

Single-Stage Primary-Regulation Controller for PFC and LED Dimmable Driving

FEATURES

Wonderful Compatibility and Performance

- Compatible with TRIAC Dimming
- Primary Side Regulation (PSR)
- Constant Current output (+/- 1.5%)
- Universal Input Voltage Range

Energy Efficient

- Boundary Conduction Mode (BCM) with Valley Switching reduces EMI and enhances Efficiency
- Power Factor Correction (PF>95%)
- High Efficiency (>85% achievable)
- Low Startup Current: 1.5µA
- Low Operating Current: 1.9mA
- Low Quiescent Current: 1.2mA

Advanced Protection and Safety Features

- Open-LED Protection
- Short-LED Protection
- Cycle-by-Cycle Current Limit
- Over-Temperature Protection

Package

SOP-8 Package Available

DESCRIPTION

The primary-side-control offline LED lighting controller, SE8320, provides accurate output current, high power factor and low total harmonic distortion. High efficiency is achieved by low operating current and valley turn-on of the primary MOSFET. Constant on-time control is utilized for a better PFC performance.

TRIAC dimming is smoothly managed without flicker. Either primary side regulation or secondary side regulation, dimming model or no dimming model can be adopted by different set of DIM and FB pins.

The multi-protection of SE8320, including open-LED protection, short-LED protection, cycle-tocycle current limit, VCC UVLO and overtemperature protection, can greatly enhance the system reliability. Current-limit threshold is adjusted automatically in a short-LED condition in order that the output current is minimized.

APPLICATIONS

- Isolated LED Driver with/without TRIAC dimming
- Non-isolated LED Driver with/without TRIAC dimming
- Secondary Side Regulation LED Driver with/without TRIAC dimming

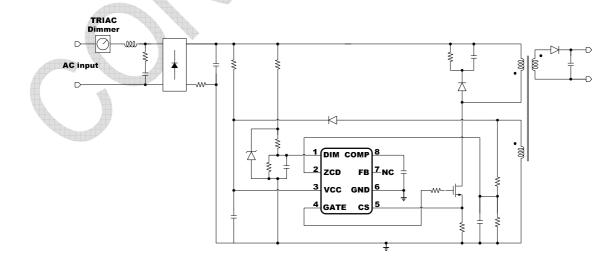


Figure 1. Typical Application



FUNCTIONAL BLOCK DIAGRAM

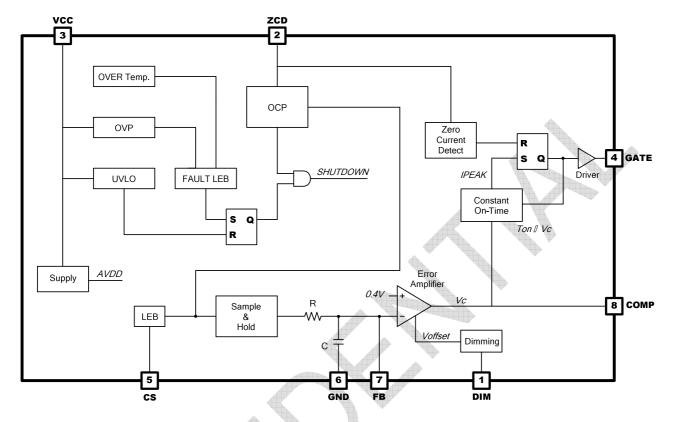


Figure 2. Functional Block Diagram

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PIN FUNCTIONS

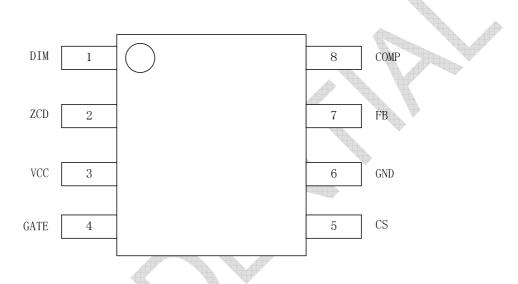
| Pin # | Name | Description |
|-------|------|--|
| 1 | DIM | Dimming pin. This pin detects the TRIAC dimming phase to control the LED lighting. This pin can be NC if TRIAC dimming is not wanted. |
| 2 | ZCD | Zero current detection pin. Connect this pin through a resistor divider from the auxiliary winding to GND in order to detect the inductor current zero crossing point. This pin also detects the output voltage for OCP. |
| 3 | vcc | Power supply pin. This pin supply power both for IC operating current and GATE driving current. |
| 4 | GATE | Gate drive output pin. The totem pole output stage is able to drive high power MOSFET. |
| 5 | CS | Current sense pin. The MOSFET current is sensed via a resistor for constant current regulation and cycle-to-cycle current limit. |
| 6 | GND | Ground. |
| 7 | FB | Feedback Pin. This pin can be NC if primary side regulation is utilized. |
| 8 | COMP | Loop Compensation pin. Connect a compensation network to stabilize the loop. |



