syllibol	Min	Max	Min	Max	
Α	3.710	5.334	0.146	0.210	
A1 X	0.381		0.015		
A2	3.175	3.600	0.125	0.142	
B	0.350	0.650	0.014	0.026	
B1	1.524 (BSC)		0.06 (BSC)		
C	0.200	0.360	0.008	0.014	
D	9.000	10.160	0.354	0.400	
E	6.200	6.600	0.244	0.260	
E1	7.320	7.920	0.288	0.312	
е	2.540 (BSC)		0.1 (BSC)		
L	2.921	3.810	0.115	0.150	
E2	8.200	9.525	0.323	0.375	



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