ORDERING INFORMATION

| ORDERING NUMBER | PINS | PACKAGE | |
|-----------------|------|---------|--|
| SE8320-SO-L | 8 | SOP | |

PACKAGE REFERENCE



ABSOLUTE MAXIMUM RATINGS

| Item | Symbol | Rating | Unit |
|-----------------------------------|-------------------|--------------|------|
| Power Supply Voltage | V _{CC} | <30 | V |
| ZCD Pin Input Voltage | V _{ZCD} | -0.3 to 7.0 | V |
| CS Pin Input Voltage | V _{cs} | -0.3 to 7.0 | V |
| DIM Pin Input Voltage | V _{DIM} | -0.3 to 7.0 | V |
| COMP Pin Input Voltage | V _{COMP} | -0.3 to 7.0 | V |
| GATE Pin Input Voltage | V _{GATE} | -0.3 to 30.0 | V |
| Maximum Junction Temperature | TJ | <150 | °C |
| Storage Temperature | T _{STG} | -55 to 150 | °C |
| Lead Temperature (Soldering, 10s) | T _{Lead} | <260 | °C |



ELECTRICAL CHARACTERISTICS

V_{cc}=20V, T_A=+27°C, unless otherwise noted.

| Symbol | Parameter | Condition | Min. | Тур. | Max. | Unit |
|-----------------------|---|--|-------|-------|-------|------|
| SUPPLY | SECTION (T _A = -40°C to 128 | 5°C) | | | | |
| V _{CC} | Operating Range | | 9 | | 24 | V |
| V _{CC-ON} | Turn-On Threshold Voltage | | 14 | 16 | 18 | V |
| $V_{\text{CC-OFF}}$ | Turn-Off Threshold Voltage | | 7 | 8 | 9 | V |
| V _{CC-HYS} | V _{CC} Hysteretic Voltage | | 7 | 8 | 9 | V |
| I _{CC} | Operating Current | Switch Period = 15µs | 1.6 | 1.9 | 2.4 | mA |
| Ι _Q | Quiescent Current | No switch | 1 | 1.2 | 1.55 | mA |
| I _{ST} | Startup Current | V _{CC} = V _{CC-ON} - 0.16V | 1.4 | 1.5 | 1.6 | μA |
| V _{OVP} | V _{CC} Over-Voltage-Protection | | 24 | 25 | 25.5 | V |
| V _{OVP-HYS} | OVP Hysteresis | | 0.9 | 1 | 1.05 | V |
| CONSTA | NT ON-TIME SECTION (T, | a = -40°C to 125°C) | | | | |
| T _{ON-MIN} | Minimum On Time | | 0.7 | 1 | 1.5 | μs |
| T _{ON-MAX} | Maximum On Time | | 20 | 24 | 27 | μs |
| ERROR A | MPPLIFIER SECTION | | | | | |
| G _M | Transconductance | | | 120 | | μA/V |
| A _{EA} | Voltage Gain | | | 9000 | | V/V |
| V_{FB} | Feedback Voltage | -40°C < T _A < 125°C | 0.384 | 0.398 | 0.405 | V |
| V _{COMP} | COMP Voltage Range | Ψ. | 0.9 | | 4 | V |
| I _{C-SOURCE} | Max Source Current | | 35 | 48 | 65 | μA |
| I _{C-SINK} | Max Sink Current | | -220 | -294 | -362 | μA |
| GATE DR | IVER SECTION (T _A = -40°C | C to 125°C) | | | | |
| V _{CLAMP} | Output Clamp Voltage | | 11 | 12 | 14 | V |
| I _{G-SOURCE} | Max Source Current | | 0.97 | 1.2 | 1.4 | А |
| I _{G-SINK} | Max Sink Current | | -1.2 | -1.7 | -2.1 | Α |
| ZERO CU | RRENT DETECTOR SECTIO | DN . | | | | |
| V _{ZCD} | ZCD Threshold | | | 0.4 | | V |
| V _{ZCD_HYS} | ZCD Hysteresis | | | 0.5 | | V |
| T _{OFF-MIN} | Minimum Off Time | | | 3.6 | | μs |
| AUTO ST | ART SECTION | | | | | |
| T _{START} | Auto Start Time | -40°C < T _A < 125°C | 100 | 140 | 190 | μs |

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ELECTRICAL CHARACTERISTICS

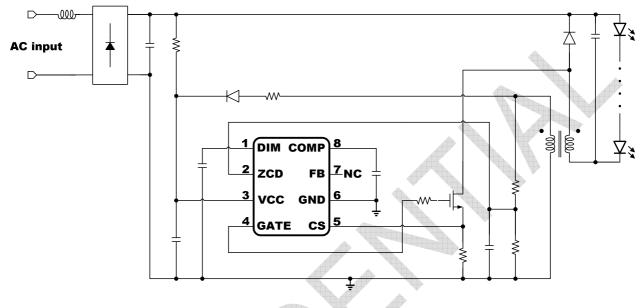
V_{cc}=20V, T_A=+27°C, unless otherwise noted.

| Symbol | Parameter | Condition | Min. | Тур. | Max. | Unit | |
|----------------------|---|-----------|------|------|------|------|--|
| OVER-CU | OVER-CURRENT PROTECTION SECTION | | | | | | |
| V _{OCP-H} | CS High Threshold Voltage for OCP | | | 3 | ~ | V | |
| V _{OCP-L} | CS Low Threshold Voltage for OCP | | | 1 | | V | |
| V _{HOCP-EN} | ZCD Threshold Voltage to Enable High OCP level | | | 0.9 | X | V | |
| $V_{LOCP-EN}$ | ZCD Threshold Voltage to Enable Low OCP level | | | 0.6 | | V | |
| T _{LEB} | CS Sampling Leading-Edge Blanking Time | 4 | | 215 | | ns | |
| OVER TE | MPERATURE PROTECTION | | | - | | | |
| T _{OTP} | Over Temperature Protection | | 140 | 150 | 160 | °C | |
| T _{OTP-HYS} | OTP Hysteresis | | 22 | 25 | 28 | °C | |
| DIMMING | DIMMING SECTION | | | | | | |
| V _{DH} | Maximum V _{DIM} at Low Dimming Angle Range | | | 3.5 | | V | |
| V _{DL} | Maximum V _{DIM} at High Dimming Angle Range | | | 2.5 | | V | |
| DS _{DH} | V_{DIM} vs. $V_{\text{FB-OFFSET}}$ Slope at High Dimming Angle Range | - | | 0.23 | | V/V | |
| DS _{DL} | V _{DIM} vs. V _{FB-OFFSET} Slope at Low Dimming Angle Range | | | 0.08 | | V/V | |

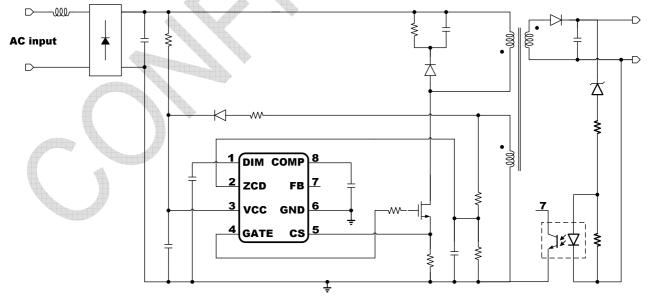




Non-Isolated Application



Secondary Side Regulation Application

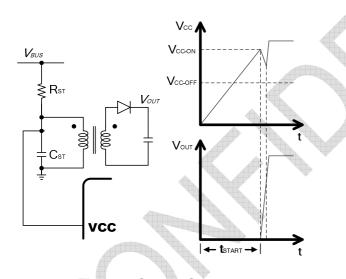




SE8320 is a TRIAC dimmable single stage Flyback and PFC controller for LED lighting applications. Primary side control is applied so that the system is simplified, and high power factor is achieved by constant on-time model. Boundary Conduction Mode (BCM) with valley switching improves efficiency and FMI performance. The multi-protection function stabilizes system and protects external components.

Startup

The capacitor C_{ST} across V_{BUS} and GND is charged by BUS through a start up resistor R_{ST} once BUS is powered on. After V_{CC} rises up to V_{CC-ON} , the internal blocks start to work and the gate driver begins to switch. Then V_{CC} will be pulled down by internal consumption until the power supply is taken over by the auxiliary winding.





Select C_{ST} for an ideal t_{ST} :

For a more stable V_{CC} , a bigger C_{ST} is needed, and R_{ST} should be decreased in order that t_{ST} is not changed. Obviously, the low I_{ST} of SE8320 makes it easier to design R_{ST} .

Boundary Conduction Mode Operation

Boundary Conduction Mode (BCM) and valley switching provides low turn-on switching losses.

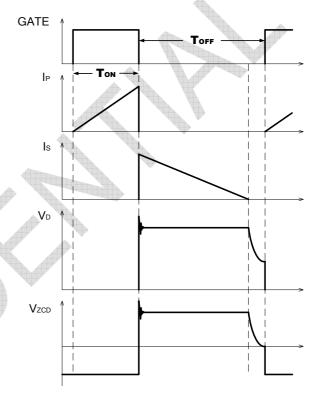


Figure 4. Boundary Conduction Mode

The voltage across drain and source of the external MOSFET is detected by the ZCD pin. The current of the inductor begins to decrease linearly as soon as the external MOSFET is turned off. When the current falls to zero, the MOSFET Drain-Source Voltage decreases, which is also detected by the ZCD pin through a resistor divider. The external MOSFET would be turned on by a turn on signal sent by the Zero Current Detector once the ZCD voltage is lower than 0.4V.

Primary Side Constant Current Control

The output mean current can be represented as

I_{OUT} = **MBS**



 V_{FB} —The feedback reference voltage, typically 0.4V.

N—The winding turns ratio of primary side to secondary side of the transformer.

 $R_{\text{CS}}\text{---}\text{The}$ current sense resistor connected between the CS pin and GND.



FUNCTIONAL DESCRIPTION

Power Factor Correction

Internal Constant On-Time Block affords a constant gate on-time T_{ON} , which is in proportion to COMP potential. The peak current of the primary side winding is

Ip = IN MUS

 L_P —The primary inductance. As L_P and T_{ON} is constant, I_P is accordingly in proportion to V_{BUS} (as shown in Figure 5).

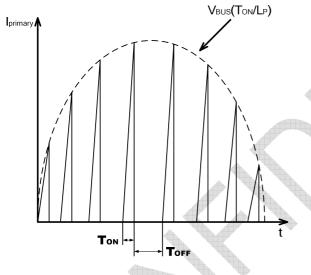


Figure 5. Power Factor Correction

As a result, the input current I_{IN} of the system follows the input voltage V_{IN} , and the Power Factor is improved.

TRIAC Dimming Control

SE8320 can implement TRIAC dimming function by internal dimming block and external passive components. The Zener diode clamp voltage is divided by two resistors, as shown in Figure 1, to obtain $V_{\text{DIM-MAX}}$ in Figure 6. Dimming angle θ_{DIM} is sensed by the Zener diode and consequently changes V_{DIM} .

When V_{DIM} is lower than V_{DH} , FB_{offset} increases as V_{DIM} decreases. This offset voltage eventually reduces the LED brightness.

The external dimming components can be replaced by a 1nF capacitor to GND to disable the dimming function. An internal current source would charge this capacitor to over V_{DH} .

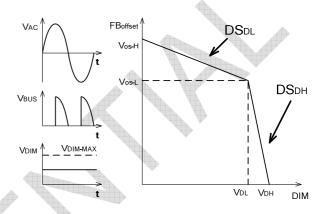


Figure 6. Dimming Function

VCC Under-Voltage Lockout (UVLO)

When the V_{CC} voltage drops below V_{CC-OFF} (typically 8V), the whole chip shuts down, and GATE switching stops. The system would not work again until VCC capacitor is charged to V_{CC-ON} through the external startup resistor.

Auto Start

A auto start block is integrated in SE8320 to avoid unnecessary shut down. The auto start block starts timing as soon as the external MOSFET turns on in every period. If the ZCD pin fails to send a turn on signal after T_{START} (typically 140µs), the auto start block would turn on GATE automatically.

When the GATE is turned on, auto start timing stops.

Minimum Off Time

To limit the maximum switching frequency and to obtain a better EMI performance, a internal block is integrated to limit the minimum GATE off time. The external MOSFET cannot turn on again in less than 3.6µs after its latest turning off.

Leading Edge Blanking (LEB)

An internal leading edge blanking (LEB) block is employed in order to avoid the premature termination of the GATE switching pulse due to



the peak voltage of the CS pin caused by the parasitic capacitor discharging when the external

MOSFET turns on. V_{CS} sampling is disabled during the LEB time.



FUNCTIONAL DESCRIPTION

Open-LED Protection

When the load of the system is open, the output capacitor is charged up rapidly. Consequently, V_{CC} rises up quickly since the VCC capacitor is charged by the auxiliary winding and its voltage reflects the output voltage.

The whole chip enters fault state, in which GATE will never be turned on, as soon as the voltage of the VCC pin rises up to V_{OVP} (typically 25V). As the external MOSFET is always off in the fault state, the auxiliary winding cannot support the V_{CC} consumption, and V_{CC} will drop. This fault state will last until UVLO.

Short-LED Protection

The voltage of the ZCD pin reflects the output voltage when the switch is on. The short-LED situation can be detected through the ZCD pin and the OCP threshold would be changed in that case.

If the ZCD voltage in GATE on period is lower than 0.6V, the cycle-by-cycle current limit would be 1V. The current limit threshold voltage becomes up to 3V when the ZCD voltage in GATE on period is higher than 0.9V. Figure 7 shows this function.

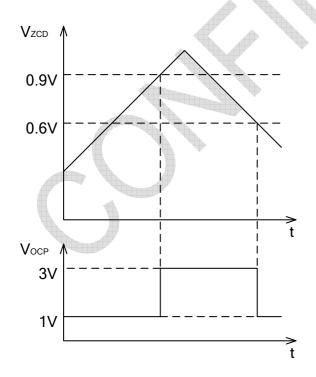


Figure 6. OCP Function

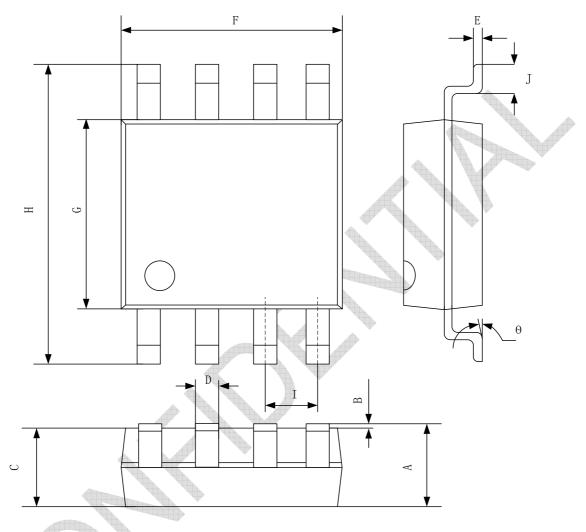
A lower OCP voltage can protect the external MOSFET as well as reduce power dissipation. As output voltage is low in the short-LED condition, the auxiliary winding cannot charge the VCC capacitor, and V_{CC} will consequently drop until UVLO.

Over-Temperature Protection (OTP)

When the junction temperature is above 150°C, the chip enters fault state, and GATE driver is shut down. The UVLO signal can relive the system of fault state if the junction temperature drops below 125°C.

PACKAGE INFORMATION

SMET



| Symbol | Dimensions in Millimeters | | Dimensions in Inches | |
|--------|---------------------------|------------|----------------------|------------|
| | Min | Мах | Min | Max |
| A | 1.350 | 1.750 | 0.053 | 0.069 |
| В | 0.100 | 0.250 | 0.004 | 0.010 |
| C | 1.350 | 1.550 | 0.053 | 0.061 |
| D | 0.330 | 0.510 | 0.013 | 0.020 |
| É | 0.170 | 0.250 | 0.006 | 0.010 |
| F | 4.700 | 5.100 | 0.185 | 0.200 |
| G | 3.800 | 4.000 | 0.150 | 0.157 |
| Н | 5.800 | 6.200 | 0.228 | 0.244 |
| I | 1.270 | (BSC) | 0.050 | (BSC) |
| J | 0.400 | 1.270 | 0.400 | 1.270 |
| θ | 0° | 8 ° | 0° | 8 ° |



SOLDERING INDICATION

This section gives a very brief insight to a complex technology. There is no soldering method that is ideal for all surface mount IC packages. Wave soldering can still be used for certain surface mount ICs, but it is not suitable for fine pitch SMDs. In these situations reflow soldering is recommended.

Reflow Soldering

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stenciling or pressure-syringe dispensing before package placement.

Several methods exist for reflowing; for example, convection or convection/infrared heating in a conveyor type oven. Throughput times (preheating, soldering and cooling) vary between 100 and 200 seconds depending on heating method.

Typical reflow peak temperatures range from 215 to 250°C. The top-surface temperature of the packages should preferable be kept below 220°C for thick/large packages, and below 235°C for small/thin packages.

Wave Soldering

Conventional single wave soldering is not recommended for surface mount devices (SMDs) or printed-circuit boards with a high component density, as solder bridging and non-wetting can present major problems.

To overcome these problems the double-wave soldering method was specifically developed.

If wave soldering is used, the following conditions must be observed for optimal results:

- Use a double-wave soldering method comprising a turbulent wave with high upward pressure followed by a smooth laminar wave.
- For packages with leads on two sides and a pitch:

- larger than or equal to 1.27 mm, the footprint longitudinal axis is preferred to be parallel to the transport direction of the printed-circuit board;

- smaller than 1.27 mm, the footprint longitudinal axis must be parallel to the transport direction of the printed-circuit board.

The footprint must incorporate solder thieves at the downstream end.

• For packages with leads on four sides, the footprint must be placed at a 45° angle to the transport direction of the printed-circuit board. The footprint must incorporate solder thieves downstream and at the side corners.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Typical dwell time is 4 seconds at 250°C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.



Manual Soldering

Fix the component by first soldering two diagonally-opposite end leads. Use a low voltage (24 V or less) soldering iron applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300°C.

When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320°C.

Suitability of Surface Mount IC Packages for Wave and Reflow Soldering Methods

| Pookago | Soldering Method | | |
|---|-----------------------------------|-----------------------|--|
| Package | Wave | Reflow ⁽¹⁾ | |
| BGA, HBGA, LFBGA, SQFP, TFBGA | Not suitable ⁽²⁾ | Suitable | |
| HBCC, HLQFP, HSQFP, HSOP, HTQFP, HTSSOP, HVQFN, SMS | Not suitable | Suitable | |
| PLCC (3), SO, SOJ | Suitable | Suitable | |
| LQFP, QFP, TQFP | Not recommended ⁽³⁾⁽⁴⁾ | Suitable | |
| SSOP, TSSOP, VSO | Not recommended ⁽⁵⁾ | Suitable | |

Notes

- 1. All surface mount (SMD) packages are moisture sensitive. Depending upon the moisture content, the maximum temperature (with respect to time) and body size of the package, there is a risk that internal or external package cracks may occur due to vaporization of the moisture in them (the so called popcorn effect).
- 2. These packages are not suitable for wave soldering as a solder joint between the printed-circuit board and heatsink (at bottom version) can not be achieved, and as solder may stick to the heatsink (on top version).
- 3. If wave soldering is considered, the package must be placed at a 45° angle to the solder wave direction. The package footprint must incorporate solder thieves downstream and at the side corners.
- 4. Wave soldering is only suitable for LQFP, TQFP and QFP packages with a pitch equal to or larger than 0.8 mm; it is definitely not suitable for packages with a pitch equal to or smaller than 0.65 mm.
- 5. Wave soldering is only suitable for SSOP and TSSOP packages with a pitch equal to or larger than 0.65 mm; it is definitely not suitable for packages with a pitch equal to or smaller than 0.5 mm.







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